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INTERNATIONAL ORGANIZATION FOR STANDARDIZATION ORGANISATION INTERNATIONAL DE NORMALISATION

ISO/TC97/SC2/WG8

CODED REPRESENTATION OF PICTURE AND AUDIO INFORMATION

ISO/TC97/SC2/WG8 N 636 November 1987

SOURCE: ISO/TC97/SC2/WG8

TITLE: Liaison between ISO/TC97/SC2/WG8 and CCITT SGXV (Mr Okubo)

During the ISO/CCITT Joint Photographic Expert Group (JPEG)[†] meeting, in Washington 26-30 October 1987, it was felt that a formal liaison between the JPEG and the CCITT SGXV working groups was urgently required.

JPEG is currently aiming at a single coding algorithm for still images to be agreed and of January 1988, and as far as JPEG knows, SGXV is also aiming at a recommendation on a unified algorithm for videoconferencing services at 384 kbit/s, during 1988.

At our a June '87 meeting in Copenhagen the three techniques given below were selected for further development +

- 1. ADCT N502
- 2. ABAC N509
- 3. GBTC N510

A full descriptions of the techniques is given in the documents annexed to this liaison letter.

The ADCT technque, N502, in particular might be worth considering for compatibility purposes, as apart from interframe coding we note the following commonalities:

- I. DCT 8x8
- 2. Zig-Zag Scanning
- 3. Uniform quantization
- 4. Thresholding
- 5. Huffman Code Tables

We would like you to consider those common parts to see if for the intraframe coding mode we might be able to harmonize our specifications, which it is felt that would be beneficial to both parties.

*JPEG is the Joint Photographic Experts Group set up by ISO TC97/SC2/WG8 and CCITT SGVIII NIC(JPEG).

INTERNATIONAL ORGANISATION FOR STANDARDIZATION ORGANISATION INTERNATIONALE DE NORMALISATION

ISO/TC97/SC2/WG8



CODED REPRESENTATION OF PICTURE AND AUDIO INFORMATION

ISO/TC97/SC2/WG8

N 640

/adctg N

Source: ISO / ADCTG

Title: Adaptive Discrete Cosine Transform Coding Scheme

for Still Image Telecommunication Services

Dear Hiroshi, please find, herein, the documentation on ISO-ADCT.

It, very likely still remains few errors...

of different kinds. But this is the best issue
we have been able to achieve by

January the 8th.

Very Kind Regards Alain Leger

Copie: Henning POULLEN

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(n. 1920)

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ISO ADAPTIVE DISCRETE COSINE TRANSFORM CODING SCHEME FOR STILL IMAGE TELECOMMUNICATION SERVICES

issued on january, 8th, 1988

Published by ISO/TC97/SC2/WG8 - ADCTG

ABSTRACT

This documentation presents the state of development of the Adaptive Discrete Cosine Transform (ADCT) coding technique, as proposed on january the 8 th, 1988, by ISO - TC97/sc2/wg8/adctg. A previous version said "the june version" or n502, was presented in Copenhagen (june 87) during an ISO - TC97/ sc2/wg8/jpeg meeting.

The main steps forward relate to:

- enhancements of the picture quality at very low bit rate (e.g. 0.08 bpp and 0.25 bpp)
- enhancements of the psychovisual impression of progressive update
- o introduction of predefined Variable Length Code tables as default values of an adaptive entropy coding scheme, in order to gain full symetry of the Codec
- o production of a replica of the original image at about 8.0 bit per pel in a final stage of a progressive or sequential updates

Note, also, that in any case this has not to be considered as a final version, refinements – open or not in a final recommendation – are still possible in some places, but, as expected, without modifications to the basic coding structure.

This documentation is the expression of a collaborative work within the ISO - TC97/sc2/wg8 adct group of experts from AT&T - Bell labs, CCETT, CSELT, DEC, KTAS, NEC and NTT.

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1.0 General

1.1 INTRODUCTION

Numerous ADCT (Adaptive Discrete Cosine Transform) Coding schemes known so far, have always produced very good results at low bit rate and were very often considered as First Class Compression techniques. This has led to a great variety of implementations on a worldwide scale that in turn have entailed much consideration at international standardization and collaboration levels: CCITT, CCIR, CEPT, RACE, ESPRIT,...

As a matter of facts, the ADCT coding technique remains one of the lead – ing coding schemes of the "first generation" for late 80's and very likely the 90's, for very wide applications, from still image services up to live image services (visiophony, visioconference and digital TV).

The technique we are thereafter describing is one of this COMPATIBLE family, specifically designed STILL grey or colour images. Its applicability ranges from low end Photovideotex services, up to Graphics arts, Press and Medical applications.

Special care has been taken to design a SIMPLE compression scheme that proves very EFFICIENT. This of much concern for the implementors of the technique who have to understand and to master EASILY.

Furthermore, the REAL TIME constraints of still image services for the ISDN or higher data rate channel, have also, always been considered, for both the coder and the decoder, and as a consequence the coding technique is fully SYMMETRICAL.

At the editing or coding level, the technique is fully ADAPTIVE to the im – age content and to the SERVICE REQUIREMENTS.

1.2 BRIEFING ON PRINCIPLES OF THE CODING ALGORITHM

TECHNICAL

The technique is simple and is basically funded on the following principles

- 1. NO IMAGE MODELS are asssumed, either in the spatial or the trans-form domain
- 2. PSYCHOVISUAL THRESHOLDS are used for the selection of the most relevant coefficients in the transform domain by implicit thresholding of the lowest decision level .
- 3. LINEAR QUANTIZATION with unit step equals to the PSYCHOVISUAL threshold of the related coefficient to be quantized.

- 4. LOSSLESS coding of the quantized coefficients and their relative positions via predefined Variable Length Code tables.
- 5. ADAPTIVITY is achieved by suitable thresholding matrices for a particuliar set of components (e.g. Y, Cr, Cb) and resolution of the image to be coded, and by carefully predefined or customized Variable Length Code tables

From these foundations, it follows that:

- 6. COMPRESSION is chiefly achieved by a LOSSLESS scheme, applied on very stable statistics in the transform domain
- 7. QUANTIZATION is performed at the nominal picture quality, for mini mizing the VISIBLE Quantization ERROR.

SER VICE

- The PICTURE QUALITY and consequently the COMPRESSION factor is varied upon changing linearly the Psychovisual thresholding matrices one Central point of the algorithm –, which act as adaptive zonal sampl ing and as adaptive quantizing laws.
- FLEXIBILITY for FUTURE improvements is provided very easily and at insignificant cost, by the possible Redefinition of VLC tables and of Thresholding matrices through overhead information e.g. according to progress in psychovisual perception and understanding of image.
- VERSATILITY for not yet defined or unforseen applications, by the same means outlined in (2)
- o The SEQUENTIAL and PROGRESSIVE build up of the received image, are available through the appropriate sequencing of the transformed and quantized coefficients in a large number of display stages as required by the service.
- o For LOW data RATE channels, e.g. 1200 bit/s on PSTN, very low coding rates are available, i.e 0.08 bit/pel,...,0.25bit/pel, through a NON-MANDATORY preprocessing structure that is embedded in the general coding structure, and that caters for good recognisability of the picture at the early stages of the progressive up-date.

1.3 OVERVIEW OF THE FUNCTIONALITIES

1.3.1 Coding Structure for progressive update

1.3.1.1 Introduction

In some applications, low bit – rate coding of images is utilized not to repre – sent pictures in final form, but instead as an interim step in a progressive transmission that allows the viewer to have something to look at far earlier than if the image were sent with final resolution and quality in only one stage of coding. One typical application migth be Photovideotex on narrow – band channels (e.g. 4800 bit/s). Other applications, not constrained by band limitations (e.g. > 64 kbit/s), may desire to display a picture in a more pleasing way than suc – cessive rows of full resolution decoded image (i.e. sequential update), and so to display in a certain number of full scans, giving a better subjective impression.

In the following, these two kinds of aspects are considered and are globally called "Progressive Coding Structure". In fact, it is composed of a "hierarchical structure" that may not be mandatory and a "spectral selection structure" – as previously described in n502 – which may give at each MAIN – stage of the "hi – erarchical structure" a large number of progressive SUB – stages by sending, for example, one coefficient per block at a time.

Finally, in such cases, we may wish to code in the early stages at equivalent bit – rates very much smaller than that required for final picture quality. For ex – ample, progressively better quality color images might be sent, respectively, at 0.08 0.25 0.75 and finally 2.25 bits per pel.

At such low early – stage bit – rates, reproduced subjective image quality of unadorned transform coding is relatively poor compared with the original picture. In fact, a better approach usually is to low – pass filter and subsample the ori – ginal picture before early – stage coding, in order to reduce the number of pels that have to be coded. The transform can then be carried out at a higher bit – rate per transmitted pel, and the image degradation due to low – pass filtering will usually be far more acceptable than the block boundary distortion due to coding at a very low bit – rate per transmitted pel.

This is similar to the approach used in classical pyramid coding. Coding always takes place at a relatively high bit – rate per transmitted pel. The number of pels sent is simply reduced to achieve the desired bit – rate per image.

Having already sent a low resolution version of the picture at one stage, it is not necessary to start from scratch in sending the higher quality picture at the next stage. The earlier stage quantized/decoded image can be up – sampled, with appropriate interpolating filters, and used as a prediction for the coding at the next stage. We simply substract the prediction from the original, and transform code the differential pel blocks as described above.

Note that this is different than traditional pyramid coding, in which an unquantized/uncoded low – pass filtered image is substracted to get the differential image to be transmitted. With this method, coding distortion at an early stage cannot be corrected at a later stage, unlike the presently described technique.

At the earlier stages, the picture quality of pyramid transform coding is noticeably better than with either transform coding. However, at the final stage, there may be discrepancies between this hierarchical coding structure and onepass transform coding at the final bit-rate. Nevertheless, the full match with

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"hierarchical" (i.e. progressive) coding and "one - pass" (i.e. sequential) coding of the same picture is ALWAYS possible, by simply reordering the data in a buffer.

1.3.1.2 Principles of operation of the hierarchical structure

First, the image is low – pass filtered and subsampled, for example, by 4:1 in both directions to give an overall reduction of 16:1 in the number of pels. We call this image I_{16} :



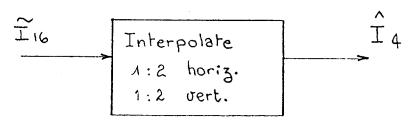
Similarly, we can produce an image subsampled 2:1 in both directions for an overall 4:1 pel reduction. We call this image I_4 :



Alternatively, I_{4} could be filtered and subsampled by an additional 2:1 in both directions, to give I_{16} .

In any event, I_{16} is then coded by the original ISO-ADCT (n502) coding scheme with high quality at say 1.3 bits per pel. Since there are 16 times fewer pels, the net bit-rate is 1.3/16 = 0.081 bit per original pel. The decoded image is \widehat{I}_{16} .

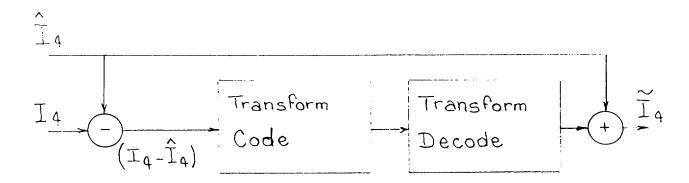
This decoded/subsampled image can be upsampled 1:2 by interpolation in both dimensions to give \hat{I}_4 , which acts as a prediction for coding I_4 . \hat{I}_4 can be further upsampled 1:2 in both dimensions to give a full sized image for display. Simple bilinear interpolation is sufficient for most imaging applications, especially taking into account that interpolated values will be automatically corrected at latter stages in the coding.



Following this, the residual prediction – error picture $I_4 - \widehat{I}_4$ can be transform coded. Note that \widehat{I}_4 is available at the receiver too. So the decoded version \widehat{I}_4 is obtained from :

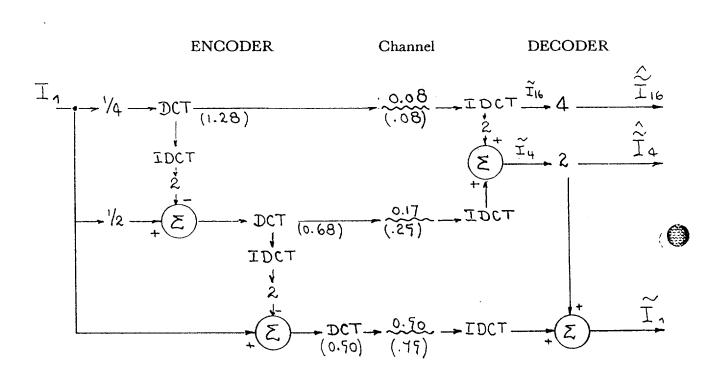
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(1.77a (1755)



As before, I_4 can be interpolated 1:2 in both dimensions to give I_1 for display and prediction. If $(I_4 - \hat{I}_4)$ is coded with high quality at about 0.67 bit per pel, then since there are 4 times fewer pels the overall bit – rate is 0.67/4 = 0.167 bit per pel. So the total to achieve this second stage display is 0.081 + 0.167 = 0.25 bit per pel.

Similarly, coding the residual error image $I_1 - \hat{I}_1$ at high quality at about 0.5 bit per pel enables a third stage display \hat{I}_1 to be obtained within 0.75 bit per total bit – rate.



note: "DCT" and "IDCT" = ISO n502

Figure 1 - Summary of the Principles of the Hierarchical Structure

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1.3.1.3 Coder Block Diagram

The Progressive Transform Coder (fig. 1) takes in video and stores the frame to be coded in Frame Store 1. Next the image is filtered and subsampled by modules 2 and 3 to produce the first signal to be sent (I_{16} in the above example). At this time Frame Store 7 contains no information, and its outputs is zero. Thus, I_{16} passes through Substractor 4 unchanged to DCT Coder 5, which transforms and quantizes the coefficients that are to be transmitted.

The quantized coefficients then pass to variable world – length Coder 6, which produces bits for the channel, as well as to inverse DCT Decoder 9, which produces decoded pel values. Delay 8 is a short delay needed to match that of the forward and inverse DCT modules 5 and 9. Since Frame Store 7 is output – ting zeros, so is Delay 8. Thus, the decoded pels pass through Adder 11 un – changed to the bilinear 1:2 Interpolator 10, which upsamples the decoded image, quadrupling the number of pels. This interpolated image (Î4 in the above example) then passes to Frame Store 7 to be used as a prediction in the next stage of progressive transmission.

In the next stage, the original picture is filtered and subsampled by modules 2 and 3 to produce I_4 at the "+ input" of Substractor 4. However, during this stage, Frame Store 7 outputs \hat{I}_4 , and thus substractor 4 outputs the differential signal $I_4 - \hat{I}_4$, which is then coded, sent and decoded respectively, by modules 5, 6 and 9.

Now, the output of the Delay 8 is \hat{l}_4 , which is added to the decoded differential signal $\hat{l}_4 - \hat{l}_4$ at the output of DCT Decoder 9 to produce the decoded signal \hat{l}_4 at the output of Adder 11. This then passes to Interpolator 10 where it is upsampled 1:2 to produce \hat{l}_1 , which is then stored in Frame Store 7 to be used in the next coding stage.

In the final stage of coding, the original signal I_1 passes unchanged through modules 2 and 3. The prediction \widehat{I}_1 from Frame Store 7 is substracted, and the differential signal $I_1 - \widehat{I}_1$ is DCT coded and sent as above, thus completing the transmission of the picture.

1.3.1.4 Decoder Block Diagram

The Progressive Transform Decoder (fig. 2) is essentially the inverse of the coder. Initially, Frame Store 17 contains nothing and outputs zeros.

At the first stage, the received quantized transform coefficients (I₁₆ in the above example) are produced from the the received channel bits by variable world-length Decoder 13. These then pass to inverse DCT Decoder 16, which produces decoded pels values. Since Frame Store 17 is outputting zeros, so is Interpolator 14. Thus, the decoded pels I₁₆ pass through Adder 15 unchanged to the Frame Store 17.

Frame Store 17 has two outputs, which operates at different rates. Output B operates at the display rate and passes to programmable interpolator 12, which produces the display picture. During the first stage, Interpolator 12 upsamples I_{16} by 1:4 in both dimensions. During the next stage of progressive transmission, output A of Frame Store 17 passes to the bilinear 1:2 Interpolator 14, which upsamples the previously decoded image (I_{16} here), quadrupling the number of pels. This interpolated image (\widehat{I}_{4} here) then passes to Adder 15. During this stage the Decoders 13 and 16 produce the decoded differential signal \widehat{I}_{4} - \widehat{I}_{4} , which is added to \widehat{I}_{4} to produce the decoded signal \widehat{I}_{4} at the output of Adder 15. \widehat{I}_{4} then passes into Frame Store 17.

Note that the programmable interpolator 12, allows for other upsampling ratio than 1:2 or 1:4, as described herein, and thus, the decoder could adapt the input – resolution to the display resolution.

In a final stage of coding, Decoders 13 and 16 produce the decoded differential image $\tilde{I}_4 - \tilde{I}_4$, Interpolator 14 produces \tilde{I}_4 , and Adder 15 outputs a final decoded image \tilde{I}_4 , thus completing the transmission of the picture.

Note that the final coding stage could be repeated if we desired to improve the quality of the transmitted picture. To do so, we simply bypass the bilinear interpolators 10 and 14, and perform another coding of the residual error image $I_{\Lambda} - I_{\Lambda}$. This correction signal is then added to the previously received picture, thus improving its quality.

Furthermore REVERSIBILITY, will simply be obtained by simply bypass – ing the DCT Coder and Decoders 5, 9 and 16, then original pel values can be coded, and LOSSLESS entropy coding can be carried out by Encoder 6.

General

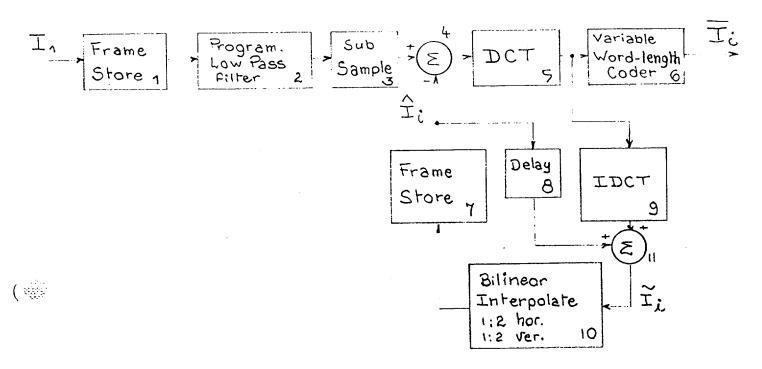


Figure 2 - HIERARCHICAL TRANSFORM CODER

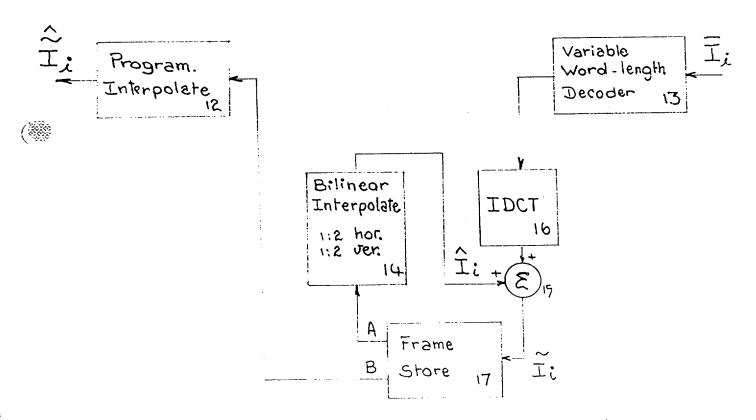


Figure 3 - HIERARCHICAL TRANSFORM DECODER

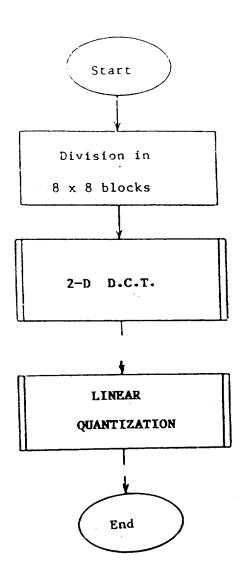
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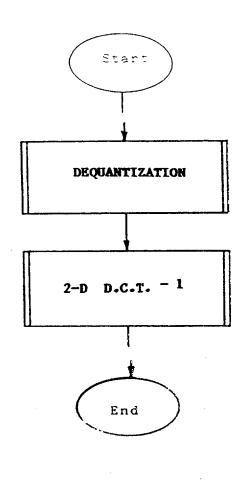
Figure 4

DCT CODER

Figure 5

Inverse – DCT DECODER





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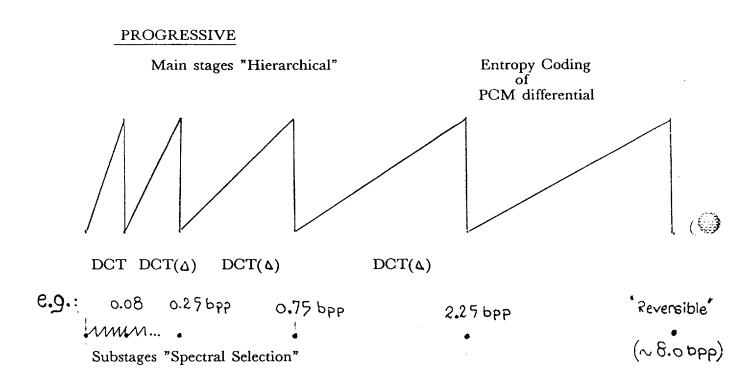
1.3.2 Coding Structure for sequential update

As aforementionned, Sequential update can be carried out in two ways. One is built on the "Progressive Hierarchical Structure", the other is built in a "one – pass" coding structure. The latter, is by far the most "natural" for a block coding technique, and requires no transmission buffer, unlike the former that imposes to resequence the data before transmission.

The drawback of the simplest one, beeing that it does not fully match the picture coded in progressive update at the same coding rate. But, does that mind the service????

General

Progressive and Sequential Schemes



SEQUENTIAL Main stage Entropy Coding PCM differential DCT DCT(4) e.g.: 0.75 bpp 2.25 bpp Reversible (~8.0 bpp)

NOTE that the first Main stage of progressive update may be equivalent to main stage of the sequential update

(*) Note that sequential might be changed before the january meeting, as agreed in n599.

General

1.3.3 Technical Description

1.3.3.1 Filtering - Subsampling - Interpolating

See Tables 4, 5, 6 and 7 in Annex A.

1.3.3.2 Transformation

A 2-dimensional Discrete Cosine Transform is applied to 8 x 8 blocks of pixels.

1.3.3.3 Linear Quantization

After transformation, the coefficients are quantized linearly, in using a quantization step dependent on the spectral position of the coefficient to be quantized. The quantization process is simply a round – off operation on the co – efficient value divided by the quantization step.

NOMINAL quantization steps are given by the quantization matrices (ta – ble...) which have been derived from a psychovisual experiment. In other terms, the quantization steps have been set to be the "Just Noticeable Ampli – tude Perception" at normal viewing distance, of each coefficient considered in – dependent. Those values used directly will lead to a very good picture quality at a coding rate < 1.0 bit per pel, but by scaling them with a common factor, in the general case, we can adjust the bit rate and correspondingly the picture quality, to nearly any desired value.

The values shown are those presently used for pictures described according to CCIR A-601, mode "4-2-2" – as required for ISO testing purposes –, but as they are part of the overhead (= 128 bytes) they can be changed for other picture resolutions or specific applications (Graphic arts, Synthetic images,...). However, an example of quantization matrices for resolution conforming to CCIR A-601, mode "2-1-1" is given. Obviously, this aspect needs future refinements, but the coding structure is fully open.

1.3.3.4 Adaptive Entropy Coding

After quantization the transform – quantized block is sparse : only few AC coefficients are non – zero.

The content of the block is transmitted by sending information on the value and the position of the coefficients, and for this purpose the transform – quantized block is described by a 1 – dimensional scan, that is the simple "zig – zag" scan, from upper left hand corner (table...)

The DC coefficients are treated seperatly, they are represented in DPCM and encoded in using particular 1 - dimensional VLC tables.

AC coefficient relative positions and number of bits to represent subsequent AC value are encoded using 2 – dimensional VLC tables. The AC value itself is coded in PCM with the corresponding number of bits expressed by the previous code.

The VLC tables may be predefined or customized to specific image(s) or application(s). In the latter, they are part of the overhead and they account for about 400 bytes in total. In any case, the recommendation would defined a standard set of predefined VLC tables that could be used as default values, as long as no particular ones are transmitted.

The present description will still be based on the full adaptive case — i.e. customized VLCs —, owing to the tight ISO schedule it was impossible to derive the faithfull predefined set, BUT proofs have been made that predefined VLCs are definitly usable with nearly NO degradation of the coding rate.

General

2.0 Coding Algorithm

2.1 FUNCTIONALITIES

2.1.1 Exchange of Parameters

Header of general parameters: Resolution, nb of stage, sequential or progressive,....This will be defined latter within JPEG.

Header for dynamic redefinition of variables: VLCs, thresholding matrices,... Some are coded in the simplified syntax (par. 2.2.2.), others will be defined latter within JPEG.

2.1.2 Number of Stages

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The available number of stages is in fact very large in regard to the actual need.

However, those stages can be splitted in main stages and in sub-stages, as follows:

- The main stages are the hierarchical ones 9 in total, (-1,2 and 4 sub-sampling in both dimensions), the extra ones based on iterations of the DCT on the decoded errors of the full resolution image this number is not fixed, it may be 5 or 10 in total -, and lastly the reversible scan, in 1 or 2 main scans.
- The substages are based on spectral selection of the coefficients according to a predefined scheme (e.g. one coefficient per block at a time), for each main stage, excepted the last PCM entropy coded reversible scan that may be divided in a large number of subsampled scans.

In total, the coding structure allows for up to one hundred scans !!!!

For edition purpose, in progressive mode, the user will simply define the desired number of stages, and then dedicated routines – not yet defined, be – cause largely dependent on the service – will automatically route the image data in the processing structure that produces the most pleasing display.

This point - typically service dependent - is clearly for further study, but in any case, the coding structure does not impose any constraint to the user or the service.

2.1.3 Resolution Control

The only parameter that is sensitive to picture resolution, is the quantization matrix (Y,Cr,Cb). However, note that we might, very likely, specify an "aver – age" matrix that could give very good results in a variety of situations, but at the expense of loss of excellence.

Nevertheless, herein (Annex...) are described the to-day best quantizing matrices for Y,Cr,Cb conforming CCIR – A601, mode "4-2-2", as required for psychovisual selection purposes by ISO. Furthermore, Quantizing matrices for Y,Cr,Cb conforming CCIR – A601, mode "2-1-1", are proposed but have not been tested yet.

Finally, it is considered that this aspect is fully THEORITICALLY solved, but practically, partially answered. Indeed, it is always possible to define a ded – icated quantization matrix to fit a specific resolution (or application) and at a cost of, only, 128 bytes (!) for transmission to the receiver. For typical resolutions or applications, average or default quantization matrices, would be stored.

For the user, in any case, it will simply consist in specifying the resolution of the image to be coded, that is a mandatory parameter of the syntax for any image communication services.

2.1.4 Level Range Control

This coding algorithm has been typically designed for 8 bit per component image data. However, its applicability to: 1 bit.....7 bit per component image data, is straightforward by simple normalization of the input value and, further – more, by eventually the use of the adaptive entropy coding option.

Above 8 bit per component image data, the present scheme has to be revised, not on the principles, but on the implementation side (e.g. DSP chip with 32 bit internal resolution instead of the consumer DCT chip) and the adaptive entropy coding has to be adapted (i.e. the number of events is increased).

2.1.5 Quality Control

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This aspect is exclusively controlled by the quantization matrices. The documentation proposes NOMINAL quantization matrices that give TYPICALLY and CONSTANTLY, for any kind of images, an excellent psychovisual picture quality.

To modify this nominal picture quality, the user has simply to scale uniformly the quantization matrices with a common factor, that has to be part of the overhead syntax. To that end, the user would be provided with a picture quality scale that automatically would convert it, to a scaling factor.

However, future refinements of this simple control are fully open, via the definition of dedicated quantization matrices, that may be part of the overhead syntax.

2.1.6 Bit Rate Control

The bit rate is also, exclusively, controlled via the quantization matrices. However, unlike quality control, the bit rate control depends on the picture complexity and, since the scheme is adaptive, two pictures of equivalent psychovisual quality, will not, in general, be coded with the same total bit. So, in order to ensure maximum subjective quality of a given main stage, the mini mum threshold factor F giving a bit rate below or equal to the desired bit rate for that stage is calculated.

The algorithm is simple:

o the relation between the bit rate B and the threshold factor F has been empirically determined and is rather well represented by the expression:

$$\log(B) = a^*\log(F) + b \tag{1}$$

 A subroutine giving B as a function of F has been implemented. It is based on a Newton - Raphson iteration that gives, in a few steps, the desired value of F.

So, if a very precise bit rate is required, two solutions are offered: i-) iter-ations of the coding process to converge to it or ii-) coding of the image in progressive update to just above the desired bit rate and then, transmission of the exact amount of the compressed image data.

2.1.7 Synchronous Operability

The coding/decoding structure is fully in synchronous operation for :

- o the sequential update
- o the progressive update with a sufficient number of main stages that are basically each in sequential format

The coding/decoding structure is nearly in synchronous operation for :

o the progressive update as soon as sub - stages of spectral selection are used

in this latter case, the synchronous operability is perceptively met (i.e.the real time constraint is fully met), but not conceptually.

Another aspect to mention, is the use of the adaptive entropy coding option, where also, the synchronous operability is conceptually not met, since preprocessing of the image is needed, before transmission.

Nevertheless, for applications on ISDN, the mentionned pre – processing times are always insignificant in regard to the transmission time – in other terms, the processors are fast enough to conceal this aspect. For broad – band channels, progressive update may be no longer needed and so, full synchronous operation is available.

2.1.8 Reversibility

The DCT is known to be difficultly reversible. So, for reversibility, the coding structure, basically, escapes from DCT to PCM, and simply makes the differ – ences between the DCT decoded image – at one appropriate coding rate – and the original, and then code them via a lossless coding scheme – typically, an Huffman coding technique.

A alternative is described in Annex B.

2.1.9 Other Features

The prominent features of this algorithm are:

- Simplicity / efficiency
- ° Free licensing
- Foreseen, intra mode coding Compatibility with visiophony at 64 kbit/s (CCITT sg XV), visioconference at 384 kbit/s (CCITT sg XV) and Dig ital Television (CCIR).

2.2 PROGRESSIVE CODING

2.2.1 Flow Charts and Explanations

CODING AND DECODING FLOW CHART OF DCT-CODING

Notation

input value for code-generation

code :input value for code-generation

code :input value for code-generation

code :integer

codesize :array(low..high) of integer; indicates the code-length of a certain coefficient

coefsize :function that tells how many bits are required to

:function that tells how many bits are required to represent coefficient. Because the number of bits in a coefficient is known on beforehand, the following bit allocation may be used:

input values	nr bits	output values
-1,1	1	0,1
-3,-2,2,3	2	00,01,10,11
-74.4-7	3	etc
-58,8-15	4	etc
-3116,16-31	5	etc
-6332,32-63	6	etc
-12764,6412	7 7	etc

diag :array(0..63) of transformcoefficients

diffdc :one dimensional array consisting of the differential DC-coefficients

distemp :integer

EOB : code meaning End Of Block

freq :frequency-statistic equal to either freque or freqdc

frequency—
statistics. Each entry determines the distance to
the next non-zero coefficient along a zig-zag
scanning path along with coefsize of the next
coefficient

freqdc :one dimensional array used for the DC-frequency-

statistics

high

:integer=constant; indicates the highest possible

coefficient

huf

:array(low..high) of integer

hufftab

:array(0..numberofcodes-1) of

record

value :integer; coefficient value huffcode:integer; coefficient huffcode

size :integer; huffcode size

end

1

:integer

j

:integer :integer

low

:integer=constant; indicates the lowest possible

coefficient

maxcodelength:integer=constant describing the length of the

longest possible huffman code. If all posibilities are >10 , which is the case for DC-coefficients as well for AC-coefficients and runlengths then

maxcodelength<21

nil

:integer

number of codes: integer; describing the number of different

values to be encoded

others

:array(low..high) of integer; tells which other

codesizes should be incremented

point

:integer

quanstep

:array(0..7.0..7) of integer described in table 1

and table 2

run

:integer

si

:integer

stage

:integer; indicates the stage in the progressive

update

temp

:integer

tresholdfactor:integer

хfгш

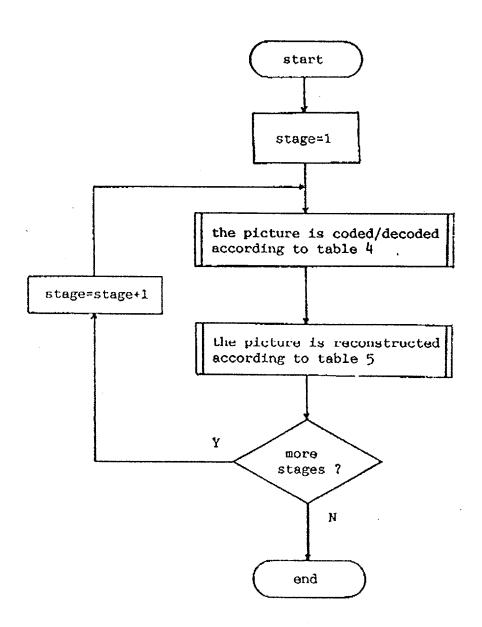
:array(0..7.0..7) of byte consisting of

coefficients

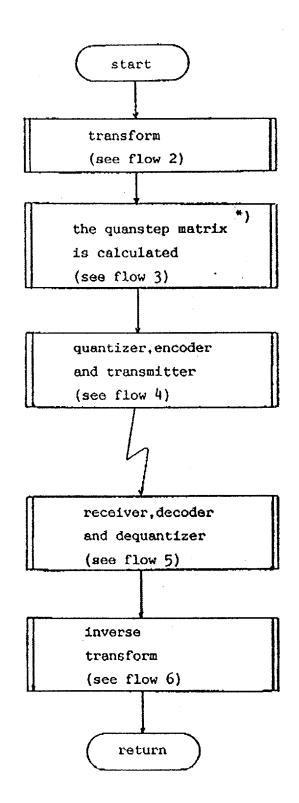
zigzag

:array described in table 3

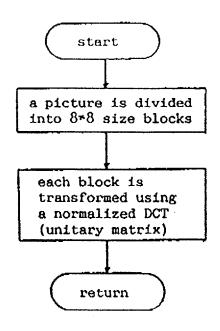
Progressive update coding/decoding.



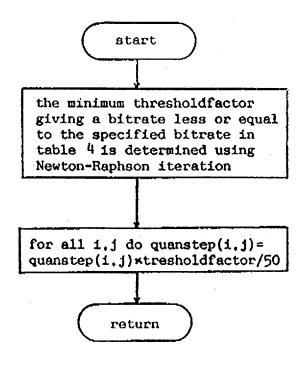
Flow 1 Index of the description of the algorithm



this part has only been implemented for ISO selection purposes, and will be removed for real applications



Flow 3 The quanstep matrix is calculated

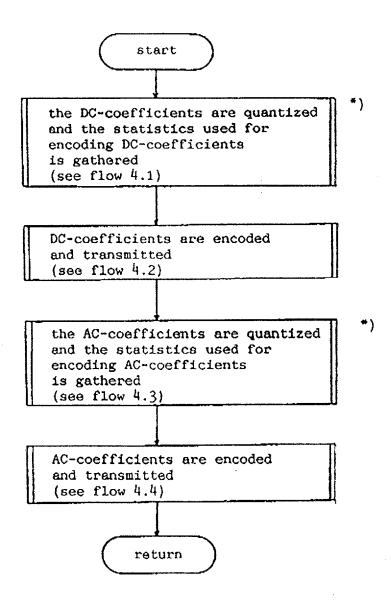


4)

this part has only been implemented for ISO selection purposes, and will be removed for real applications

Flow 4

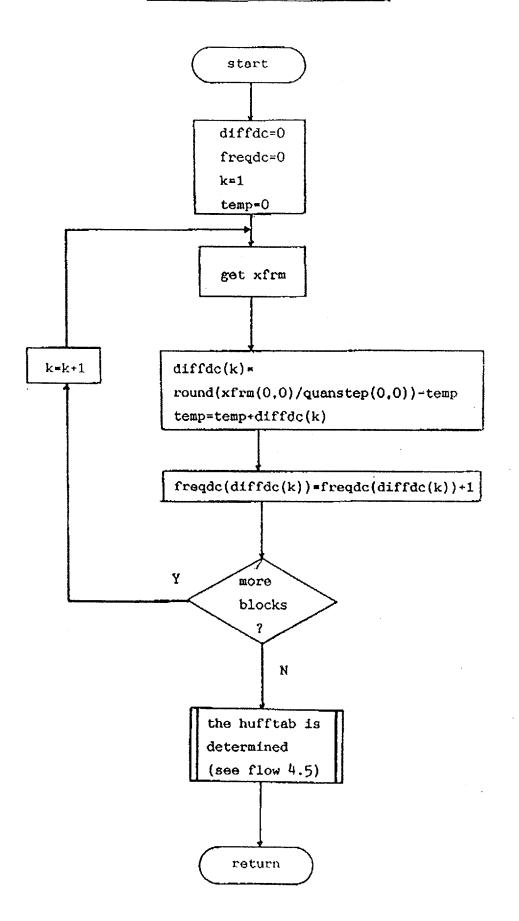
Quantizer, Encoder and Transmitter



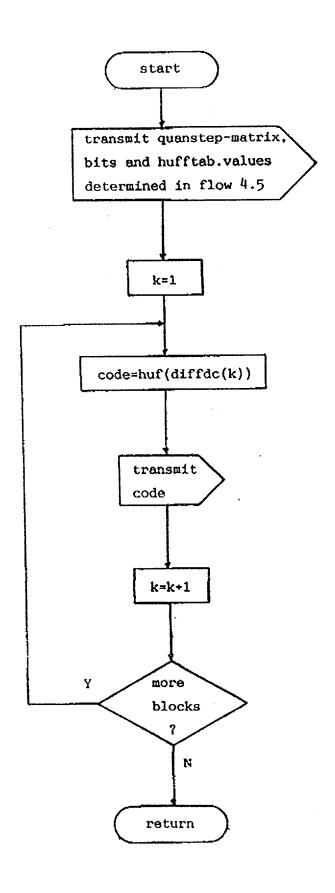
*) this part could be replaced by predefined VLC-tables in real applications

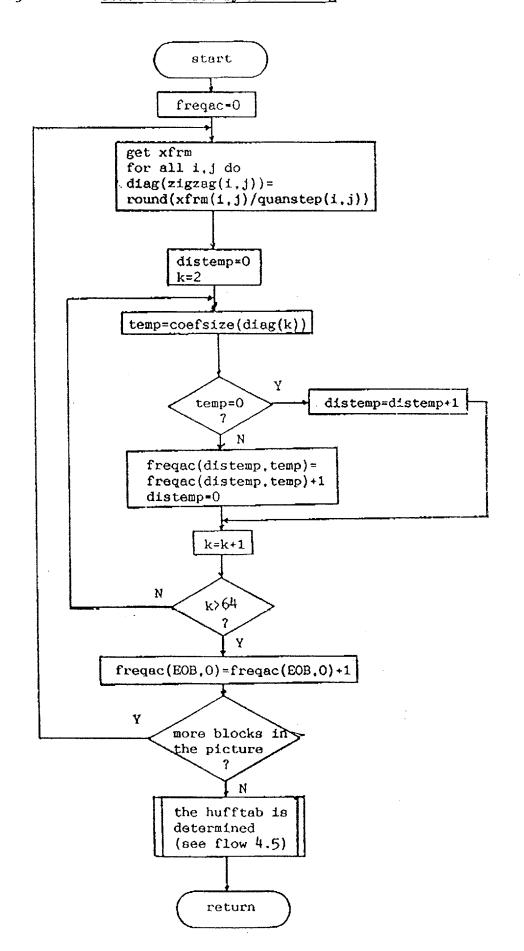
2

Flow 4.1 Statistic used for DC-encoding



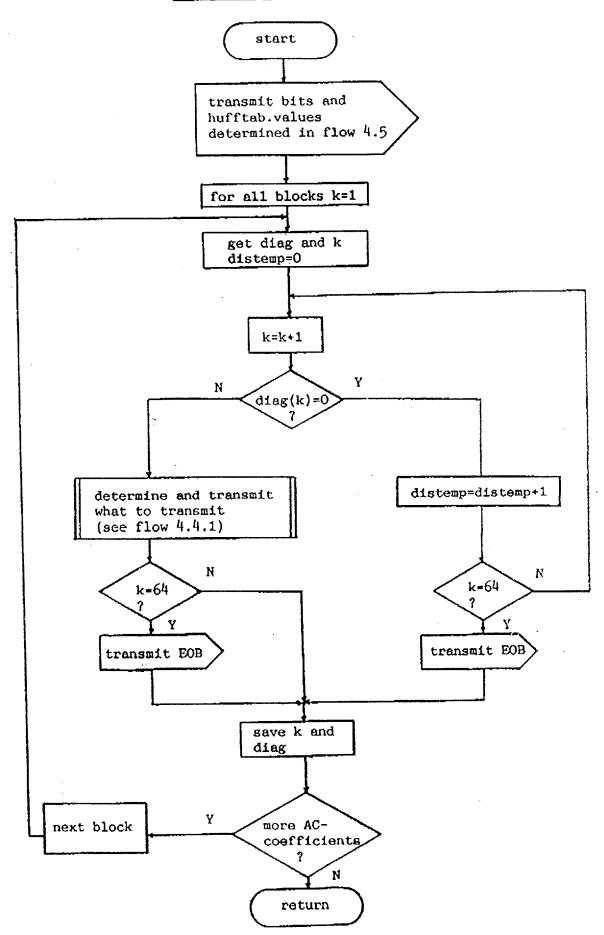
Flow 4.2 DC-coefficients encoding and transmitting



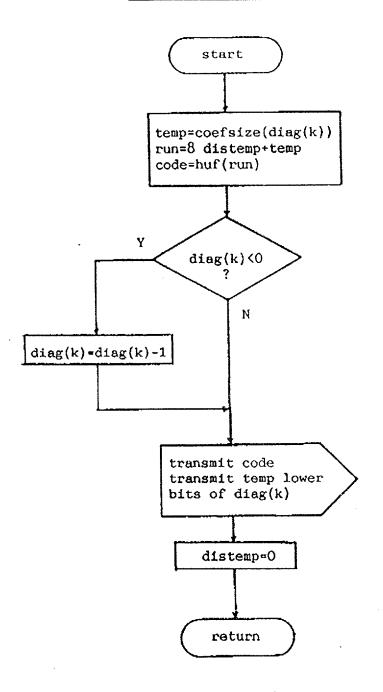




AC-encoding and transmitting

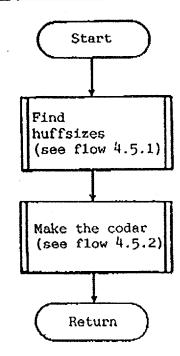


Flow 4.4.1 Determine and transmit what to transmit



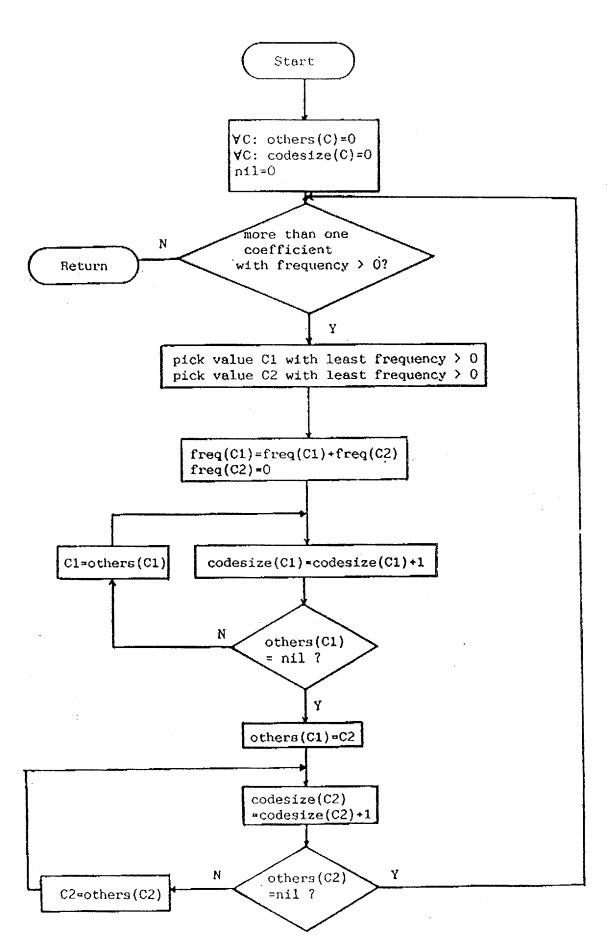
Flow 4.5 The hufftab is determined

((())



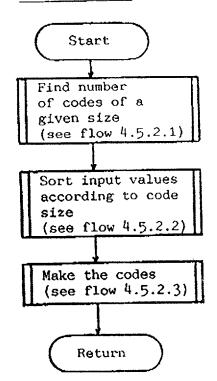
:63

Find huffsizes

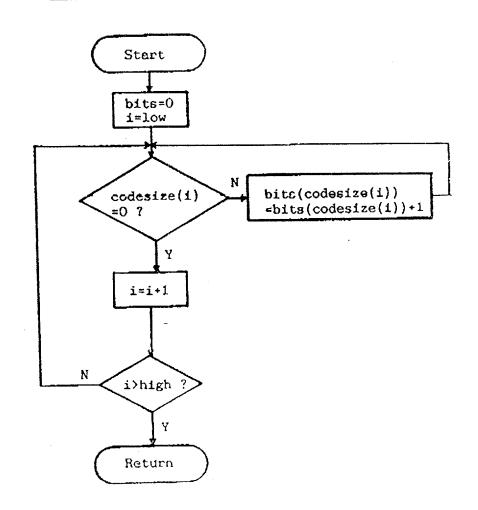


Flow 4.5.2

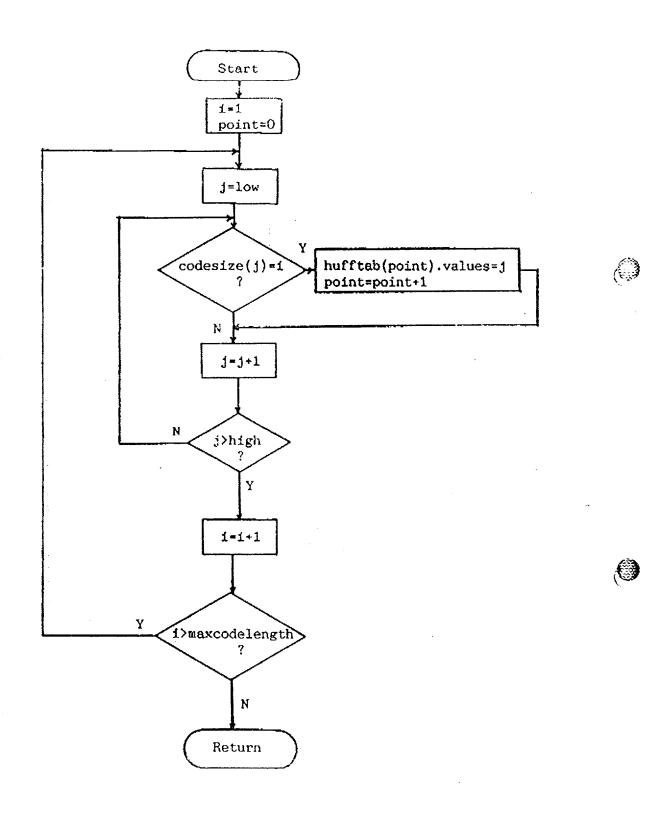
Make the codar

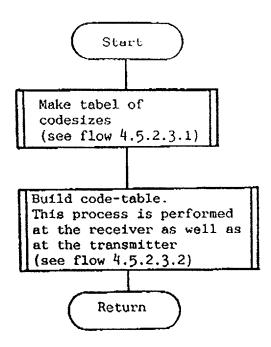


Flow 4.5.2.1 Find number of codes of a given size



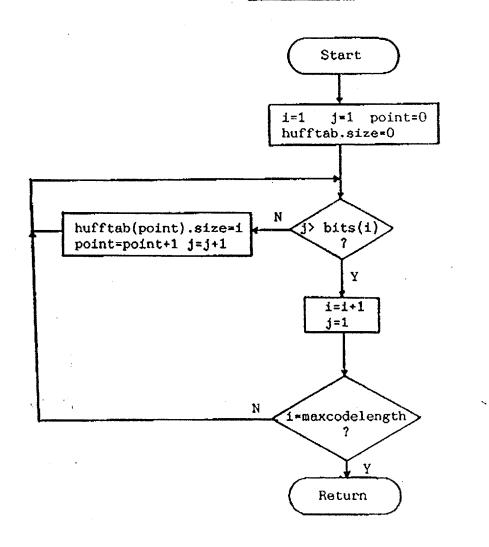
Flow 4.5.2.2 Sort input values according to code size



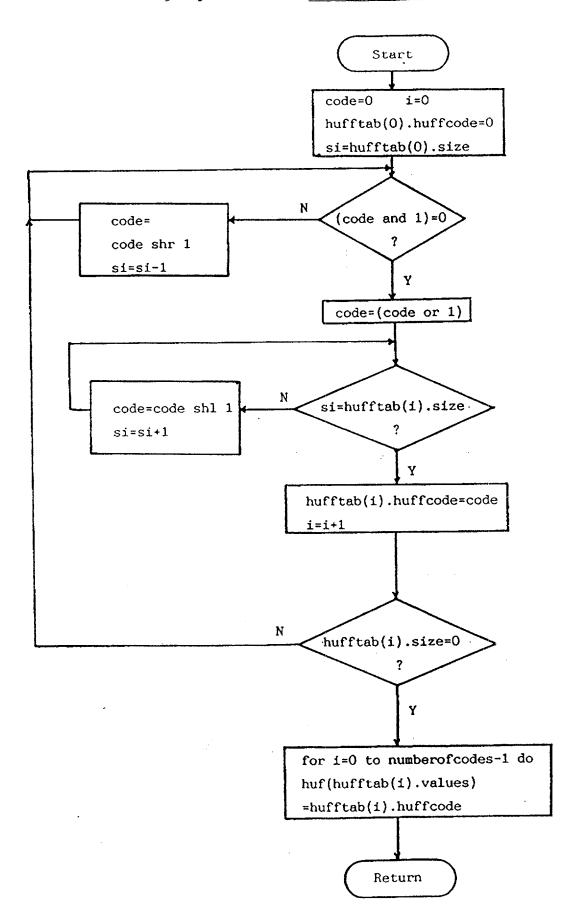


Flow 4.5.2.3.1

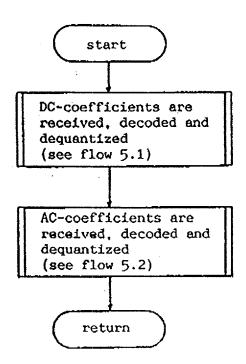
Make table of codesize



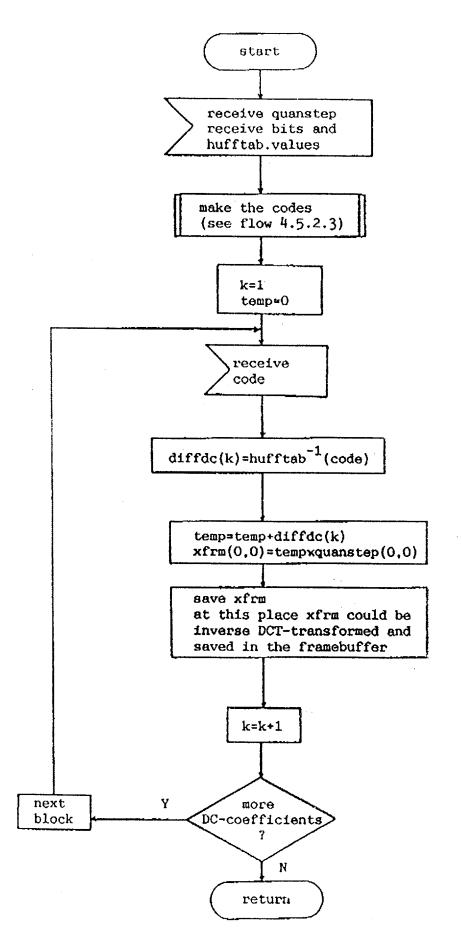
}{{};



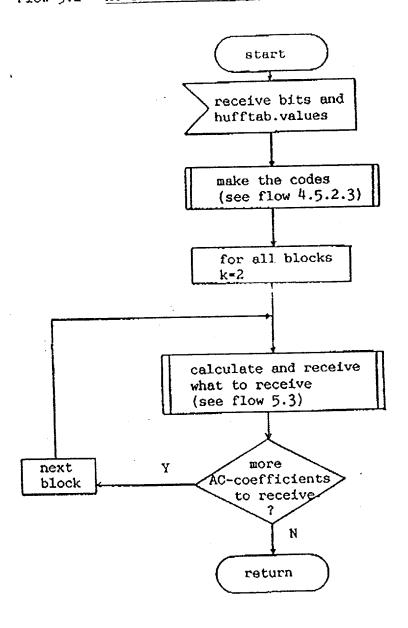
Flow 5 Receiver, Decoder and Dequantizer



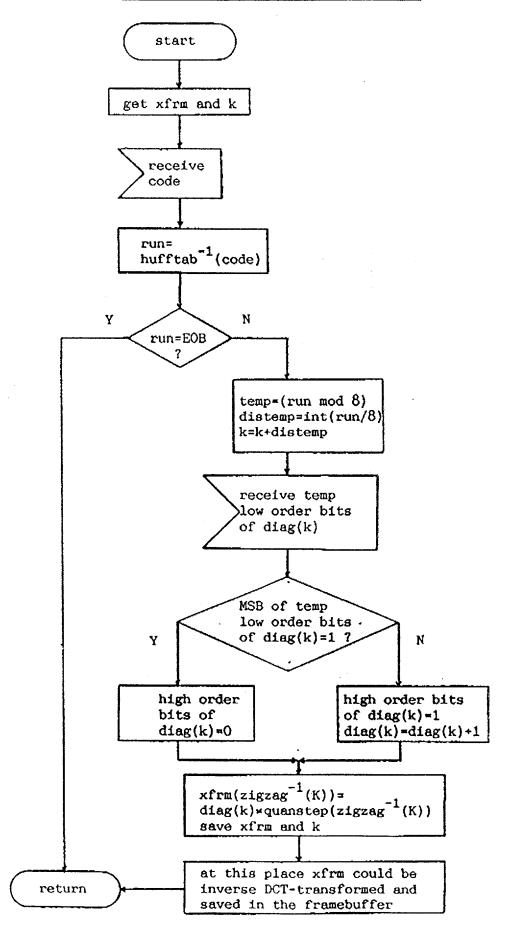
Flow 5.1 DC-coefficients are received, decoded and dequantized

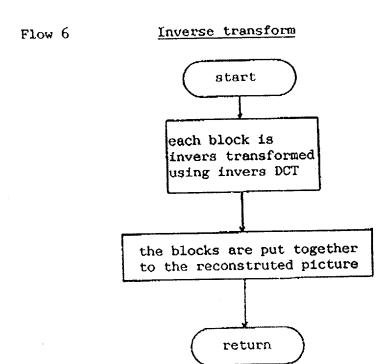


Flow 5.2 AC-coefficients are received, decoded and dequantized



Flow 5.3 Calculate and receive what to receive





Syntax for a main stage (non reversible)

File Threshold Codetable Lum.DC Codetable Chr.DC size matrices for lum.DC for chr.DC	1				<u> </u>	I	
size matrices for lum.DC for chr.DC		File	Threshold	Codetable	Lum.DC		 _
		size	matrices	for lum.DC		for chr.DC	

Codetable Codetable Lum.AC-chr.AC, lum.AC-chr.AC, for lum.AC for chr.AC

Syntax for the reversible stage

File	Codetable	Values	
size			

2.3 SEQUENTIAL CODING

2.3.1 Flow Charts and Explanations

To be delivered later as agreed in n599 rev.1

2.3.2 Syntax

To be delivered later as agreed in n599 rev.1

2.4 FUNCTIONAL RELATIONS BETWEEN PROGRESSIVE AND SEQUENTIAL

To be delivered later as agreed in n599 rev.1

Annex A. Tables

```
(( 16, 11, 10, 16, 24, 40, 51, 61), ( 12, 12, 14, 19, 26, 58, 60, 55), ( 14, 13, 16, 24, 40, 57, 69, 56), ( 14, 17, 22, 29, 51, 87, 80, 62), ( 18, 22, 37, 56, 68, 109, 103, 77), ( 24, 35, 55, 64, 81, 104, 113, 92), ( 49, 64, 78, 87, 103, 121, 120, 101), ( 72, 92, 95, 98, 112, 100, 103, 99))
```

Table 1 Visibility tresholds for the luminance.

Table 2 Visibility tresholds for crominance.

```
1 2 6 7 15 16 28 29

3 5 8 14 17 27 30 43

4 9 13 18 26 31 42 44

10 12 19 25 32 41 45 54

11 20 24 33 40 46 53 55

21 23 34 39 47 52 56 61

22 35 38 48 51 57 60 62

36 37 49 50 58 59 63 64
```

Table 3 Showing how the block may be zig zag scanned

Annex A. Tables

I	
Stage number :	Description of how the picture to be encoded is calculated:
1	The original picture is filtered using the filter described in table 6. The filtered picture is subsampled 2:1 in both x- and y-direction and saved in TEMP picture. The subsampled picture is filtered and subsampled using the same procedure. The resulting picture is coded to 0.08 bit/pix using the N502 in flow 1.
2	The picture produced in stage 1. in table 5. is subtracted from the TEMP picture. The resulting picture is coded to 0.17 bit/pix using the N502 in flow 1.
3	The picture produced in stage 2. in table 5. is subtracted from the original. The resulting picture is coded to 0.5 bit/pix using the N502 in flow 1.
4	The picture produced in stage 3. in table 5. is subtracted from the original. The resulting picture is coded to 1.5 bit/pix using the N502 in flow 1.
5	The picture produced in stage 4. in table 5. is subtracted from the original. The resulting picture is huffman-coded and transmitted.

Table 4 Describing how the picture to be encoded is calculated in the different stages.

Annex A. Tables 44

Stage number:	Description of how the reconstructed picture is updated:
1	The picture is expanded 1:2 in both x- and y-direction using the interpolation described in table 7. (Only at the decoder, the picture is expanded again to full size.)
2	The picture is added to the picture produced in stage 1. The resulting picture is expanded 1:2 in both x- and y-direction using an interpolation described in table 7.
3	The picture is added to the picture produced in stage 2.
4	The picture is added to the picture produced in stage 3.
5	The picture is received and huffman-decoded, and added to the picture produced in stage 4.

Table 5 Describing how the reconstructed picture is updated.

	Ţ	
1	4	1
4	16	4
1	4	1
1	4	1

Table 6 A low pass filter.

a	x 2	b
x 5	х1	x 3
đ	x4	c

Notation:

a.b,c,d

:pixel values picked in the subsampled

picture

x1, x2, x3, x4, x5

:pixels values calculated as follows:

x1=(a+b+c+d)/4

x2=(a+b)/2

x3=(b+c)/2

x4=(c+d)/2

x5=(d+a)/2

Table 7 Describing how the picture is expanded.

Annex B. Alternative for reversibility

DESCRIPTION OF THE METHOD USED TO GET THE REVERSIBILITY

We tried several possibilities in the pel domain (VLC of the differences, prediction of the differences, VLC of the configurations of the differences in a 2 x 2 sub-blocks) without reaching good results.

The method giving the best performances is based on two steps:
- transmission with VLCs of the remainders of the transform coefficients

- transmission of the small residual difference between original and decoded image using run-length coding of the zeroes' values and VLC of the differences.

The assumptions made for the first step are the following:

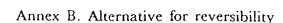
- 1. the transform coefficients are represented with 12 bits
- 2. at the previous step (2.25 bit/pel) a maximum step size of 8 is used (according the quantisation rule of fig. 1), resulting in a residual difference in the range from -3 to +4
- 3. the VLC used to code the Y coefficient differences is shown in tab. I
- 4. the VLC used to code the coefficient differences of the chrominance components (CR and CB) is shown in tab. II

The assumption made for the second step are:

38

- 1. the VLC used to code the zeroes sequences is written in tab. III.
- 2. the last word (100110) is used both to say that no errors were encountred in the 90 pels examined and to form an address greater than 720 which serves to pass to the following line
- 3. the VLC used to code the pel-value differences is shown in tab. IV.

The statistics used to compute the different VLCs can of course be improved but, we aspect, with marginal effect on the whole performances.



```
000
     0.1731457412
                               2
                                    01
-2
     0.2661795914
                                                E= 2.80 bit
                                    11
-1
     0.1864149272
                                    100
0
     0.1227117106
                                    0010
     0.0886115953
1
                                    0011
2
     0.0666907802
                                    1010
3
     0.0525824651
                                    1011
     0.0436631963
                           Table 1
     0.2682926953
                               2
                                    01
-3
-2
     0.3902438879
                                    1
                               3
                                    000
     0.1731707305
 0
     0.0731707290
                                    0011
                                                E = 2.25 bit
                               5
 1
     0.0414634161
 2
                                    001001
     0.0243902430
 3
                                    0010000
     0.0170731712
                                7
      0.0121951215
                                    0010001
                             Table 2
 -2
      D.0000149$60
                                 3
                                     111
                                 2
 -1
      D.3846521974
                                     10
      D.6152132154
                                     0
      D.0001196476
                                     110
```

Table 3

(93.55

```
0
     0.0747560295
                                     110
 1
     0.0906299502
                                     0000
2
     0.0784915835
                                     กกรก
3
     0.0741735846
                                     0100
 4
     0.0661373138
                                     0110
 5
     0.0599241965
                                     0111
     0.0511442684
 6
                                     1010
 7
     0.0487453826
                                     1011
 8
     0.0427001864
                                5
                                     00010
 9
     0.0394377001
                                5
                                     00011
10
     0.0351436920
                                5
                                     01010
11
     0.0307777189
                                5
                                     10000
     0.0282828771
                                     10010
12
                                S
     0.0241807792
13
                                5
                                     11101
14
     0.0220937487
                                5
                                     11111
      0.0191431176
15
                                6
                                     001101
16
      Q. d182795189
                                6
                                     001111
17
      0.0170320980
                                6
                                     010110
18
      0.0140814660
                                6
                                     100011
19
      0.0133618005
                                6
                                     100111
20.
      0.0113707241
                                6
                                     111100
21
      0.0113227461
                                6
                                     111101
22
      0.0096915029
                                7
                                     0011001
23
                                7
      0.0088758813
                                     0011101
24
      0.0079403156
                                7
                                     0101111
                                7
25
      0.0063330615
                                     1110001
26
      0.0065489612
                                7
                                     1110000
                                7
27
      0.0055174399
                                     1110011
28
      0.0046778293
                                8
                                     01011100
29
      0.0048217629
                                8
                                     00111001
30
      0.0038382190
                                8
                                     10001001
31
      0.0039821523
                                 8
                                     10001000
32
      0.0034304082
                                 8
                                     10001011
      0.0030465864
33
                                 8
                                     11100101
34
                                 5
      0.0027107422
                                     001100001
                                 9
35
      0.0024708535
                                     001110000
36
      0.0024948423
                                 9
                                     001100011
      0.0019191095
                                 9
37
                                     010111011
38
      0.0017751763
                                 9
                                     100010101
39
      0.0013913544
                                10
                                     0011000000
      0.0012234324
40
                                10
                                     0011000101
41
      0.0010075325
                                10
                                      1000101000
42
      0.0011754546
                                10
                                     0011100011
43
      0.0013913544
                                10
                                      0011000001
44
      0.0010555102
                                10
                                      0101110101
45
      0.0006956772
                                      00110001000
                                11
      0.0005757328
46
                                11
                                      00111000101
 47
      0.0007676438
                                10
                                      1110010001
 48
      0.0004317996
                                      10001010011
                                11
 49
      0.0005757328
                                11
                                      01011101000
50
      0.0006237106
                                11
                                      00111000100
 51
       0.0003838219
                                11
                                      11100100001
 52
       0.0004557885
                                11
                                      10001010010
 53
       0.0002638776
                                12
                                      010111010010
       0.0003358442
 54
                                12
                                      001100010010
 55
       0.0001919109
                                12
                                      111001000001
       0.0002398887
 56
                                12
                                      010111010011
 57
       0.0001919109
                                12
                                      111001001100
       0.0001439332
 58
                                13
                                      0011000100111
 59
       0.0001679221
                                13
                                      0011000100110
```

```
111001001101
     0.0001919109
                              12
60
                                    111001001010
                              12
     0.0001719109
61
                                    111001001011
                              12
     0.0001717109
62
                                    1110010000001
                              13
     0.0000759555
63
                              14
                                    11100100000001
     0.0000479777
64
                                    111001001000
                               12
     0.0001717109
65
                               14
                                    11100100111100
      0.0000479777
66
                               14
                                    11100100111101
      0.0000479777
67
                               15
                                    1110010011111100
      0.0000239889
68
                                    11100100100110
      0.0000479777
                               14
69
                                    11100100100111
                               14
      0.0000479777
70
                                     111001001111101
                               15
      0.0000239889
71
                                     11100100000000
                               14
      0.0000719666
72
                               14
                                     11100100100100
      0.0000479777
73
                                     11100100100101
                               14
74
      0.0000479777
                                     111001001110010
                               15
75
      0.0000239889
                                     11100100111010
      0.0000479777
                               14
76
                                     11100100111011
                               14
      0.0000479777
77
                                     1110010011111111010
      0.000d000001
                               18
78
                                     111001001110011
      0.0000239889
                               15
79
                                     111001001110000
                               15
80
      0.0000239889
                                     1110010011111111011
      0.0000000001
                               18
 81
                                     1110010011111111000
                               18
      0.0000000001
 82
                                     1110010011111111001
                               18
 83
      0.0000000001
                                     111001001110001
                               15
      0.0000239889
 84
                                     111001001111111111
                               18
      0.000000001
 85
                                     1110010011111111111
      a.aaadaaaaa1
                               18
 88
                                18
                                     1110010011111111100
      0.0000000001
 87
                                     1110010011111110
                                15
      0.0000239889
 88
                                     1110010011111111101
                                18
       0.00000000001
 89
                                     100110
 90
      0.0138895549
```

Table 4

Annex C. Results with predefined VLCs

ISO

INTERNATIONAL ORGANISATION FOR STANDARDIZATION ORGANISATION INTERNATIONALE DE NORMALISATION

ISO/TC97/SC2/WG8

CODED REPRESENTATION OF PICTURE AND AUDIO INFORMATION

ISO/TC97/SC1 7C8 N

/adctg N 39

Source: SAT/CCETT

Title: Symetry for the ADCT scheme - N 502

Introduction

As agreed at our Washington meeting, CCETT has continued the work for the full symetry of N 502. We have always STRESSED the need for this specific item, for at least the following reasons:

- 1. Simplicity of implementation e.g.on silicon
- 2. ONE PASS Coding technique
- 3. Greater Compatibility with Visiophony, Teleconference and Digital T.V. Coding techniques
- 4. Better fitness to Real Time constraints, and especially for Face to Face services

Results

The following results are based on either the use of Customized VLC as described in N 502, or the use of Predefined VLC tables derived from the four ESPRIT test images. In most cases the impairments on the coding rate are negligible for "natural images", and very limited on "binary images" – note that the learning phase does not include "binary images", yet.

Pictures	Coding rate n502	Coding rate with predefined VLCs
Barbara	1.03	1.02
Binary1	1.35	1.38
Binary2	0.59	0.63
Binary3	0.66	0.70
Binary4	0.97	1.01
Zelda	0.55	0.56
Clown	0.86	0.86
Flower	0.85	0.85
Boy	0.99	. 0.99
Iba	0.92	0.92
Toys	0.79	0.79
EbuChar	0.73	0.74
Boats	0.92	0.92
Tree	1.51	1.52
Violin	1.56	1.58
Young	0.90	0.91

Annex D. Overall Results

List of the illustrations

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