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TITLE: Description of two of the Video Codecs used in the United States

with Primary Rate Transmission Systems

Attached are descriptions of two of the video codecs used in the United States. One is provided by Compression Labs, Incorporated (ANNEX A) and one is provided by NEC America (ANNEX B). These are offered to the Specialists Group on Coding to further its work in addressing Part 3 of Recommendations H.120 and H.130.

ANNEX A

ANNEX A

1.0 INTRODUCTION

This contribution outlines a codec which is capable of transmitting 525 NTSC video, voice and data information over a 1.544 Mbps communications channel.

The codec is suitable for use in visual telephony and is offered for discussion relating to Part 3 of Recommendations H-120 an H-130.

2.0 COMMUNICATIONS

The communications portion of the system is described below. Following a general introduction, the communications frame structure is discussed in detail, covering the frame sync, format, system communications, audio, video field, high speed data, forward error correction, encryption and multipoint conferencing.

2.1 COMMUNICATIONS FRAME STRUCTURE

The communications frame provides a means of separating the various components that make up the teleconferencing signal. The proposed structure is based on the HDLC standard in which the various components are time division multiplexed and referenced to a synchronizing word. Figure 2.1 illustrates the format.

As a feature of the system, an identical structure is used via both a high speed serial data interface (RS449) and a TI/DSl interface. In the standard form of the Tl bit stream, the data is inserted within the Tl frame so that there is no specific allocation of particular data, such as audio, to particular channel slots. The codec frame is asynchronous with respect to the Tl framing structure. This is illustrated in Figure 2.2.

In variation thereof, the audio is treated differently in the multi-point conferencing mode. This is described in section 2.1.9.

2.1.1 Frame Sync

The Frame Sync consists of a 9-bit pattern (001111101) which has been selected for its superior autocorrelation characteristics. The framing circuitry also ignores one bit errors in the received frame pattern for added error immunity.

2.1.2 Format

The format field consists of 3 bits of status information which has been error protected by encoding it into a (7,3) BCH code. This coding allows a single error correction/double error detection scheme.

The status information specifies whether user data (RS-232 or RS-449) is present in the data fields.

2.1.3 System Communications

Inter-codec commands and system status signaling are handled via a 16 bit field which also serves as a low speed (1200 baud max) RS-232 user data communications channel. A single bit in the format field serves to separate the command mode data from the RS-232 user data.

The command mode serves to coordinate field sizes for the high speed data (video and user channels), graphics frame transfers and handle status information transfers for the user data ports (RS-232, RS-449).

2.1.4 Audio

The audio channel is encoded at a 64KHz rate using a 176 bit field. The coding algorithm is covered in section 3.3.

2.1.5 <u>Yideo Field</u>

The video field consists of the following major subfields: Sync, header, buffer status, and video block data.

2.1.5.1 Sync

The sync word is a unique 16-bit pattern that signifies the start of a physical video frame or strip. A strip occurs once for every 16 lines of video (see section 3.1). Since physical frames occur asynchronously with respect to transmission frames, a sync word may occur anywhere within the video field. Bit insertion/deletion protects against the accidental occurrence of a sync pattern within the video block data.

2.1.5.2 Header

The header is a five-bit subfield that follows the sync. It helps to differentiate between a frame sync and a strip sync. It also signals various operating modes of the codec, such as scan mode, pixel density, full motion video or graphics mode.

2.1.5.3 Buffer Status

The buffer status subfield consists of a four-bit code that indicates how full the bit rate buffer is (see section 3.1.6).

2.1.5.4 Video Block Data

The video block data consists of a concatenation of variable-length Huffman codes with block separators. The block separators consist of a motion status code at the beginning of each block and an end-of-block (EOB) code at the end. The Huffman codes, if present, indicate the amplitudes and position of any significant DPCM coefficients (see section 3.1.5).

2.1.6 High Speed Data

Two full duplex data ports on each codec communicate via this optional field. The presence and source of the data is indicated by bits in the format field. Field size is determined by the number of channels and their aggregate bit rate. The user data field, when allocated, is at the expense of the video field:

VIDEO + USER DATA = 3840 BITS

In order to facilitate buffer control, every eighth bit in the user data field is a buffer status bit used to indicate an end of message condition.

The user data rates run from 9.6 Kbps up to 448 Kbps in each channel, limited by total available bandwidth.

2.1.7 Forward Error Correction

As the control field contains its own error checking, only the command, audio user data and video data fields are error corrected. The algorithm used is a single burst error correction code (4080, 4098) capable of correcting any burst error up to 9 bits by using a 32 bit syndrome. This syndrome is generated via a Burton code polynomial ($_{\rm X}32 + _{\rm X}24 + _{\rm X}22 + _{\rm X}20 + _{\rm X}8 + _{\rm X}6 + _{\rm X}4 + _{\rm X}1$) and forms the last field in the frame.

2.1.8 Encryption

The system communications field and all data fields may be encrypted with a 56 bit user defined key.

The DES algorithm is employed in the Cipher Feed Back mode which links the 64 bit encryption blocks.

2.1.9 Multipoint Conferencing

In a multi-point conference, the audio is treated differently than described above. The audio field is removed from the communications frame as illustrated in Figure 2.3. In the case of communications over a Tl network, the data stream is inserted in channels one through 23 as illustrated in Figure 2.4. This allows the audio to be separately inserted in channel 24 outside of the codec encryption and forward error correction field. It may then be extracted at a nodal point and combined with audio from many sources before being distributed.

In a satellite network employing earth stations on customer premises, the audio may be carried on a separate SCPC or TDMA channel and combined at a nodal point as above.

The method for coding and controlling audio in a multipoint conferencing environment is left for further study.

3.0 CODING

This section discusses the methods employed to code and decode the video and audio data in terms of video processing and sampling, input memory and input processing, the cosine transform process, DPCM coding, Hutfman coding, rate buffer control and output processing.

3.1 VIDEO CODING

Figure 3.1 illustrates the major subsystems in the video coding section.

3.1.1 Video Processing and Sampling

In this portion of the system, the composite NTSC signal is separated into three components, one luminance (Y) and two chrominance components (R-Y and B-Y) and then converted into digital form. The sample rate used for luminance is twice the color subcarrier frequency (7.2 MHz) and chrominance is half the subcarrier frequency (1.8 MHz). This provides a luminance bandwidth of 2.75 MHz and a chrominance bandwidth of 680 KHz.

As it is only necessary to code and transmit the active video information, some of the sampled data is discarded. Of the 455 luminance samples per line, 386 are coded. Of the 114 chrominance samples per line, 96 are coded. A similar economy is performed in the vertical dimension. Of the 525 lines in an NTSC signal only 480 luminance lines and 512 chrominance lines are transmitted. Figures 3.2 and 3.3 illustrate this video format. The area of chrominance sampling is greater than that of luminance. After subsampling by 4:1 vertically (resulting in 128 lines), an integer number of a 16 x 16 pixel chrominance block is obtained.

3.1.2 Input Memory and Processing

This subsection performs a scan conversion on the video signal. The luminance frame of 480 lines by 368 pixels is re-scanned as 690 blocks of data, each containing 256 pixels arranged in a 16 X 16 pixel format. The position of each block with respect to a frame is illustrated in Figure 3.3.

The chrominance information is vertically subsampled four to one and scan converted into 48 blocks of R-Y data and 48 blocks of B-Y data, each 16 X 16 pixels in size. To further reduce bandwidth, the video frames are subsampled 2:1. This produces a data stream of 15 frames per second.

To enhance performance, the signal may be temporally filtered and noise reduced. This is optional and does not effect inter-operability.

3.1.3 Cosine Transform

A two-dimensional cosine transform is performed on the 16X16 pixel blocks. Both luminance and chrominance blocks are treated identically. This operation allows the removal of large amounts of redundant spatial information. The one-dimensional cosine transform for N points is given by

$$F(u) = 2C(u)/N \sum_{i=0}^{N-1} f(i) \cos[(2i+1)u\pi/2N]$$

for u = 0,1, ..., N-1, where

C(u) =1/
$$\sqrt{2}$$
 for u = 0
for u = 0.

The inverse transform is given by

$$f(i) = \sum_{u=0}^{N-1} C(u) F(u) \cos[(2i+1)u\pi/2n]$$

When the forward transform is performed both horizontally and vertically on a square block of pixels, a block of coefficients is formed in which the coefficient in the upper left corner represents the d-c average brightness within the block; energy, or coefficients progressing to the represent right increasing horizontal spatial frequencies; coefficients progressing downwards represent increasing vertical spatial frequencies. The cosine transform has been found to have superior energy compaction properties for typical images, as compared with all transforms possessing fast computational algorithms.

3.1.4 DPCM Coding

An adaptive DPCM loop is used to remove both spatially redundant and temporally redundant information from the stream of transform coefficients. The DPCM-coded coefficients are quantized to a degree determined by rate buffer fullness. Sixteen degrees of quantization severity are used, according to a four-bit buffer fullness code which is updated for every horizontal strip of blocks.

3.1.5 Huffman Coding

The quantized DPCM coefficients are scanned in a diagonal zig-zag within each block, starting from the d-c term in the upper left corner of the block and ending with the highest frequency term in the lower right corner. The amplitude of each nonzero DPCM coefficient is assigned a variable-length Huffman code derived from the statistics for typical teleconferencing images. Runs of zero amplitudes are coded from a separate table of Huffman codes. A run that extends to the end of the block is replaced with a single end-of-block code.

3.1.6 Rate Buffer Control

A rate buffer memory is used to regulate the flow of bits onto the channel. A four-bit word representing buffer fullness is updated for every horizontal strip of blocks and included in the video codestream. This word is used by the forward and inverse DPCM loops to control the degree of quantization used.

3.2 VIDEO DECODING

The video decoding process is complementary to the coding process, i.e., the functions performed by the Forward Transform logic and the Coding subsystems detailed above (sections 3.1.1 through 3.1.6) are basically reversed.

3.2.1 Output Processing

In the Output Memory subsystem, the scan conversion is reversed and the output frames are interpolated to restore a 30 frame per second video signal. The output video processing encodes the Y, R-Y and B-Y into an NTSC signal with conventional syncs and blanking.

3.3 AUDIO CODING

(for further study)

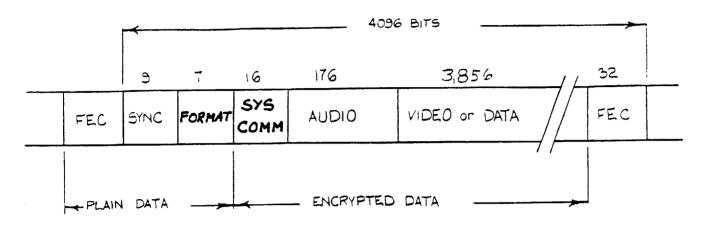


FIGURE 2.L COMMUNICATIONS FRAME FORMAT

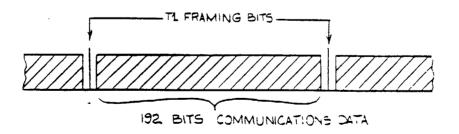


FIGURE 2.2 T1 FRAME FORMAT

				4096 BITS	——	
	9	7	16	4032	32	
FEC	SYNC	FORMAT	SYS COMM.	VIDEO/ DATA	FEC	

FIGURE 2.3

COMMUNICATIONS FRAME FORMAT - MULTIPOINT CONFERENCING

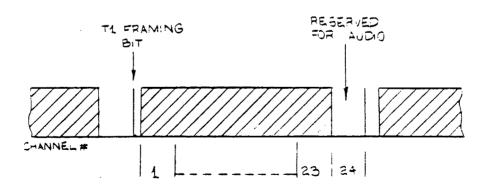
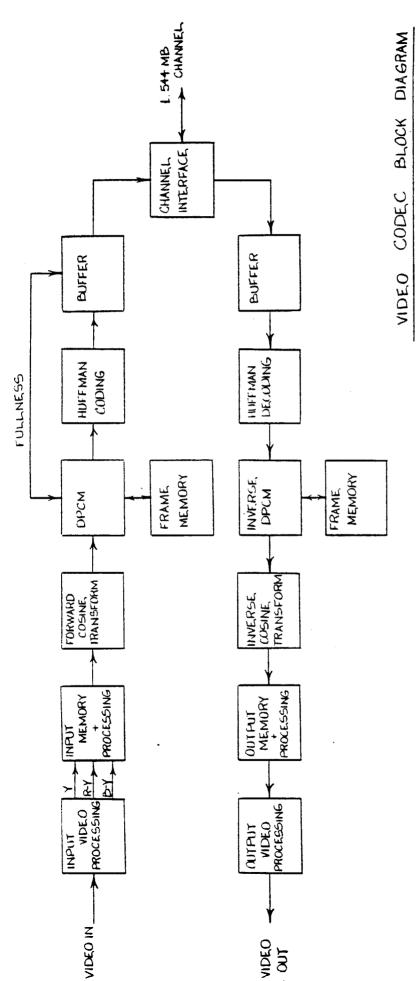


FIGURE 2.4
T1 FRAME FORMAT - MULTIPOINT CONFERENCING

13



10

FIGURE 3.1

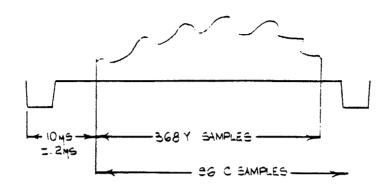


FIGURE 3.2 VIDEO SAMPLING WINDOW

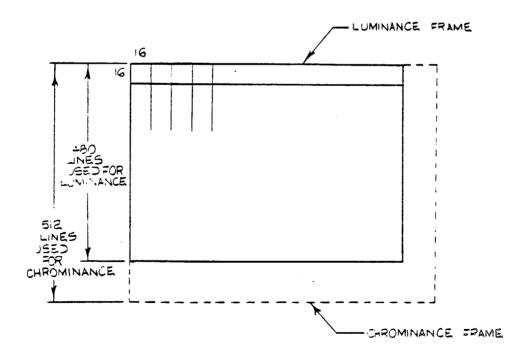


FIGURE 3.3 VIDEO FRAME FORMAT

ANNEX B

1. Introduction

CCITT Recommendations H.120 and H.130 have been established on the codec and frame structure for videoconferencing for Part 1 and Part 2a. However, Part 3 for intra-regional 1544 kbit/s transmission of videoconferencing signals that employ 525 lines and 60 field/s standard is still under study.

This contribution proposes a codec and frame structure for completing the Part 3 of Recommendations H.120 and H.130.

2. Outline

2.1 Overall Configuration

The codec overall blockdiagram is shown in Fig.1. The coder (Fig.1(a)) is composed of three blocks, video source coding, transmission coding and audio source coding.

In the video source coder, the input analog NTSC signal is first digitized, color decoded, and time division multilplexed. The prefilter removes noise components. The prefilter output is adaptive predictive encoded to remove redundancy. The prediction error signal is then entropy encoded. The entropy-encoded error signal is fed to buffer memory to smooth out the data rate to be transmitted. Buffer occupancy information is fed back to the adaptive predictive coder to prevent buffer overflow.

In the transmission coder, encoded video and audio signals are encrypted on an optional basis, and then forward error corrected and scrambled. The encoded video, encoded

audio and optional user data signals are multiplexed into 1544 kbit/s digital data.

In the audio source coder, the input audio signal is encoded into 64 kbit/s data. The encoded audio signal is delayed to compensate for the video source coding delay, and fed to the transmission coder.

The decoder (Fig.1(b)) carries out the reverse operation.

2.2 Video Input/Output

525/30 NTSC composite color/monochrome television signal.

2.3 Digital Output/Input

Conforming to CCITT Recommendation G.703.

2.4 Digital Video Format

Video sampling frequency is 4 fsc (color subcarrier frequency). The sampling clock is locked to the horizontal synchronization of the NTSC signal. An NTSC composite signal is converted to luminance (Y) and chrominance(C_1 and C_2) components, which are then multiplexed to TDM (Time Division Multiplex) format.

2.5 Coding Algorithm

Adaptive predictive coding:

- (1) interframe prediction with motion compensation, and
- (2) intraframe prediction

Entropy Coding:

- (1) variable word-length coding, and
- (2) run-length coding

2.6 Mode of Operation

Full duplex. Also one-way broadcasting mode is available.

2.7 Audio Channel

Monaural audio signal (50~7000 Hz).

2.8 Data Channel

Optional 64 kbit/s user data channel.

2.9 Transmission Error Correction

BCH forward error correcting code.

Demand refresh mode available.

2.10 Additional Facilities

Encryption, graphics mode and multipoint communication.

2.11 Codec Processing Delay

Approximately 165 ms.

3. Video Interface

The codec video interface is 525/30 NTSC color/monochrome television signal. Nominal amplitude is 1 V_{p-p} at 75 ohm unbalanced.

4. Video Source Coding

4.1 Analog-to-digital and Digital-to-analog Conversion

- (1) Sampling frequency $f_s = 910f_H$ $(f_H: Horizontal scanning line frequency),$
- (2) Precision=8 bits/sample(Linear PCM, two's complement),
- (3) Level Definition:

Sync tip level(-40 IRE)=-124(10000100), White level (100 IRE)= 72(01001000).

4.2 Color Decoding and Encoding

(1) Digital Demodulation and Separation at the Encoder.

Luminance component : Y

Chrominance components : C₁/C₂

(2) Digital Incorporation and Modulation at the Decoder.

4.3 TDM Signal (see Fig.2)

- (1) Time compression of C_1 and C_2 Compression Ratio = 6:1
- (2) Horizontal blanking interval of Y is substituted line-sequentially for either of the C_1 and C_2 signals.
- (3) TDM signal consists of Burst (7 samples), Color signal (64 samples of C_1 or C_2), and Luminance signal (384 samples).

4.4 Pre- and Post-filtering

- (1) At the Encoder;
 - · Temporal filtering for noise reduction,
 - ·Spatial filtering for folding-over suppression in subsampling.

- (2) At the Decoder;
 - · Spatial filtering for interpolating eliminated samples in Subsampling.
 - Spatio-temporal filtering for interpolating eliminated fields in Field Repetition.
 - •Temporal filtering for reduction of noise due to video source coding.

4.5 Predictive Coding

4.5.1 Source Coder and Decoder (See Fig. 3)

- (1) At the Coder;
 - ·Motion Vector Detection,
 - Predictive Encoder based upon adaptive intra-interframe prediction with motion compensation,
 - Entropy Coder for encoding prediction error signals, motion vector information, and coding mode commands.
- (2) At the Decoder;
 - ·Entropy Decoder,
 - ·Predictive Decoder.
- (3) Demand Refresh Capability for restoring identity of frame memory contents at the coder and the Decoder.

4.5.2 Adaptive Prediction (See Fig. 4 and Table-1)

- (1) Adaptive intra-interframe prediction with motion compensation
 - ·sample-by-sample prediction selection

when s=1

*Selection of Motion-compensated Interframe $\text{Prediction } (\hat{x}_{\text{M}} \text{=} \text{Z}^{-F+V}) \text{ , when s=0.}$

·s=Ra.Rb,

where $\mathbf{R}_{\mathbf{a}}$ is a reference signal for a previous sample, and $\mathbf{R}_{\mathbf{b}}$ for a previous line sample.

Definition of R;

$$R_{x} = \begin{cases} 0 & \text{if } |\hat{x}_{I} - \hat{x}| \ge |\hat{x}_{M} - \hat{x}|, \\ 1 & \text{otherwise,} \end{cases}$$

where \hat{x} is local decoder output signal for input x.

(3) For Subsampling mode,

$$\hat{x}_{I} = z^{-2},$$

 $S=R_{c}.R_{d}.R_{e}$, when samples a and b are eliminated.

(4) For Refresh mode,

 $R_{y}=1$ for all samples in lines under Refresh mode.

4.5.3 Motion Compensation (See Fig.5)

- ·Block size
- 8 lines x 16 samples
- · Motion Vector Direction

Positive for delay increasing direction in Interframe prediction loop.

4.5.4 Quantization (See Table-2)

(1) Four quantizing characteristics

$$Q_0$$
 (57 levels), Q_1 (57), Q_2 (51) and Q_3 (37).

4.5.5 Subsampling

- (1) Subsampling using 2:1 line quincunx pattern.
- (2) Interpolation of eliminated samples
 - ·Four neighboring samples (two-dimensional) for luminance signal interpolation.
 - •Two horizontally adjacent samples for color signal interpolation.

(See Table 1)

4.5.6. Field Repetition (See Table 1)

- (1) 2:1 Field Omission
- (2) Interpolation of omitted fields by vertical average using two lines in the previous field.

Same component $(C_1 \text{ or } C_2)$ in the previous field is used for color signal interpolation.

4.5.7. Stop of Coding

- (1) Coding operation halt for excessive generated information rate in order to prevent buffer memory from overflowing.
- (2) Both prediction error signal and motion vectors are forcibly set zero.

4.5.8 Refresh

- (1) Refresh for renewal of frame memory contents.
- (2) Intraframe prediction for Refresh
- (3) Two types of Refresh
 - ·Demand Refresh for communication with backward channel
 - ·Cyclic Refresh for broadcasting type communication without backward channel.

4.6 Coding Parameter Control

(1) Normal mode

full sampling

(2) Subsampling

2:1 sample elimination

(3) Field Repetition

2:1 field elimination

(4) Stop

Coding halt

(5) Refresh

renewal of frame memory contents

4.7 Entropy Coding

4.7.1 Entropy Coder (See Fig. 6)

- (1) Coding of Mode Commands
- (2) Compression of prediction error data
- (3) Compression of motion vector data
- (4) Multiplexing

4.7.2 Coding Mode Commands (See Fig. 7)

- (1) Data Structure
 - ·Frame Sync
 - ·Frame Mode Data
 - ·Line Sync
 - ·Line Mode Data
 - · Motion Vector Data
 - ·Prediction Error Data
- (2) Frame Mode Data

- .Frame Position to indicate the head line position in a video frame
- . Delay time in coder buffer memory
- · Demand Refresh (Execution and Request)
- ·Color/Monochrome State
- · Cyclic Refresh to indicate lines to be refreshed
- · Mode Flag for extension of mode information
- (3) Line Mode Data
 - ·Quantizing characteristics

$$q_0 \sim q_3$$

- Refresh
- ·Field Repetition
- ·Line Skip to indicate that there is a code representing the number of lines which can be skipped.
- (4) Motion Vector Data
 - ·Compressed Motion Vector Data
 preceded by its number (in Bytes)
 - · Entropy Coding
- (5) Prediction Error Data

 Entropy Coding

4.7.3 Prediction Error Coding

- (1) Variable Word-length coding for non-zero prediction error signal. (See Table 3)
 - The same code set is used for all quantizing characteristics.
- (2) Adaptive code switchig between Variable Word-length (V) code and Fixed Word-length (F) code on a sample-by-sample basis

ANNEX B

- ·Long V code is followed by F code
- ·F code representing small amplitude error is followed by . V code
- ·Switching threshold between Level Numbers 4 and 5. (See Fig. 8)
- (3) Run-length coding for zero prediction error signal (See Table 4)
 - ·Short run (Run-length RL<64) is expressed using a single code (Terminate code).
 - ·Long run (RL>65) is expressed using two codes, Make-up (MK) and Terminate codes.
- (4) Scanning method conversion from TV scan to Block scan, where scanning is completed within a two-dimensional block, thus resulting in longer runs.
- (5) Bridging of isolated zero amplitude error signal. V_0 or F_0 .
- (6) Eliminated samples in Subsampling are not coded.

4.7.4 Motion Vector Coding

- (1) Coding of vector difference

 Difference between two successive vectors
- (2) Variable Word-length coding for non-zero difference vector (See Fig. 9, Table 5)
- (3) Run-length coding for zero difference vector (See Table 5)
- (4) Transient code (TRANS) is inserted when Subsampling is turned on or off. The first TRANS code indicates start of Subsampling and the second end of it, and so on. (See Fig. 10)

4.8 Buffer Control

- (1) Delay time at the coder buffer memory is transmitted as a part of Frame Mode Data. Buffer memory operation at the decoder is controlled so that the total delay time at the coder and decoder is kept constant.
- (2) Buffer memory size
 180kbit/s for the coder buffer,
 more than 220kbit/s for the decoder buffer memory.
 (Total delay time is about 165ms.)
- (3) Underflow/Overflow Prevention

 For Underflow prevention, each zero amplitude prediction error signal is represented with variable word-length code V₀ instead of run length coding.

 For Overflow prevention, Stop Mode is applied, resulting in zero amptitude for all prediction error signal and motion vector.

5. Audio Coding

Audio signal bandlimited to 7kHz is encoded into 64 kbit/s. The encoding scheme is, at present, subband coding with ADPCM (SG. XVIII Delayed Contribution SE). However, provision is made for future use of more efficient coding.

6. Transmission Coding

6.1 Encryption

Video and audio signals can be independently encrypted.

6.2 Error Correction

(255, 239) double error correcting BCH code, along with interleaving.

6.3 Scrambling

8 stage scrambler.

6.4 Frame Structure

Details are shown in Annex.

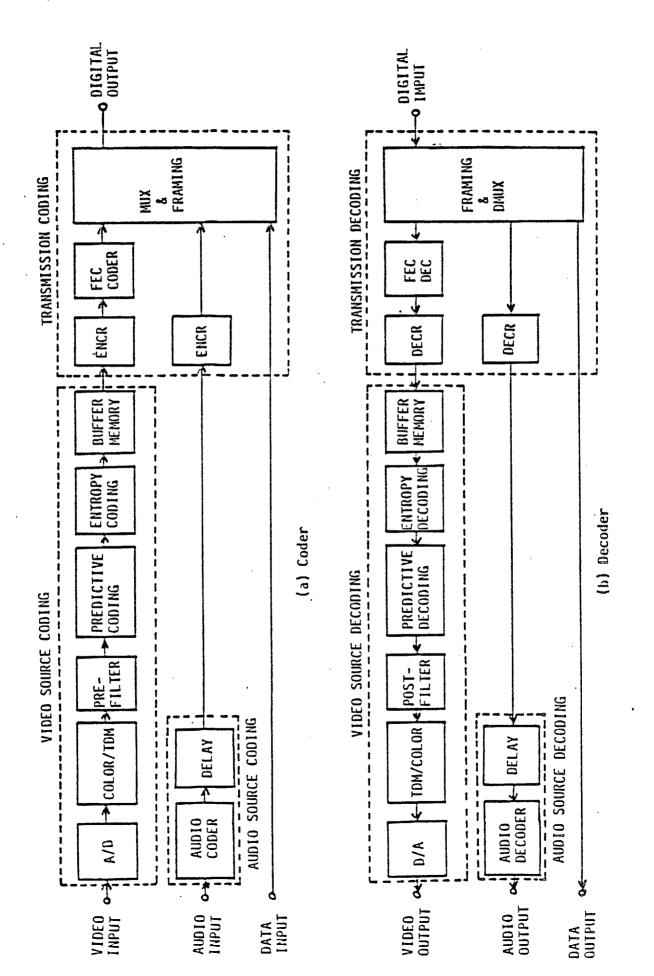


FIGURE 1. Codec blockdiagram

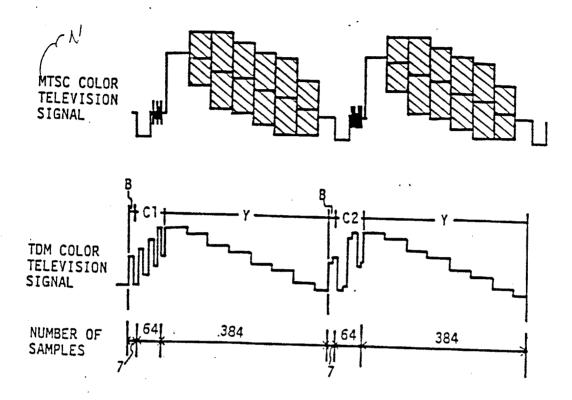
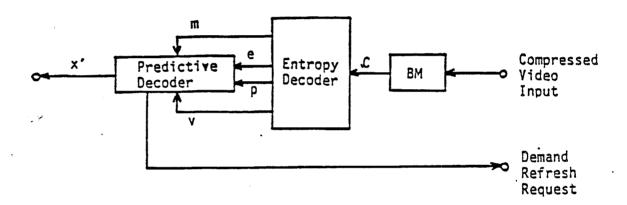


FIGURE 2. TDM Color Signal Format



x : input video signal

x': output video signal

m : coding mode

e : prediction error

v : motion vector

p : parity check information

C : compressed data

b : buffer memory occupancy imformation

BM : Buffer Memory

FIGURE 3. Source coder and decoder

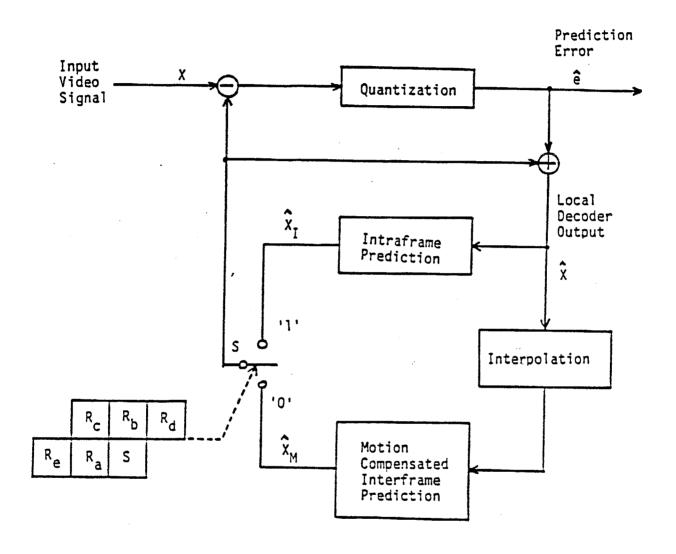
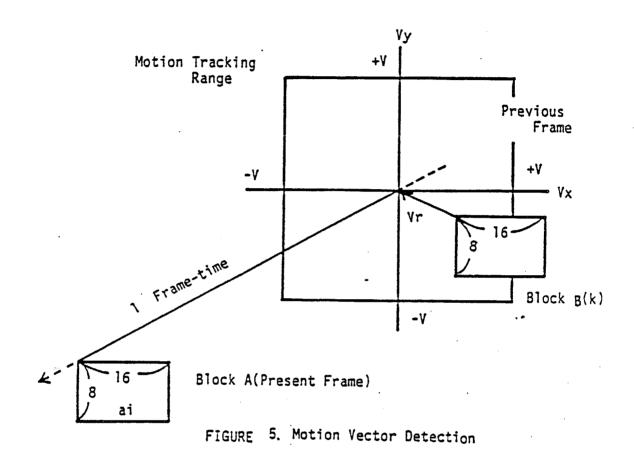
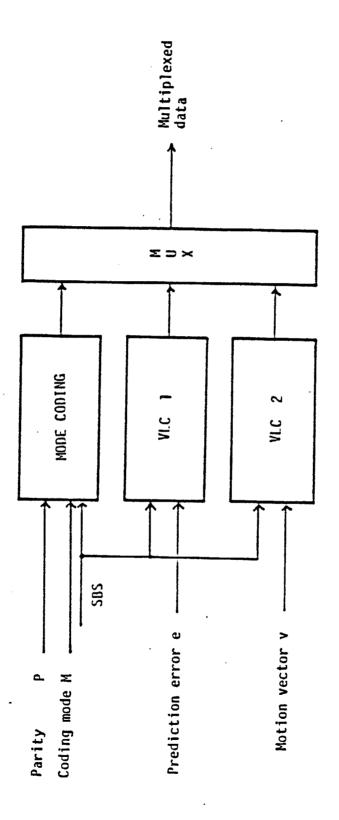


FIGURE 4. Adaptive Prediction





SBS : Subsampling VLC : Variable Wordlength coder

FIGURE 6. Entropy coder

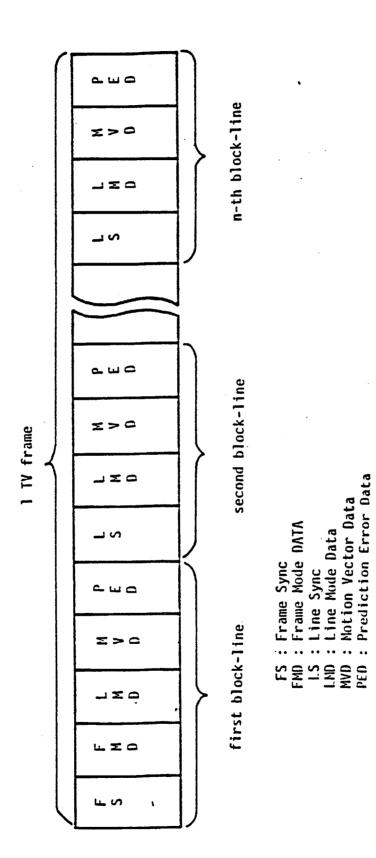
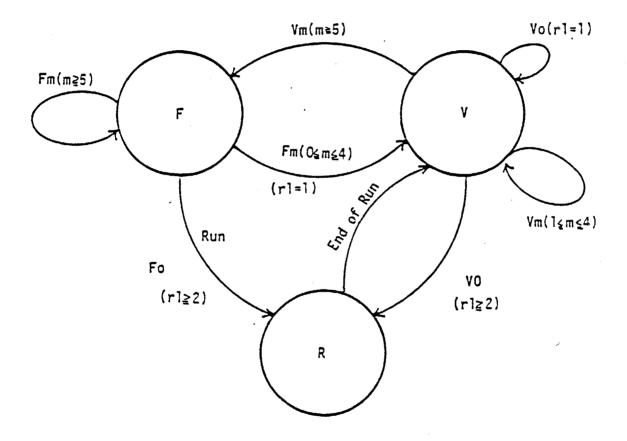


FIGURE 7. Multiplexed data format



Note: Run-length is represented by "r]".

Figure 8. Code Transition Rule for Prediction Error

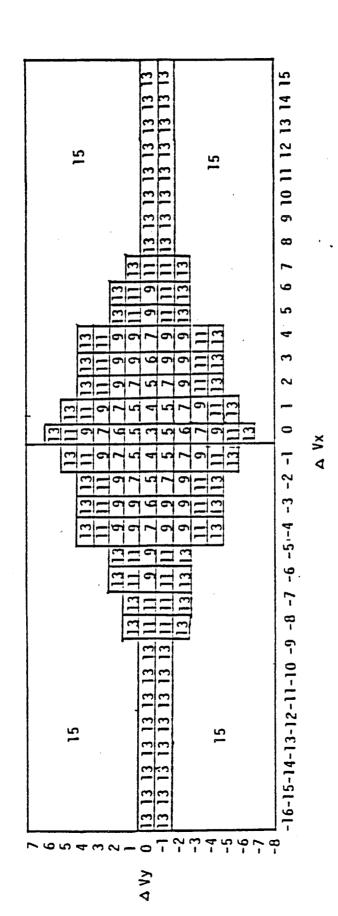
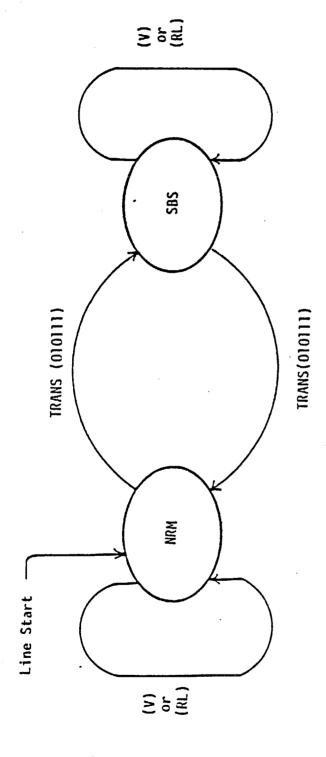


FIGURE 9. Word length for motion vector prediction errors



Note: The same motion vector coding method is applied

for both NRM and SBS modes.

Normal/Subsampling mode change

FIGURE 10. Code Transition rule for motion vector data and

TABLE 1. Prediction and interpolation functions.

	(2)				ield field			
(Z)1 SI	(z) ⁸ 1	·			irst f econd			
Interpolation Functions 1(2)	(z) ³ 1			$\frac{1}{2} \left\{ \frac{1}{2} \left(z^{-1} + z^{+1} \right) + \frac{1}{2} \left(z^{-1} + z^{+1} \right) \right\}$	Z ^{-283H} ;first field Z ^{-282H} ;second field			
ation F		-		1)	Z Z Z Z Z Z Z Z Z Z Z Z Z Z Z Z Z Z Z	_		11)
terpol	$I_{\gamma}(z)$		ı	₊ Z+ ₋ Z).	5211 ⁺ Z-2			$\frac{1}{2}(z^{-1}+z^{+1})$
I				$\frac{1}{2} \left(\frac{1}{2} \right) $	$\frac{1}{2} (z^{-262H_{+Z}-263H}) \begin{vmatrix} z^{-283H}; \text{first field} \\ z^{-282H}; \text{second field} \end{vmatrix}$			1 2
	P _B (Z)	j_2.						
		ζ.						
ediction Function P(Z)	P _c (2)							
Function		; S=1	. S=1	Fined)	ined)	Z-1.	2-2	ined)
ction 1	$P_{\gamma}(Z)$	z ⁻¹ z-F+V	Z-2 Z-F+V	(not defined)	(not defined)		.2	(not defined)
Predi	3			J	(u			ě)
				pa	eq	Ð		ed
Kind	ot Pel	Coded	Coded	Omitted	Omitted	Coded	Coded	Omitted
	ode				noi	ZZZ	v∝	Š
	Coding Mode	Normal	Sub came June		Field Repetilion		Refresh	

TABLE 2. Quantizing characteristics

		Q0			(21					Q2		
	Input Rar	nge 0	utput	Inp	ut Ra	nge	Output		Input	Ra	nge	Outpu	a [
			Level				Level					Leve	1
	0~	1	0		0~	3	0		0	~	4	0	
		2	1		4~	6	3		5	~	8	5	1. [
		3	2		7~	8	6		9	~	12	10	
	4~	5	3		9~	1 0	9		1 3	~	17	1 5	
	6~	7	5	1	1~	1 3	12		18	~	2 2	20	
	8 ~	9	7.	1	4~	16	15		2 3	~	27	25	
	10~	1 1	1 0	1	7~	19	18		28	~	3 2	3 0	1 [
	12~	1 4	13	2	0~	2 2	2 1		3 3	~	3 7	3 5	
	15~	17	16	2	3~	2 6	2 4		3 8	~	42	4 0] [
	18~	20	19	2	7~	3 0	28		4 3	~	47	4 5	Ī [
	21~	23	22	3	1~	3 4	3 2		4 8	~	5 2	5 0	1 [
	24~	26	2.5	3	5~	3 9	3 7		5 3	~	5 7	5 5	
	27~	2 9	28	4	0~	44	4 2		5 8	~	6 2	6.0] [
	30~	3 2	3 1	4	5~	4 9	47		63	~	67	6 5	Ī [
	33~	3 7	3 5	5	0~	5 41	5 2		6 8	~	72	70	Ī [
	38~	4 2	4 0	5	5~	5 9	5 7		73	~	77	7 5	ī [
	43~	4 8	4 5	6	0~	6 4	62		78	~	8 2	8 0	Ī [
	49~	5 4	5 1	6	5~	6 9	67		8 3	~	8 7	8 5	Ī
!	55~	6 0	5 7	7	0~	7 4		• •	88		9 2	9 0	
	61~	6 71	6 4	7	5~	7 9		. ,	9 3	~	9 7		
	68~	7 4	7 1		0~	8 41	8 2		98	~ 1	0 2		-
	75~	8 1	78		5~	8 91	8 7	. ,	103			1 0 5	•
		8 8	8 5			9 41		. ,				1 1 0	-
	89~		9 2			9 91		: 1				1 1 5	-
	96~1	<u>_</u>	·				102	}				121	-
	103~1						107					127	-
	110~1						1 1 3						. ن
}				1									

_								
I	пр	u t	R	an	ge		Ou	tp
							_	ev
		0	~			6	Ù	
		7	~		1	1		
	1	2	~		1	7		1
	1	8	~		2	4		2
	2	5	~		3	1		2
	3	2	~		3	8		3
	3	9	~		4	5		4
	4	6	~		5	2		4
	5	3	~		5	9		<u>4</u> 5
	6	0	~		6	6		6
		7	~		7	3		6 7 7
	6 7	4	~		8	0		7
	8	1	~		8	7		8
	8	8	~		9	4	İ	9
	9	5	~	1	0	1		9
1	0	2	~	1	0	8	1	0
1	0	9	~	1	1	5	1	1
1	1	6	~	1	2	3	1	1
<u>1</u>	2	4	~	2	5	5	1	2

Q3 ·

Note : Characteristics are symmeterical with respect to zero

TABLE 3. Variable length code for non-zero amplitude prediction errors

		1			
Level	Code	V Code	Level	Code	F Code
Number	Length		Number	Length	
V O	4	0 1 1 1	F0	4	0001
1	2	1 S	1 1	6	111115
2	5	0 1 1 0 S	2	6	111105
3	7	0101115	3 4	6	111015
4	7	0101105	4	6	11100S
5	8	01010115	5	6	110115
6	8	01010105	6	6	110105
7	8	01010015	7	6	110015
8	8	01010005	8	6	110005
9	9	010011115	9	6	101115
10	9	010011105	10	6	101105
1 1	9	010011018	1 1	6	101015
12	9	010011005	1 2	6	101005
1 3	9	010010115	13	6	100115
1.4	9	010010105	14	6	100105
1 5	9	010010015	15	6	100015
16	9	010010005	16	6	100005
17	10	0100011115	17	6	011115
18	10	0100011108	18	6	011105
19	10	0100011015	19	6	011018
20	10	0100011005	20	6	011005
2 1	10	0100010115	2 1	6	010115
2 2	10	0100011108	22	6	010105
23	10	0100010015	23	6	010015
2 4	10	0100010005	- 24		010005
25	10	0100001115	25	1	001115
26	10	0100001105			001105
27	10	0100001018	27	1	001015
28	10	0100001005		1	001005

Note: S denotes Sign.S=0 for positive, S=1 for negative

TABLE 4. Run length code for zero amplitude prediction errors

RL	Code	Code Word	Remark
	Length		
2	5	0 0 0 0 1	
3	5	0 0 0 0 0	
4	6	0 0 1 0 1 0	
5	6	0 0 1 0 0 1	
6	6	0 0 1 0 0 0	
7	7	0 0 1 0 1 1 1	
8~11	7	0 0 1 1 0 X X	$ X = 1 \cdot 1 - R \cdot L $
1 2	8 .	0 0 1 1 1 1 0 1	
1 3	8	0 0 1 1 1 1 0 0	
14~17	8	0 0 1 1 1 0 X X	X = 17 - RL
18~25	9	0 0 0 1 1 1 X X X X	X = 25 - RL
26~33	1 0	0 0 0 1 1 0 0 X X X	X = 3 3 - R L
34~37	1 0	0 0 0 1 0 1 0 0 X X	X = 37 - RL
38~64	1 2	0 0 0 1 0 0 1 X X X X X	X = 69 - RL
MK 1	1 3	0 0 1 0 1 1 0 Y Y Y Y Y Y	
MK2	14	0 0 1 1 1 1 1 1 1 Y Y Y Y Y Y	$Y = 0 \sim 63$
мкз	1 4	0 0 1 1 1 1 1 0 4 4 4 4 4 4	
MK 4~ 7	1 5	0 0 0 1 1 0 1 X X Y Y Y Y Y Y	X = 7 - MK
MK 8~15	16	0 0 0 1 0 1 1 X X X Y Y Y Y Y Y I	X = 15 - MK
MK16~19	16	0 0 0 1 0 1 0 1 1 X X Y Y Y Y Y Y Y	X = 19 - MK
MK20~34	18	0 0 0 1 0 0 0 1 X X X X Y Y Y Y Y Y	X = 35 - MK
MK35~49	19	0 0 0 1 0 0 0 0 1 X X X X Y Y Y Y Y	Y X = 5 0 - MK
MK50~57	19	0 0 0 1 0 0 0 0 0 1 X X X Y Y Y!Y Y	Y X = 57 - MK

Note: $RL=64\times(MK \text{ number})+1+Y,0\leq Y\leq 63$

ANNEX B

TABLES. Variable length code and run length code for motion vector data

. 4 V x	ΔVy	Code	Code Word	Number
		Length		of Codes
± 1	0	4	0 0 1 S x	2
± 1	± 1	5	1 1 1 SxSy	
0	± 1	5	1 1 0 1 Sy	8
±2	0	5	1 1 0 0 Sx	
.0	± 2	6	1 0 1 1 1 Sy	4
± 3	0	6	1 0 1 1 0 S x	
± 1	± 2	7	1 0 0 1 1 SxSy	
±2	. ±1	7	100 10 SxSy	12
0	± 3	7	100 011 Sy	
±4	0	7	100010 Sx	
± 3	± 1	9	1010 111SxSy	
± 1	±3	9	1010110SxSy	
±2	± 2	9	1 0 1 0 1 0 1 SxSy	
± 3	± 2	9	1010100SxSy	
±4	± 1	9	1 0 1 0 0 1 1 S x S y	3 0
±4.	±2	9	1010010SxSy	
±5	0	9	10100011 Sx	
±6	0	9	10100010 Sx	_
0	± 4	9	10100001 Sx	
-8~7	-5~+5	1 1	100001 XXXX Sy [X]=\(\Delta\) Vx	3 2
	(See Fig.13)			
-16~15	- 6~+6	1 3	0 1 0 0 0 0 1 X X X X X Sy [X]= \(\times V \times \)	6 4
	(See Fig.13)			
-16~15		1 5	$1\ 0\ 0\ 0\ 0\ X\ X\ X\ X\ Y\ Y\ Y\ Y\ [\ X\] = \triangle V\ x$	360
	(See`Fig.13)		[Y]=△Vy	

RL	Code	Code	Word	Number
·	Length		·	of Codes
1	3	000		1
2	4	0 1 1 1		1
3~6	6	0 1 1 0 X X	XX = 6 - RL	4
7~12	7	0 1 0 1 X X X	X X X = 12 - R L	6
13~20	8	0 1 0 0 1 X X X	XXX = 20 - RL	8
21~28	9	0 1 0 0 0 1 X X	XXX = 28 - RL	8
				
TRANS	6	0 1 0 1 1 1		1

ANNEX B

ANNEX

Frame Structure for Use with the Proposed Codec

1. Outline

This annex summarizes the frame structure for use with the proposed codec operating at 1544 kbit/s.

A frame consists of 193 bits. A multiframe and a supermultiframe consist of 12 frames and 4 multiframes, respectively. To satisfy the format restrictions specified in Recommendation G.703 (one's density), provision is made by means of a stuffing system. Three frames are observed, and when needed, stuffing is carried out to control one's density.

2. Frame Structure

Figure A1 shows the line data frame format. The upper figure depicts the normal mode, where no stuffing is carried out. On the other hand, the lower figure shows the stuffing mode, where 1's are stuffed at the bit positions indicated by dashed lines.

Figure A2 shows the multiframe structure.

The abbreviated names for the bits in Figures A1 and A2 are shown in Table A1.

Table A1 Bit Allocations

Abbr.	Information
F _T	Transmission Line Frame Sync.
F _D	Codec Frame Sync.
·	
x ₀ -x ₄	Stuffing Flags
	Majority Decision (3 out of 5)
	1 : Normal Mode
	2 : Stuffing Mode
С	Codec-to-Codec Information
A ₁ -A ₈	Audio Data
. מ	Audio Frame Information
V	Video Data

							_		_		141112	
	FD X2 X3						4.8	1 \$\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \				
	V 32 V I FT FD	X4 1 C A1 X 1 FT FD X0 X1		12	FT10 FT11 FT12	1 X0 X1 X2 X3 X4 1	12	S4 S5 F2 S				
- 193BITS	1 U 8 V V V V V V V V V	X2 X3 C A1 B I FI FD X4			F16 F17 F18 F19 (1)	 1 X0 X1 X2 X3 X4	·	F1 S1 S2 S3 (1) (0)		36 42 48	FG F7 F8	(1) (1) (1)
	C A1 A2 1 A3 A	1 C A 1 3 1 L FT FD X2 X3		3	FT3 FT4 FT5 (0) (1)	 X2 X3 X4 1 X0 X1 X2 X3 X4		S3 S4 S5) (0) (1) (1)		18 24 30	F3 F4 F5	(1) (0) (1)
<u></u>	TETED XO X1	 1 FT FD XO X1	A. J. (Apr.) 174	icu -	F11 F12 (1)	 X 0 X	- 5	S1 S2 (1) (0)	·	6 12	FI F2	(0) (0)
	DATA	DATA			T SEQUENCE	X SEQUENCE		FD SEQUENCE			a JN a II C a o	7 2540510

Fig. A2 Multiframe Structure

✓ 193BITS	<-193BITS><193BITS>
 X0 X1 C A1 A2 1 A3 X A8	FT FD X2 X3 C
 STUFFING MODE	
 FT FD X0 X1 C A1 A2 A3 A8 D 14 15 16 17 19 1	2 3 4 5 193 2 3 4 5 193 FT FD X4 C

160 152 STUFFING BIT POSITIONS (Indicated 123 128 136 144 32 र ८ 104 112 œ

Fig. Al Frame Format