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English Version

Source: AT&T Bell Laboratories

Title: Progress on Inverse DCT Accuracy Specifications

This document contains copies of correspondence between parties interested in chip manufacture, algorithm simulation and flexible hardware experimentation.

From:
CCITT SGXV
Working Party XV/1
Specialists Group on Coding
for Visual Telephony

January, 1988

To: Respondents on Designs for DCT Chips

Thank you for your submissions and revisions on DCT and inverse DCT chip design information. We are endeavoring to design our videoconferencing codec and to specify requirements for transforms in such a way that all of your chips will be suitable for this application. However, we cannot at this time guarantee it.

Thus, there are several points that need some further study and clarification. In particular, we would like to run your 8 x 8 inverse DCT (12 bits in, 9 bits out, 2's complement signed) in our video codec computer simulations. Therefore, could you send computer source code that duplicates the digital input-output behavior of your 8 x 8 separable, inverse DCT chip? We would need this within two weeks to meet our very stringent deadlines. Could you send the code via PC floppy, 9-track mag tape or electronic mail to

Barry G. Haskell
Room 4C-538
AT&T Bell Laboratories
Holmdel, NJ 07733
Tel: 201-949-5459
Fax: 201-949-3697
E-Mail Address: ucbvax!vax135!bgh

He will then distribute the software to those with video codec simulation capability so that we can adjust our codec parameters accordingly. Thank you in advance for your cooperation.

On a related topic...Would you be willing to participate in a standards setting procedure for DCT/IDCT chips for more general application, including still-image data compression, medical imaging, aerial photography, etc. We may be able to organize something through the IEEE Standards Committee. If so, please let Barry Haskell know.

TO: Dr. Barry Haskell
AT&T Bell Labs
Room 4C 538
Crawfords Corner Road
Holmdel, New Jersey 07733

Fax: +1 201 949 6172
/3697

FROM: David Hein
Compression Labs, Inc.
2860 Junction Avenue
San Jose, CA 95134

Phone: +1 408 435 3000
Fax: +1 408 922 5429
Telex: 171619

DATE: February 19, 1988

Dear Barry:

You will find attached the results of the IDCT mismatch experiment run at 320 Kbps. Also attached is a possible candidate for the test vectors to be used for testing the various IDCT algorithms. Figure 1 shows the coding performance of RM4 as a function of refresh period and mismatch error. The figure gives the results for both 320 Kbps and 768 Kbps. The 768 Kbps results were also given in the CCITT document #281, excluding the $N=10$ case.

The following conclusions can be drawn from Figure 1:

1. The relative loss in quality is greater for 768 Kbps than it is for 320 Kbps. For the case where $N=12$, there is a 1.4 dB loss for 768 Kbps, and a 0.3 dB loss for 320 Kbps.
2. The optimum refresh period is shorter for 768 Kbps than for 320 Kbps. When $N=12$ the optimum refresh period is around 4.5 seconds for 768 Kbps, versus 8 seconds for 320 Kbps.
3. The coding performance is more sensitive to the refresh period for 320 Kbps than it is for 768 Kbps. With a $MME=0$ and a refresh period of 2 seconds there is a 0.5 dB loss at 768 Kbps versus a 0.9 dB loss at 320 Kbps.

Based on the above conclusions it would seem that the standard could allow for the refresh period to be a function of rate, longer for low rates and shorter for high rates. This probably complicates things a bit too much however. I would be more inclined to retain the requirement that was put forth in document #281. In document #281 the refresh period is defined in terms of actual coded frames instead of absolute time. In a simple implementation the refresh period would be 61 divided by the frame rate. At very high channel rates, such as 1.536 or 1.920 Mbps, the coded will probably run at 30 frames per second, and therefore the refresh period will be 2 seconds. However, at 384 Kbps codecs

will probably be operating at 15 frames per second, and have a 4 second refresh period. If the Nx384 algorithm is applied at 64 Kbps, the codec will probably have to operate at 10 frames per second or less. The refresh period would be greater than 6 seconds.

In addition, the proposal in document #281 relaxes the refresh requirement for the uncoded blocks, since these blocks do not contribute to the mismatch error. In theory, a fixed block never requires refresh. In practice some refresh is required to handle transmission errors. By relaxing the refresh requirements for fixed blocks we have helped the low channel rate even more, since there will be a greater percentage of fixed blocks at low channel rates than at the higher rates.

Taking a second look at the curves in Figure 1, it seems that the refresh requirement could be relaxed to 120 frames, 8 seconds at 15 frames per second, without very much loss in performance. For low MME cases there is actually an increase in performance.

Now on the subject of the IDCT test vectors. Table 1 shows my initial cut at defining the test vectors. As I mentioned over the phone, these were derived by taking a 32x32 window out of one of the frames from the Salesman sequence. The first 16 blocks in Table 1 were generated by breaking the 32x32 image up into 16 blocks, taking the FDCT, and rounding to 12 bits. Blocks 17 to 32 were generated by taking the frame difference of the 32x32 image with the next frame and then producing the 12 bit coefficients as described for the first 16 blocks.

I have generated MME's based on these test vectors and got results similar to those obtained with a random number generator. Therefore, rather than proposing the test vectors in Table 1, I would propose the following method for generating the test vectors.

1. A psuedo-random number generator is used to generate approximately uniformly distributed, uncorrelated 9 bit numbers.
2. The DCT coefficients are produced by performing a forward DCT using a 32 bit floating point implementation.
3. The coefficients are rounded to 12 bits.

Random number generators can be well defined and I would suggest using one based on the linear congruential method, which is defined as:

$$X(n+1) = (A * X(n) + C) \text{ mod } D$$

Values of $A=65539$, $C=0$, $D=2^{**}32$ and $X(1)=65539$ are used in the well known RANDU function. I have used values of $A=21677$, $C=19117$, $D=2^{**}15$ and $X(1)=31415$.

If this is acceptable then the problem of defining the MME arises. The MME as defined in document #281 is simply the MSE of the 9 bit results produced by the baseline IDCT and the target IDCT. Two other potential measures could be the maximum absolute error and the mean error.

My feeling at this point is that there should be 5 parameters that are measured. Based on these measurements, the standard could then specify limits on one or more of the measurements. I suggest that the following parameters be measured:

1. The MSE averaged over all pixels.
2. The maximum MSE, which is computed by measuring the individual MSEs for each of the 64 pixels, and selecting the largest MSE.
3. The mean error averaged over all the pixels.
4. The maximum mean error, which is computed by measuring the individual means for each of the 64 pixels, and selecting the one with the largest absolute value.
5. The maximum absolute error.

I would suggest using a 32 bit floating-point implementation of the IDCT as the baseline, instead of the one I proposed in document #281.

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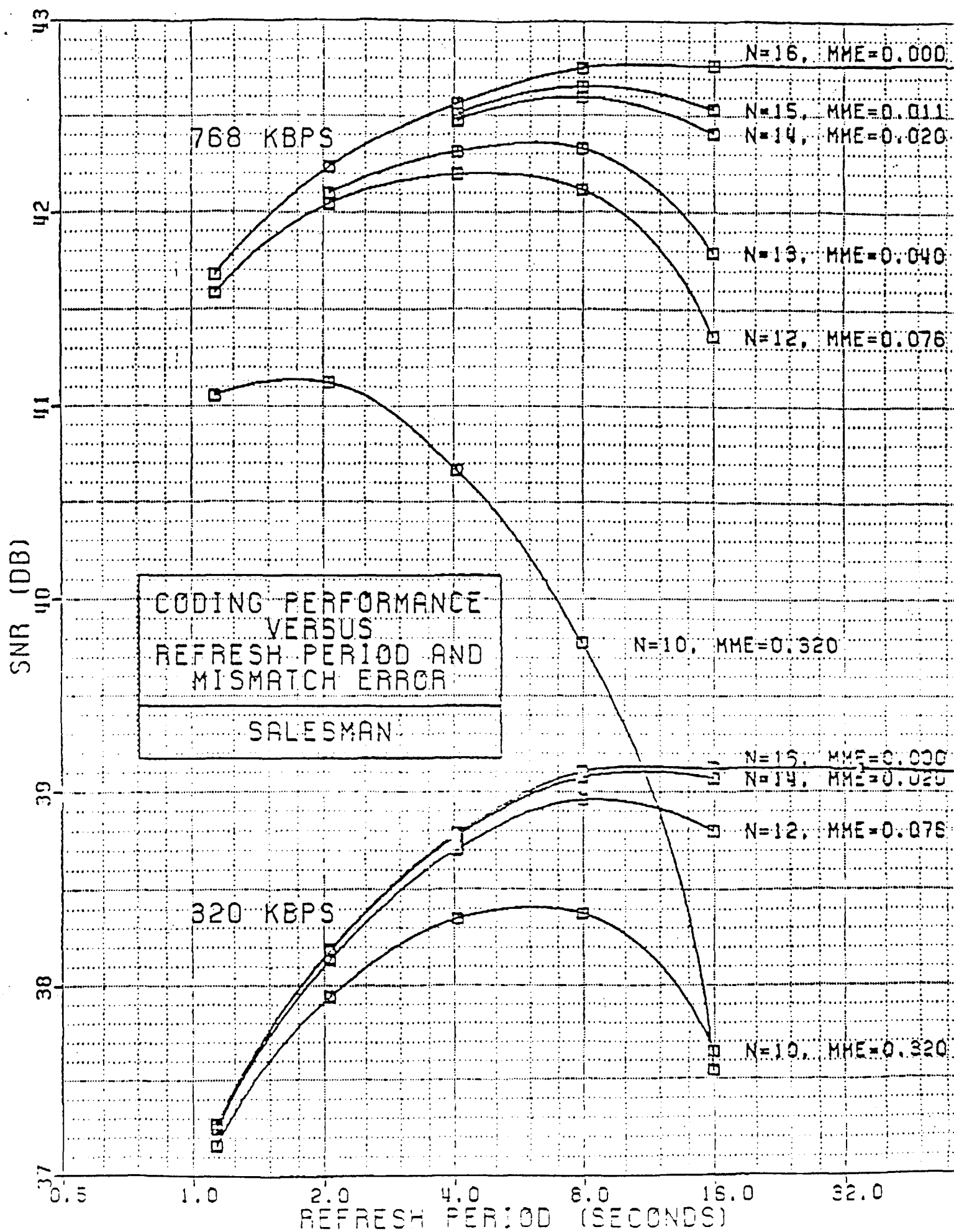


Figure 1

	C ₀₀	C ₀₁				C ₀₆	C ₀₇	C ₁₀	C ₁₁						C ₁₃	
#1	107 000 FFF FFF	00A 000 000 000	008 000 000 000	005 000 001 001	004 001 000 FFF	004 001 001 FFF	002 000 000 001	001 000 000 001	001 FFF 000 001	FFF 000 000 000	FFF 000 001 FFF	FFE 000 000 000	FFE 000 000 000	FFE 000 000 000	FFF 000 000 000	
#2	1D3 002 002 FF8	F7E FFD FFF 005	FF0 FFD 001 FFF	000 003 FFE 001	003 003 000 000	FFD 003 000 003	004 FFF 000 FFE	FFF 003 000 007	004 002 FFE 000	000 FFD 000 001	FFF 002 001 000	001 001 000 000	004 FFF 001 000	FFC 001 000 000	002 000 000 000	000 001 000 000
#3	309 003 FFD FF8	FD9 001 004 002	FF8 FFC 001 FFF	FF5 002 FFF 002	FFD FFE FFF 001	FFA 002 000 004	002 000 FFE 000	FFB 005 001 00B	FEF 004 002 FFE	00E FFC 000 001	002 001 FFF 000	FFF 003 000 000	FFE 001 001 000	FFE 000 000 000	000 FFF 001 001	FFE 001 000 000
#4	398 FF7 001 FF6	FDB FFD FFB 005	003 004 FFD 003	FF6 FFF 000 001	FFB FFE 000 001	001 001 001 003	FFF 002 000 000	FFE 006 000 00B	FE6 004 005 FFF	FFE FFC 002 002	001 002 FFD PFE	FFB 002 FFF 000	001 001 000 FFF	FFD 002 000 000	000 000 000 000	FFE FFF 000 000
#5	172 000 015 009	04B FE0 00B 001	007 FE6 00A 001	FF6 002 00A FFF	FFE 009 002 FFF	FF9 FFA FF6 FFD	004 FFD FFD 001	FFD FFD FFE FFC	F9E 010 FE7 001	FD6 01D FF0 004	010 007 FFA 000	011 FF3 FFF FFE	004 FF5 003 003	00D 006 006 000	001 006 000 000	004 002 001 004
#6	213 FFC 004 004	FE5 FEE FFE FFD	FD9 011 002 000	FE5 003 005 002	FF3 008 FFD 000	FFD 001 000 002	001 000 001 FFC	FFD 000 FFC FF9	FF5 FF1 FFC FF9	FBB 00F FFD 002	FFC FF6 001 002	FFC 008 FFD 000	006 004 002 000	002 FFD FFF 000	000 001 000 FFF	004 FFD FFF 005
#7	150 054 01E 018	FF8 FDA FF3 FFC	023 008 FF7 FFC	FEC FE5 FF8 000	007 006 003 FFD	005 002 FFE FFD	006 001 000 000	FFF 001 FFD FFC	045 FF5 FFD 003	02C FF0 001 FFE	FEA FF0 002 001	01D FF0 FFF 000	FF7 FF0 003 002	FFA FFD FFF 000	FFC FFF 001 001	001 FFC FFF 004
#8	274 073 010 00E	F7A 02C 004 005	030 009 006 001	002 FFA FFD FFF	FFB 001 000 002	FFD 001 000 FFE	008 001 000 000	005 001 FFB FFC	03A 006 000 005	012 003 FFF 006	FD5 000 FFA 002	000 002 000 003	003 FFD FFF FFE	FFD FFF FFF 002	FFF FFF 001 FFF	FFE FFE FFE 007
#9	24C 014 FFD 00B	00E 00B 009 001	FF3 FE2 005 FF9	008 FFC FFF FFE	FFE 00F FFC 001	FEF FF4 FFA FFE	00F FF5 000 000	004 FFD FFA FFC	001 FD4 FFE FFF	022 FFA FFC FFE	00C 00C FF5 002	FFD 00C 000 FFE	001 002 001 FFE	FF8 FF8 000 FFE	FFA FFC FFF 001	FFA 000 002 FFD
#10	200 FE5 FFB 005	FE3 014 002 FFF	000 001 001 001	FFF FFD 000 FFE	FFE FFD FFC FFF	FFE 003 001 000	001 001 000 FFF	FFE 001 000 FFB	FD7 FF0 FFA 007	024 000 002 FFD	000 FFF 000 000	FF8 000 001 000	FFF FFC FFF 000	FFE 000 001 FFD	000 002 FFF 000	000 000 003 FFD
#11	34D FF0 FFE FF8	FF6 PDF FF3 006	00E 008 001 004	FFC FF3 FF1 FFB	002 003 002 002	004 002 FFF FFE	FFF 003 003 000	003 FFF FFB FFC	FFB FE4 FF1 FFF	FE4 FFF 004 004	012 FF3 005 001	006 000 001 000	005 000 002 000	FFF 002 000 000	000 002 000 000	000 004 003 FFB

Table 1 IDCT Test Vectors

#12	320	FC5	063	FED	FFD	006	002	FFD	FDB	002	002	FF2	006	006	FFA	002
	001	FF2	00D	007	009	FFC	002	002	000	FFE	007	002	FFE	FFD	007	001
	FFE	000	004	005	002	003	000	FFD	FF8	FFF	000	001	FFD	FFF	002	003
	001	FFC	001	002	FFF	FFF	001	FFC	003	FFF	000	FFF	FFF	FFE	001	FFA
#13	27B	01B	FE4	FF3	004	FE9	FFE	00A	FF3	FF7	012	019	006	FFC	00B	002
	FF5	FF8	FFA	FFF	FFD	FFC	FFP	006	010	00A	004	005	004	005	FFF	FFF
	00C	000	FF2	FF7	003	003	000	000	005	002	FFB	FFC	FFF	000	FFF	001
	FF3	002	004	002	FFD	002	FFD	008	FFC	FFF	FFF	FFF	FFF	000	001	001
#14	296	FCA	015	FF7	000	000	001	000	017	FE9	000	005	000	004	FFF	000
	007	FFA	FFD	005	000	003	FFE	003	006	FFD	001	000	FFF	000	001	000
	003	FFE	002	001	FFF	000	000	000	000	002	FFF	000	FFF	FFF	000	000
	FF5	002	FFF	002	000	003	FFF	009	FFF	000	001	000	000	000	FFF	000
#15	354	001	FE9	000	FF4	001	FFF	000	025	FFB	00D	FFC	007	FFE	FFF	FFF
	FF7	000	002	001	FFE	FFF	002	006	004	FFE	003	000	002	001	FFF	000
	002	001	FFF	FFF	001	000	000	FFF	FFF	FFF	FFF	000	000	FFF	000	000
	FF6	002	000	004	000	003	000	00A	001	FFF	000	000	000	FFF	000	001
#16	32B	F7C	02E	010	002	013	FEC	00B	009	016	018	001	FFD	006	007	004
	FEF	00C	007	001	FFC	FFB	008	FFF	00D	FFD	FFF	002	001	004	FFC	001
	FF9	005	FFE	FFF	002	FFF	001	FFF	003	FFE	000	FFF	000	000	000	FFF
	FF8	001	000	002	000	005	FFF	00A	FFF	000	002	000	000	FFF	001	FFF
#17	000	FFF	000	001	002	001	001	000	FFE	FFF	FFF	001	000	FFF	FFF	FFE
	000	000	FFE	FFF	000	000	000	FFF	002	001	FFF	FFF	FFF	001	FFF	001
	FFF	FFF	001	002	000	001	000	000	000	001	000	001	000	000	000	000
	001	001	000	000	FFF	FFF	FFF	001	000	001	000	001	FFF	000	000	000
#18	012	000	FF9	FFF	FFF	FFD	001	000	00F	FFB	FF9	FFE	008	FFE	004	000
	FFC	FFA	005	003	000	025	FFD	001	007	FFE	000	FFE	FFF	001	FFF	000
	004	020	001	FFE	000	000	FFE	001	FFD	000	003	FFF	001	000	000	000
	FFF	001	FFE	000	001	000	002	000	FFE	000	001	001	001	000	000	000
#19	017	FFE	004	FFF	002	FFE	002	FFD	FE4	01C	008	000	FFE	FFC	002	FFC
	022	FE6	FFC	004	FFF	001	FFF	000	FF2	005	007	003	FFF	FFE	FFF	FFF
	003	003	FFE	FFE	FFE	000	FFE	002	FFC	002	FFF	000	001	000	001	000
	004	001	FFE	FFF	001	001	000	000	FFE	000	FFF	001	000	000	000	000
#20	010	000	004	FFE	FFE	000	001	000	FE0	FF4	000	FFE	FFF	FFF	001	FFA
	02B	001	007	FFD	000	000	001	003	FF1	FF6	004	004	000	001	000	FFF
	008	FFB	FFB	001	000	000	000	000	002	FFE	FFD	FFE	001	001	000	000
	001	005	FFD	000	000	000	001	000	FFF	001	FFE	FFF	FFF	002	000	FFF
#21	FCF	FFE	00D	FFF	006	006	002	003	0CE	FE3	FF6	00E	002	FFC	FFE	FFC
	007	004	FE1	FFC	003	007	001	FFE	027	02D	022	FFC	FFE	003	004	003
	004	002	008	015	006	FF2	FF4	000	FF4	FF1	FF3	FF4	004	004	004	001
	00E	007	000	000	FFD	FFC	001	000	00E	00C	003	FFE	FFE	001	001	FFF
#22	FE9	FFC	000	004	FFD	001	002	003	01A	FD8	015	FFF	000	000	000	001
	FE0	010	000	002	000	FFD	FFC	FFF	000	010	FEF	010	FFE	FFF	000	FFF
	FFD	003	002	001	FFA	004	FFE	001	000	000	001	FFC	001	FFC	FFF	001
	FFE	004	FFF	004	FFF	002	001	000	FFD	001	FFE	000	000	000	000	FFF

Table 1 (continued)

#23	FFE	FFC	FFC	001	FEB	003	004	001	0C1	FF7	FF4	008	007	FF9	FFE	000
	025	FC6	010	FDA	014	006	001	003	FF9	003	FEF	016	FE3	000	FFA	FFD
	005	FE8	FEE	FF2	00A	000	004	000	FED	00E	006	004	004	FFE	001	FFE
	021	FEA	FFD	FFE	FFE	003	FFE	FFF	FF4	009	FFF	FFF	003	FFE	000	000
#24	021	FCC	FD7	00E	FEA	FF1	FFA	009	0AB	058	FED	FF3	013	FFE	003	FFE
	051	027	023	FFF	002	00F	007	002	FE5	FF3	FF1	FFB	FFA	FF9	FFB	002
	FFF	005	00F	FFF	006	004	001	FFF	001	FFD	FF8	FFF	FFF	000	000	000
	00D	009	003	001	003	001	001	000	000	FFF	003	000	FFE	001	000	001
#25	FF6	00C	000	FFB	00C	FF5	FF7	001	00B	00B	FF1	001	009	FF4	FFD	000
	FED	FFB	FF7	006	00A	FFC	FFC	FFE	FC4	FFE	02A	00E	FFA	FFC	001	FFF
	008	004	FF2	FF8	FFE	001	003	001	00E	FFD	FF1	002	004	001	002	000
	00D	001	FF2	FFA	001	004	002	001	FFE	002	001	FFD	FFF	003	001	000
#26	000	00E	FF6	003	000	FFF	003	001	FD0	021	005	FF9	001	002	FFE	002
	FF2	007	000	FFF	FFA	003	000	000	FF3	009	FFF	002	FFA	000	002	000
	FFD	FFD	003	002	FF9	003	FFE	001	FFE	001	FFF	000	001	001	001	000
	004	FFD	001	FFF	FFE	001	FFF	001	001	FFE	001	001	FFF	FFF	FFF	000
#27	FD6	FF0	00C	FFD	009	007	001	006	FD5	FD4	009	FF2	003	006	001	002
	00C	FD9	002	FF3	004	004	005	001	00E	FDC	FF9	FF0	005	006	003	002
	006	FF2	FFD	FF1	000	001	003	000	FFA	003	001	FF4	001	003	000	001
	FF8	00A	003	FF6	002	000	000	000	FF8	006	003	FFA	001	FFF	000	002
#28	FFF	FD3	00D	00F	FEF	00A	002	FFF	FEB	FF4	014	000	003	003	000	FFE
	012	FFB	00F	005	FFF	003	000	001	008	FFA	008	000	000	004	006	001
	FFB	007	006	008	004	005	FFE	002	FFB	FFC	001	002	FFB	FFD	003	001
	FFC	FFF	000	002	FFD	001	001	002	FF9	001	FFF	002	001	000	001	001
#29	009	00B	009	003	001	FFD	FFB	004	FF1	FE9	FEF	FFD	011	003	004	002
	015	012	000	001	FF9	FF9	FFC	003	008	FFD	FF5	FFE	00D	00B	FFD	FFE
	010	FFF	FE8	FEE	FFF	001	000	001	000	001	001	FFF	FFD	000	FFF	000
	FF7	FFE	009	002	FFC	000	000	002	FFA	002	004	000	FFD	FFE	000	000
#30	015	FEC	006	005	FFC	004	FFE	001	013	FEB	FFC	005	FFF	005	FFE	FFF
	005	000	FFB	002	FFC	002	000	001	005	FFD	002	000	FFD	FFF	001	FFF
	002	FFE	002	001	000	000	000	000	FFE	001	FFF	FFF	FFF	000	000	000
	FFF	FFF	FFF	000	000	000	000	000	FFE	FFF	000	000	000	001	FFF	001
#31	010	012	000	008	000	FFD	001	000	004	004	007	FFD	002	001	000	FFF
	FFB	FF6	007	FFD	FFF	FFE	000	001	001	000	002	000	004	FFF	FFF	FFF
	002	003	FFD	FFF	001	000	000	000	000	FFD	FFE	000	001	FFF	000	FFF
	000	001	000	003	FFF	000	000	FFF	000	FFF	FFF	000	FFF	000	001	000
#32	051	FE3	FDA	00C	FF5	024	FDE	009	FB2	021	019	004	FE9	008	006	006
	024	FF0	FFA	002	003	FFC	005	FFE	004	FFC	005	006	005	001	FFD	001
	FEF	008	FFB	FFF	001	FFE	003	FFF	008	001	FFF	FFE	002	000	000	FFF
	FFE	FFD	002	FFF	000	000	FFF	000	FFF	003	001	FFF	FFE	FFF	001	FFF

Table 1 (continued)

From: Barry G. Haskell
AT&T Bell Labs
Holmdel, NJ 07733
USA

phone: +1 201 949 5459
fax: +1 201 949 3697

To: Mr. T. Koga
Second Development Department
Transmission Division
NEC Corporation
1753, Shimonumabe, Nakahara-ku
Kawasaki City 213 JAPAN

phone: +81 44 433 1111
fax: +81 44 433 5214

Dear Mr. Koga,

Recall our lengthy discussions on the problem of mismatch between the coder and decoder inverse transform. You said you might be able to carry out some mismatch experiments on the flexible hardware. In particular, the decoder inverse transform coefficients could be altered or truncated somewhat to simulate a mismatch. Then the effect on picture quality of various refresh rates could be studied.

Have you been able to make any progress on this problem? We would only need to know what you did to the inverse transform coefficients and what refresh rate you needed to maintain good picture quality. We could then compute the Mean Square Mismatch error using random data and specify the allowable MME for 384 kbs.

I have talked with David Hein, and his preliminary results are encouraging. Some mismatch at 384 kbs seems allowable, as long as refresh occurs at a reasonable rate. However, it would be much more satisfactory to have some real-time hardware results to back it up.

Operation at 64 kbs seems to be no problem.

I await your reply, and thank you in advance.

Sincerely,

Barry Haskell

To Mr. B.G. Haskell
AT&T Bell Labs
Holmdel, NJ 07733

Phone +1 201 949 5459
Fax +1 201 949 3697

From Toshio Koga (22-3130) Phone +81 44 433 1111 Ex 3134
Second Development Department Fax +81 44 433 5214
Transmission Division
NEC Corporation
1753 Shimonumabe, Nakahara-ku
Kawasaki City 213

Subject IDCT Mismatch Problem (Two pages including this.)
Date March 1, 1988

Dear Mr. Haskell:

Thank you for your facsimile letter on the problem of mismatch between the coder and decoder inverse transform.

Using our Flexible Hardware, we observed reproduced sequences when the accuracy of IDCT in the decoder loop was changed. As seen from the attached bit diagram, truncation at ① ($2^{**}-5$), ② ($2^{**}-4$), and ③ ($2^{**}-1$) was carried out. The first two approximations were made to the output data between the first and the second 1D-IDCT. According to the current Flexible Hardware specification, it has a 16-bit (11.5) accuracy.

Truncation at $2^{**}-5$ (①) hardly produces visible degradation for teleconference video sequences even ten seconds after the truncation gets started. Although truncation at $2^{**}-4$ (②) affects picture quality about several seconds after, rounding at $2^{**}-4$ will produce no visible distortion within ten seconds according to our previous simulation work. Rounding at $2^{**}-4$ will be simulated soon to make it sure. Truncation at $2^{**}-1$ (③) gives no good result.

In this experiment, we did not change matrix element values, since they do not affect the experimental results as long as their accuracy is higher than, for instance, 12 bits (currently 16 bits)

At the moment, the experimental results give us a feeling that truncation at $2^{**}-5$ for the output data from the first IDCT or its equivalent approximation accuracy by some IDCT algorithms other than the matrix approach will be possible, and in addition rounding gives a better result.

Refresh cycle may be about ten seconds or longer.

Note : Approximation should carefully be made to sequences with highly saturated color, for instance, Color Bar.

cf. Doc #142

(1)

TRANSFORM DOMAIN INPUT

MATRIX FROM ELEMENT 1ST 1-D

PRODUCT OF ABOVE TWO

SUM OF EIGHT PRODUCTS

OVERFLOW CHECK BITS

1ST 1-D OUTPUT

MATRIX FROM ELEMENT 2ND 1-D

PRODUCT OF ABOVE TWO

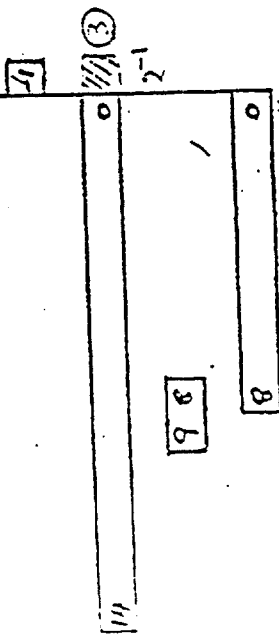
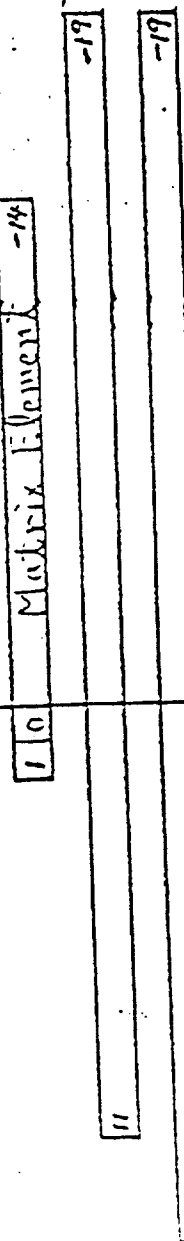
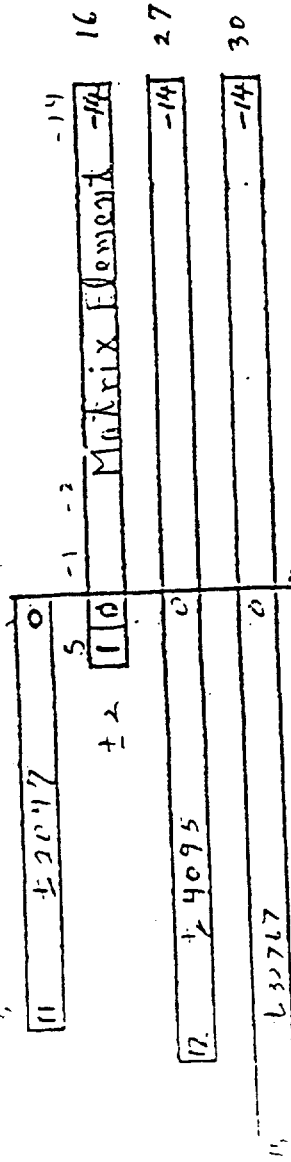
SUM OF EIGHT PRODUCTS

ROUNDING DETERMINATION BIT

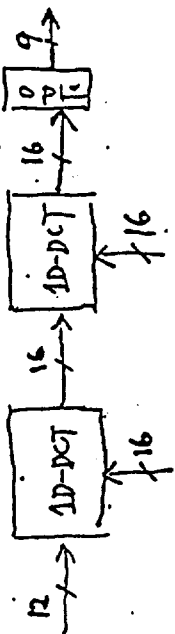
INTEGER VALUE

OVERFLOW CHECK BITS

REL DOMAIN OUTPUT



BINARY POINT



2-D INVERSE TRANSFORM BIT SPANS

From: Barry G. Haskell
AT&T Bell Labs
Holmdel, NJ 07733
USA

phone: +1 201 949 5459
fax: +1 201 949 3697

To: Mr. T. Koga
Second Development Department
Transmission Division
NEC Corporation
1753, Shimonumabe, Nakahara-ku
Kawasaki City 213 JAPAN

phone: +81 44 433 1111
fax: +81 44 433 5214

Dear Mr. Koga,

Thank you for your results on mismatch experiments.

We did corresponding computer simulation noise measurements using random pel data, and got the following results:

Operation	Maximum pel MME	Overall MME
1. Truncation to 15-bits	0.083	0.011 OK
2. Truncation to 14-bits	0.116	0.026 Not OK
3. Rounding to 14-bits	0.066	0.021 OK

I assume your studies were using 320 kbs for the video. Was there any systematic refresh during the measurements? We assume not.

On the basis of information from you and David Hein, I sent the attached letter to chip manufacturers. We will hear their response. If you have any further thoughts or results, please let me know.

I await your reply, and thank you in advance.

Sincerely,

Barry Haskell

To Mr. B.G. Haskell Tel +1 201 949 5459
AT&T Bell Labs Fax +1 201 949 3697
Holmdel, NJ 07733

From Toshio Koga (22-3130) Tel +81 44 433 1111 Ex3134
Second Development Department Fax +81 44 433 5214
Transmission Division
NEC Corporation
1753, Shimonumabe, Nakahara-ku,
Kawasaki City, 211

Subject IDCT
Date March 16, 1989

This page only.
(~~pages including this.~~)

Thank you for your letter on Proposed Specifications.
I would like to make a few comments on that.

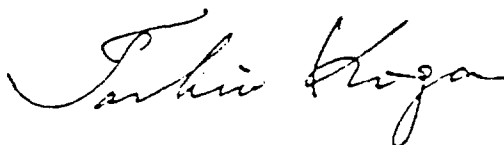
1. In our hardware measurements, transmission bit rate for video was 320kb/s. Reproduced sequence was obtained without cyclic refresh. Immediately after fast updating sequence in Intra-mode, truncation was carried out and then reproduced picture was observed until visible distortion appeared.
Your assumption is correct.
2. Distortion is much more visible for highly saturated pictures or parts of picture such as Color Bar.
We should be careful about this result.
3. In practical IDCT calculation, we can distinguish two kinds of distortion, one being due to the approximation (rounding to 9 bits) and the other being due to mismatch. Base-line approach corresponds to a certain approximation.
This will lead us to use real number computation to obtain "reference" as you did.
4. Two examples of truncation to 14 and 15 bits will be demonstrated using VTR in the Hague meeting.

Unfortunately, I can not attend at the Hague meeting. Mutsumi Ohta, who participated in the Tokyo meeting last January, will act for me.

I hope we can have happy results.

Best regards.

Toshio Koga



From: Barry G. Haskell
HO 4C-538
AT&T Bell Labs
Holmdel, NJ 07733
USA

phone: +1 201 949 5459
fax: +1 201 949 3697
telex: 219879 BTLH UR

To: Respondents to CCITT Request for DCT

Dear Sirs and Mesdames:

Thank you for your responses so far to the CCITT request for DCT chip information. Based on software received as well as computer simulations and hardware codec studies, we think we are near to a performance specification that would work for ISDN video codecs. We are thinking of a procedure (see attached) that would be suitable for computer simulation.

Could you examine the proposed method and let us know your thoughts? The peak error constraint is fairly firm, whereas the mean-square error constraints may possibly be relaxed slightly as we gain more experience with the codecs.

We will need your responses by March 18 at the latest. If I do not yet have your fax number, could you send it? Thank you.

Sincerely yours,

Barry Haskell

CCITT Specialists Group on Visual Telephony

Proposed Specification for Inverse DCT Chips

1. Generate random integer pixel data values in the range -256 to +255. Arrange into 8x8 blocks.
2. For each 8x8 block, perform a separable, orthonormal, matrix multiply, Forward Discrete Cosine Transform (FDCT) using at least 32-bit floating point accuracy.
3. For each block, round the 64 resulting transformed coefficients to the nearest integer values. Then clip them to the range -2048 to +2047. This is the 12-bit input data to the inverse transform.
4. For each 8x8 block of 12-bit data produced by step 3, perform a separable, orthonormal, matrix multiply, Inverse Discrete Cosine Transform (IDCT) using at least 32-bit floating point accuracy. Round the resulting pixels to the nearest integer, and clip to the range -256 to +255. These blocks of 8x8 pixels are the "reference" IDCT output data.
5. For each 8x8 block of 12-bit data produced by step 3, use the proposed IDCT chip or an exact-bit simulation thereof to perform an Inverse Discrete Cosine Transform. Clip the output to the range -256 to +255. These blocks of 8x8 pixels are the "test" IDCT output data.
6. For each of the 64 IDCT output pixels, measure the peak and mean square mismatch error between the "reference" and "test" data.
7. For any pixel, the peak error should not exceed 1 in magnitude. For any pixel, the mean square error should not exceed 0.06. Overall, the mean square error should not exceed 0.02.



**Bell
Communications
Research**

331 Newman Springs Road
Box 7020
Red Bank, New Jersey 07701-7020
(201) 758-2000
Rm. 8X-301
Ext. 2889
Fax: 758-0889

FAX: 949-3697

March 15, 1988

Dr. B. G. Haskell
AT&T Bell Labs
HO 4C-538
Crawfords Corner Road
Holmdel, New Jersey 07733

Dear Dr. Haskell:

Thank you for sending us the "Proposed Specification for Inverse DCT Chips." Following are our comments to the enclosed document

1. Item 7, line 2 - Shouldn't the "pixel" be "block"?
2. Shouldn't the random number generator algorithm be specified?
3. How many blocks of random numbers should be simulated?
4. Shouldn't the speed requirements of the chip be specified?
5. The mean-square error specification may be relaxed a little bit

Please call me if you have any further questions.

Sincerely,

Ming-Ting Sun
Member of Technical Staff
Signal Processing Systems Group

Attachment

CNTR NO: 2631

To: B G Haskell
AT&T Bell Labs
Holmdel, NJ
USA

From: Colln Smith, Inmos, Bristol, UK
cc Peter Cavill

Date: 18th March 1988.

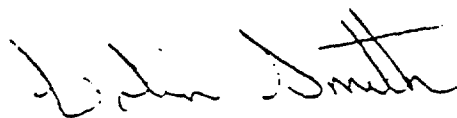
Subject: Proposed Specification for Inverse DCT chips

Dear Mr. Haskell,

Thank you for your specification of a procedure for the DCT coding. Our simulations are broadly in line with the proposed method and any minor differences will be corrected in line with your fax, so that we can estimate the error values for the random integer pixel data input.

However in order to compare resultant error rates it is important to know if any scaling has been applied to the coefficients when performing the DCT. We think that this information should be added to the proposed specification.

Regards,



Manager, DSP Products.

FAX to: +1 201 949 3697

Dr. B. G. Haskell

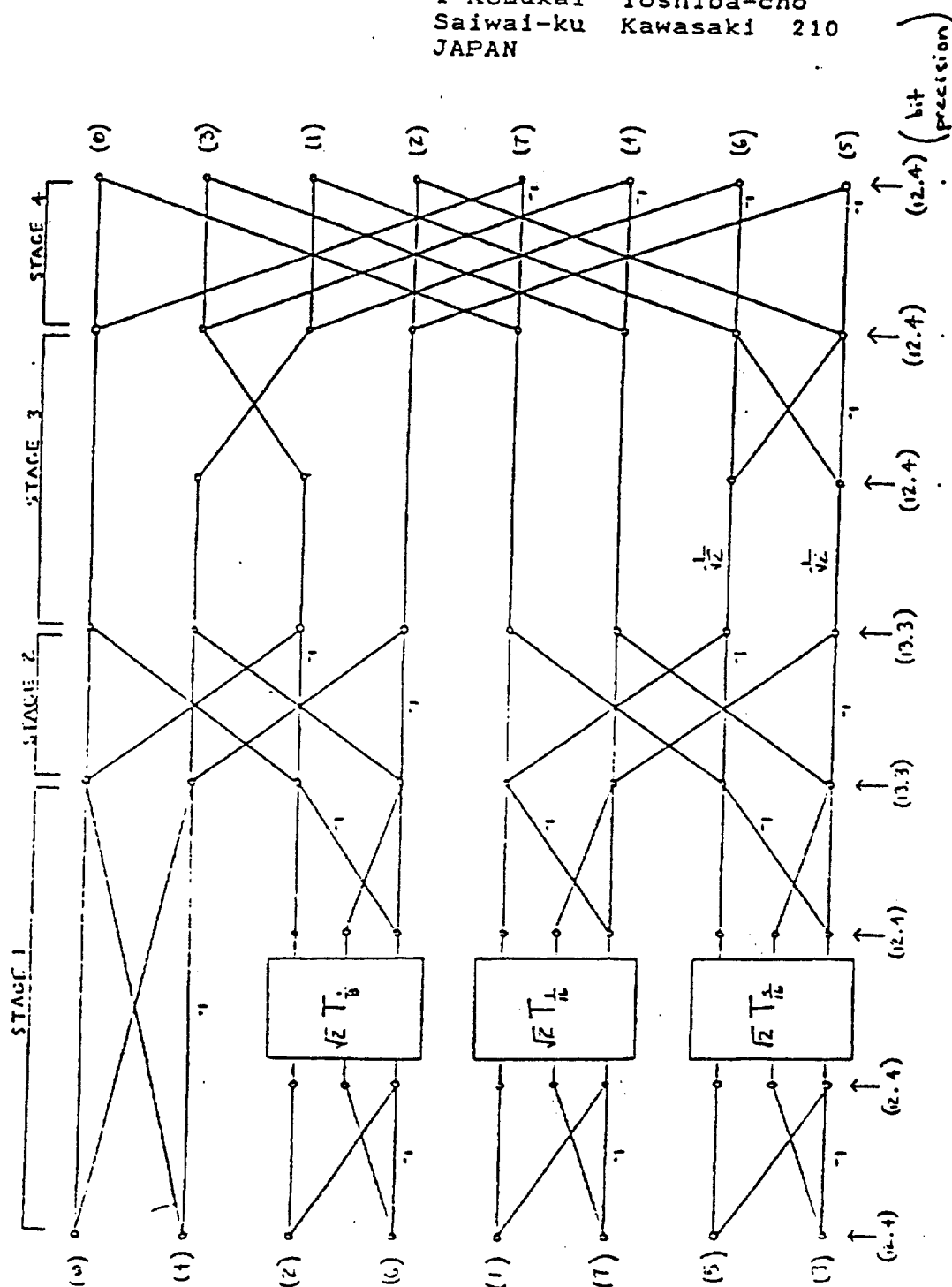
HO 4C-538 AT&T Bell Labs

Holmdel, NJ 07733

USA

Video Sys. & Tech. Lab.
R & D Center TOSHIBA
1 Konukai Toshiba-cho
Saiwai-ku Kawasaki 210
JAPAN

FIGURE 1: FAST INVERSE COSINE TRANSFORM (1-DIM)



FAX to: +1 201 949 3697

Dr. B. G. Haskell
HO 4C-538 AT&T Bell Labs
Holmdel, NJ 07733
USA

Video Sys. & Tech. Lab.
R & D Center TOSHIBA
1 Komukai Toshiba-cho
Saiwai-ku Kawasaki 210
JAPAN

NUMBER OF INPUT DATA BLOCKS: 20000

PEAK ERROR

1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0

MSE (PIXEL-BY-PIXEL BASIS)

0.04	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.03	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.02	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.01	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.02	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.02	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.04	0.02	0.02	0.02	0.02	0.02	0.02	0.02
0.04	0.02	0.02	0.02	0.02	0.02	0.02	0.02

OVERALL MSE = 0.02

Three of Three



IEEE

CIRCUITS AND SYSTEMS SOCIETY

Ming L. Liou
President

PLEASE REPLY TO:
Bell Communications Research
Room NYC 3X-303
331 Newman Springs Road
Red Bank, NJ 07701 USA
Telephone: (201) 758-2949
Telex: 275 318
FAX: (201) 758-0889

March 14, 1988

Dr. B. G. Haskell
AT&T Bell Labs
Crawfords Corner Road
Holmdel, New Jersey 07733

Dear Barry:

In response to the CCITT Specialists Group on Coding for Visual Telephony's inquiry to encourage manufacturers and IEEE to contribute to making an industry standard for DCT chips, the IEEE Circuits and Systems Society has agreed to undertake such a task.

A Standards Committee, chaired by Dr. Ming-Ting Sun of Bellcore, has been established within the Circuits and Systems Society. The committee will first submit a Project Authorization Request (PAR) to the IEEE Standards Board for approval. After approval of the project, various organizations within IEEE expressing interests will be asked to participate.

Since the CCITT, ISO, and other organizations have already done some work on DCT standards, the Circuits and Systems Standards Committee will rely heavily on these results during the standardization process.

Sincerely yours,

Ming L. Liou

From: Barry G. Haskell
HO 4C-538
AT&T Bell Labs
Holmdel, NJ 07733
USA

phone: +1 201 949 5459
fax: +1 201 949 3697
telex: 219879 BTLH UR

To: F. Molo
Telettra Vimercate

Dear Mr. Molo

Thank you for sending your software for simulating your DCT chip. Unfortunately, I was unable to get it to work properly, I think.

We really only need the inverse transform. Perhaps you could change the software slightly to provide a subroutine

INVDCT(BLOCCI, BLOCCO)

where BLOCCI is an 8x8 input matrix of DCT integer coefficients in the range -2047 to +2047, and BLOCCO is an 8x8 output matrix of pixel integer values in the range -255 to +255.

I am concerned about your statements ...

"sine and cosine of angles different from canonical ones"

"line and column DCT are not exactly the same"

Any chip to be used in a standard codec would have to work with other codecs that do use the canonical, seperable, orthonormal Discrete Cosine Transform.

If you have any more advice, please let me know. Thank you.

Sincerely yours,

Barry Haskell

February 22, 1988

From: Barry G. Haskell
AT&T Bell Labs
Holmdel, NJ 07733
USA

phone: +1 201 949 5459
fax: +1 201 949 3697
/6172

To: Mr. Richard C Nicol
British Telecom Research
Martlesham Heath, Ipswich IP5 7RE
UNITED KINGDOM

phone: +44 473 642710
fax: +44 473 643791

Dear Richard,

I got your Telex of February 19! Our letter (copy attached) went out on February 2 to NCR, INMOS, Thomson, Bellcore, Siemens, Telettra, Philips and AT&T. It went out to Plessy on Feb. 17.

We have received software from Telettra, Bellcore and AT&T. INMOS called two weeks ago and said they would send it. Could you see if that is still their intention? Call Peter Cavil (0454 616616).

I talked with David Hein and sent a fax (copy attached) to Koga at NEC.

I think we'll be able to converge on something fairly quickly.

Thanks,

Barry Haskell

ANT Nachrichtentechnik GmbH - Postfach 1120 - D-7150 Backnang

Dr. B.G. Haskell
AT&T Bell Laboratories
HO 4C-538
Holmdel,
NJ 07733 USA

Copy:
Mr.G.Zedler
Fernmeldetechnisches
Zentralamt, N11
Postfach 50 00
6100 Darmstadt

Ihre Zeichen/Nachricht vom

Unsere Zeichen / Bearbeiter

☎ (07191) 13-0
Durchwahl 13-

Backnang

MX/V,Bö/kl / Börner

2589

11.03.88

CCITT Video Coding Experts Group
Solicitation of information about DCT chips

Dear Sir,

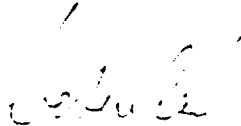
We have been informed about your a.m. request, and we are pleased to address you in this matter.

Our company is engaged since years in advanced techniques for image processing, including elaboration of coding/decoding algorithms allowing transmission bit rates of $n \times 64$ kbit/s. In the context of this work, DCT has been used, and a pertinent semicustom chip is under realisation which could be a candidate solution for a CCITT Video Coding standard. Whilst sample quantities are available on short term, volume production will depend upon market request and could be envisaged for 1989.

As we are interested in video conferencing techniques, we are prepared to support the realization of CCITT standards. Consequently, we very much would appreciate to have your indication along which terms CCITT intends to identify a standard solution out of various candidates. We are willing to consider adequate arrangements including patent and licencing aspects.

We will appreciate your advice how to pursue this matter further. You are kindly invited to contact us if we can assist in CCITT standardisation.

Sincerely yours,


Dr. Helzenbein


Börner