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Interactive systems for digital television distribution

**Downstream RF interface for cable modem
termination systems**

ITU-T Recommendation J.210



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Downstream RF interface for cable modem termination systems

Summary

This Recommendation defines the downstream radio frequency interface (DRFI) specifications for:

- an edgeQAM (EQAM) modular device; or
- an integrated cable modem termination system (CMTS) with multiple downstream channels per RF port; or
- an integrated CMTS beyond DOCSIS 2.0.

Source

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ITU-T Recommendation J.210

Downstream RF interface for cable modem termination systems

1 Scope

1.1 Scope

The DOCSIS Rec. J.112 and [ITU-T J.122] define the requirements for the two fundamental components that comprise a high-speed data-over-cable system: the cable modem (CM) and the cable modem termination system (CMTS). This Recommendation provides physical layer requirements for CMTS transmitters in the DOCSIS architecture. It applies to head-end components built according to the M-CMTS architecture ([ITU-T J.212] and [ITU-T J.211]) as well as integrated CMTS systems.

This Recommendation defines the downstream radio frequency interface (DRFI) specifications for:

- an edgeQAM (EQAM) modular device; or
- an integrated cable modem termination system (CMTS) with multiple downstream channels per RF port; or
- an integrated CMTS beyond DOCSIS 2.0.

1.2 World wide use

There are differences in the cable spectrum planning practices adopted for different networks in the world. Therefore three options for physical layer technology are included, which have equal priority and are not required to be interoperable. One technology option is based on the downstream multi-program television distribution that is deployed in the Americas using 6 MHz channels. The second technology option is based on the corresponding European multi-program television distribution using 8 MHz channels. The third technology option is based on the corresponding Japanese multi-program television distribution using 6 MHz channels. All options have the same status, notwithstanding that the document structure does not reflect this equal priority. The first of these options is defined in clauses 5, 6, 7, whereas the second and the third are defined by replacing the content of those clauses with the content of Annexes A and B respectively. Correspondingly, [ITU-T J.83-B] and [CEA-542-B] apply only to the first option, [ETSI EN 300 429] applies only to the second, and [ITU-T J.83-C] applies only to the third option. Compliance with this Recommendation requires compliance with at least one of these implementations, not necessarily with more than one. It is not required that equipment built to one option shall interoperate with equipment built to the others.

A DRFI-compliant device may be a single-channel only device, or it may be a multiple-channel device capable of generating one or multiple downstream RF carriers simultaneously on one RF output port. An EQAM may be a module of a modular cable modem termination system (M-CMTS) and be used for delivering a high-speed data service or it may serve as a component of a digital video or video-on-demand (VoD) system, delivering high quality digital video to subscribers. These specifications are crafted to enable an EQAM to be used without restriction in either or both service delivery scenarios simultaneously. "Simultaneous" in the early deployments means that if a RF output port has multiple QAM channels, some channel(s) may be delivering high-speed data while some other channel(s) may be delivering digital video. This specification enables future uses, wherein a single QAM channel may share bandwidth between high-speed data and digital video in the same MPEG transport stream.

Conceptually, an EQAM will accept input via an Ethernet link, integrate the incoming data into an MPEG transport stream, modulate one of a plurality of RF carriers, per these specifications, and deliver the carrier to a single RF output connector shared in common with all modulators.

Conceivably, a single EQAM RF channel could be used for data and video simultaneously. The reason that an EQAM RF channel can be used for either is that both digital video and DOCSIS data downstream channels are based on [ITU-T J.83-B] for cable networks in the Americas, [ETSI EN 300 429] for cable networks deployed in Europe and [ITU-T J.83-C] for cable networks in Japan. On downstream channels complying to [ITU-T J.83-B], typically, the only difference between an EQAM RF channel operating in a video mode and an EQAM RF channel operating in DOCSIS data mode is the interleaver depth (see clauses 6.3.1 and 6.3.3). DOCSIS data runs in a low latency mode using a shallow interleaver depth at the cost of some burst protection. DOCSIS data can do this because if a transmission error occurs, the higher layer protocols will request retransmission of the missing data. For video, the sequence of frames in the program is both time sensitive and order sensitive and cannot be retransmitted. For this reason, video uses a deeper interleaver depth to provide more extensive burst protection and deliver more of the program content without loss. The penalty video pays is in latency. The entire program content is delayed by a few milliseconds, typically, and is invisible to the viewers of the program. The conflicting demands for interleaver depth are what prevent a single EQAM RF channel from being used optimally for video and DOCSIS data simultaneously. A traditional integrated CMTS, however, is used solely for DOCSIS data.

2 References

2.1 Normative references

The following ITU-T Recommendations and other references contain provisions which, through reference in this text, constitute provisions of this Recommendation. At the time of publication, the editions indicated were valid. All Recommendations and other references are subject to revision; users of this Recommendation are therefore encouraged to investigate the possibility of applying the most recent edition of the Recommendations and other references listed below. A list of the currently valid ITU-T Recommendations is regularly published. The reference to a document within this Recommendation does not give it, as a stand-alone document, the status of a Recommendation.

- [ITU-T H.222.0] ITU-T Recommendation H.222.0 (2006) | ISO/IEC 13818-1:2006, *Information technology – Generic coding of moving pictures and associated audio information: Systems.*
- [ITU-T J.83-B] ITU-T Recommendation J.83 (1997), *Digital multi-programme systems for television, sound and data services for cable distribution. Annex B: Digital multi-programme System B.*
- [ITU-T J.83-C] ITU-T Recommendation J.83 (1997), *Digital multi-programme systems for television, sound and data services for cable distribution. Annex C: Digital multi-programme System C.*
- [ITU-T J.122] ITU-T Recommendation J.122 (2002), *Second-generation transmission systems for interactive cable television services – IP cable modems.*
- [ITU-T J.211] ITU-T Recommendation J.211 (2006), *Timing interface for cable modem termination systems.*
- [ETSI EN 300 429] ETSI EN 300 429 V1.2.1 (1998), *Digital Video Broadcasting (DVB); Framing structure, channel coding and modulation for cable systems.*
- [IEC 60169-24] IEC 60169-24 (1991), *Radio-frequency connectors – Part 24: Radio-frequency coaxial connectors with screw coupling, typically for use in 75 ohm cable distribution systems (Type F).*

2.2 Informative References

- [ITU-T J.212] ITU-T Recommendation J.212 (2006), *Downstream external physical layer interface for modular cable modem termination systems*.
- [NSI] *Cable Modem Termination System Network Side Interface*, SP-CMTS-NSI-I01-960702, 2 July 1996, Cable Television Laboratories, Inc.
- [M-OSSI] *Modular CMTS Operations Support System Interface*, CM-SP-M-OSSI-I02-051209, 9 December 2005, Cable Television Laboratories, Inc.
- [CEA-542-B] CEA-542-B: *CEA Standard: Cable Television Channel Identification Plan*, July 2003.
- [CMCI] *Cable Modem to Customer Premises Equipment Interface*, CM-SP-CMCI-I10-050408, 8 April 2005, Cable Television Laboratories, Inc.
- [ERMI] *Edge Resource Manager Interface*, CM-SP-ERMI-I02-051209, 9 December 2005, Cable Television Laboratories, Inc.
- [Article 23-(1)] *Regulations for Enforcement of the Cable Television Article 23-(1)*, Ministry of Internal Affairs and Communications (MIC), Japan.

2.3 Reference acquisition

- Cable Television Laboratories, Inc., <http://www.cablelabs.com/>
- EIA: Electronic Industries Alliance, http://www.eia.org/new_contact/
- ETSI: European Telecommunications Standards Institute, http://www.etsi.org/services_products/freestandard/home.htm
- ITU: International Telecommunication Union (ITU), <http://www.itu.int/home/contact/index.html>
- ISO: International Organization for Standardization (ISO), <http://www.iso.org/iso/en/xsite/contact/contact.html>
- MIC: Ministry of Internal Affairs and Communications (MIC), http://www.soumu.go.jp/joho_tsusin/eng/index.html

3 Terms and definitions

This Recommendation defines the following terms:

3.1 ceiling (ceil): The ceiling function rounds a number up to the nearest integer or nearest multiple of significance. Use: Ceiling (number, significance).

3.2 cable modem (CM): A modulator-demodulator at subscriber locations intended for use in conveying data communications on a cable television system.

3.3 customer premises equipment (CPE): Equipment at the end user's premises; may be provided by the service provider.

3.4 carrier-to-noise ratio (C/N or CNR): The ratio of signal power to noise power in a defined measurement bandwidth. For digital modulation, $CNR = E_s/N_0$, the energy-per symbol to noise-density ratio; the signal power is measured in the occupied bandwidth, and the noise power is normalized to the modulation-rate bandwidth. For analog NTSC video modulation, the noise measurement bandwidth is 4 MHz.

3.5 decibels (dB): Ratio of two power levels expressed mathematically as $dB = 10\log_{10}(P_{OUT}/P_{IN})$.

- 3.6 decibel-millivolt (dBmV):** Unit of RF power expressed in decibels relative to 1 millivolt over 75 ohms, where $\text{dBmV} = 20\log_{10}(\text{value in mV}/1 \text{ mV})$.
- 3.7 decibel-microvolt (dBμV):** Unit of RF power expressed in decibels relative to 1 microvolt over 75 ohms, where $\text{dB}\mu\text{V} = 20\log_{10}(\text{value in }\mu\text{V}/1 \mu\text{V})$.
- 3.8 Electronic Industries Alliance (EIA):** A voluntary body of manufacturers which, among other activities, prepares and publishes standards.
- 3.9 edgeQAM modulator (EQAM):** A head end or hub device that receives packets of digital video or data. It re-packetizes the video or data into an MPEG transport stream and digitally modulates the digital transport stream onto a downstream RF carrier using quadrature amplitude modulation (QAM).
- 3.10 forward error correction (FEC):** A class of methods for controlling errors in a communication system. FEC sends parity information with the data which can be used by the receiver to check and correct the data.
- 3.11 harmonic related carriers (HRC):** A method of spacing channels on a cable television system with all carriers related to a common reference.
- 3.12 hybrid fibre/coax system (HFC):** A broadband bidirectional shared-media transmission system using optical fibre trunks between the head-end and the fibre nodes, and coaxial cable distribution from the fibre nodes to the customer locations.
- 3.13 incremental related carriers (IRC):** A method of spacing NTSC television channels on a cable television system in which all channels are offset up 12.5 kHz with respect to the [CEA-542-B] standard channel plan except for channels 5 and 6.
- 3.14 media access control (MAC):** Used to refer to the layer 2 element of the system which would include DOCSIS framing and signalling.
- 3.15 modulation error ratio (MER):** The ratio of the average symbol power to average error power.
- 3.16 M/N:** Relationship of integer numbers M,N that represents the ratio of the downstream symbol clock rate to the DOCSIS master clock rate.
- 3.17 multiple system operator (MSO):** A corporate entity that owns and/or operates more than one cable system.
- 3.18 national television systems committee (NTSC):** Committee which defined the analog, colour television, broadcast standards in North America. The standards television 525-line video format for North American television transmission is named after this committee.
- 3.19 NGNA LLC:** Company formed by cable operators to define a next-generation network architecture for future cable industry market and business requirements.
- 3.20 physical media dependent sublayer (PMD):** A sublayer of the physical layer which is concerned with transmitting bits or groups of bits over particular types of transmission link between open systems and which entails electrical, mechanical and handshaking procedures.
- 3.21 QAM channel (QAM ch):** Analog RF channel that uses quadrature amplitude modulation (QAM) to convey information.
- 3.22 quadrature amplitude modulation (QAM):** A modulation technique in which an analog signal's amplitude and phase vary to convey information, such as digital data.
- 3.23 radio frequency (RF):** A portion of the electromagnetic spectrum from a few kilohertz to just below the frequency of infrared light.

- 3.24 radio frequency interface (RFI):** Term encompassing the downstream and the upstream radio frequency interfaces.
- 3.25 root mean square (RMS):** Square root of the mean value squared a function.
- 3.26 self-aggregation:** Method used to compute the headend noise floor by summing measured noise from a single device over a specified output frequency range.
- 3.27 standard channel plan (STD):** Method of spacing NTSC television channels on a cable television system defined in [CEA-542-B].
- 3.28 upstream channel descriptor (UCD):** The MAC management message used to communicate the characteristics of the upstream physical layer to the cable modems.
- 3.29 video-on-demand (VoD):** System that enables individuals to select and watch video content over a network through an interactive television system.

4 Acronyms, abbreviations and conventions

4.1 Acronyms and abbreviations

This Recommendation uses the following abbreviations and acronyms:

CMCI	Cable Modem to CPE Interface
CMTS	Cable Modem Termination System
CW	Continuous Wave
dBc	Decibels relative to carrier power
DEPI	Downstream External-PHY Interface
DOCSIS	Data-Over-Cable Service Interface Specification
DRFI	Downstream Radio Frequency Interface
DTI	DOCSIS Timing Interface
ERMI	Edge Resource Manager Interface
M-CMTS	Modular Cable Modem Termination System
MPEG	Moving Picture Experts Group
NGNA	Next Generation Network Architecture, see <i>NGNA LLC</i>
OSSI	Operations Support System Interface
PHY	Physical Layer
ppm	parts per million
Q	Quadrature modulation component
S-CDMA	Synchronous Code Division Multiple Access

4.2 Conventions

Throughout this Recommendation, the words that are used to define the significance of particular requirements are capitalized. These words are:

"MUST" or "SHALL"	This word or the adjective "REQUIRED" means that the item is an absolute requirement of this Recommendation.
"MUST NOT" or "SHALL NOT"	This phrase means that the item is an absolute prohibition of this Recommendation.

"SHOULD"	This word or the adjective "RECOMMENDED" means that there may exist valid reasons in particular circumstances to ignore this item, but the full implications should be understood and the case carefully weighed before choosing a different course.
"SHOULD NOT"	This phrase means that there may exist valid reasons in particular circumstances when the listed behaviour is acceptable or even useful, but the full implications should be understood and the case carefully weighed before implementing any behaviour described with this label.
"MAY"	This word or the adjective "OPTIONAL" means that this item is truly optional. One vendor may choose to include the item because a particular marketplace requires it or because it enhances the product, for example; another vendor may omit the same item.

5 Functional assumptions

This clause describes the characteristics of a cable television plant, assumed to be for the purpose of operating a data-over-cable system. It is not a description of EQAM or CMTS parameters. The data-over-cable system **MUST** be interoperable within the environment described in this clause.

Whenever there is a reference in this clause to frequency plans or compatibility with other services, or conflicts with any legal requirement for the area of operation, the latter shall take precedence. Any reference to NTSC analog signals in 6 MHz channels does not imply that such signals are physically present.

5.1 Broadband access network

A coaxial-based broadband access network is assumed. This may take the form of either an all-coax or hybrid fibre/coax (HFC) network. The generic term "cable network" is used here to cover all cases.

A cable network uses a shared-medium, "tree-and-branch" architecture, with analog transmission. The key functional characteristics assumed in this Recommendation are the following:

- Two-way transmission;
- A maximum optical/electrical spacing between the DRFI-compliant device and the most distant CM of 100 miles in each direction, although typical maximum separation may be 10-15 miles;
- A maximum differential optical/electrical spacing between the DRFI-compliant device and the closest and most distant modems of 100 miles in each direction, although this would typically be limited to 15 miles.

At a propagation velocity in fibre of approximately 1.5 ft/ns, 100 miles of fibre in each direction results in a round-trip delay of approximately 1.6 ms. For further information, see Appendix VIII of [ITU-T J.122].

5.2 Equipment assumptions

5.2.1 Frequency plan

In the downstream direction, the cable system is assumed to have a passband with a lower edge between 50 and 54 MHz and an upper edge that is implementation-dependent but is typically in the range of 300 to 870 MHz. Within that passband, NTSC analog television signals in 6-MHz channels

are assumed present on the Standard (STD), HRC or IRC frequency plans, as well as other narrow-band and wideband digital signals.

5.2.2 Compatibility with other services

The CM and EQAM or CMTS MUST coexist with the other services on the cable network, for example:

- a) They MUST be interoperable in the cable spectrum assigned for EQAM or CMTS-CM interoperation, while the balance of the cable spectrum is occupied by any combination of television and other signals; and
- b) They MUST NOT cause harmful interference to any other services that are assigned to the cable network in spectrum outside of that allocated to the EQAM or CMTS. The latter is understood as:
 - 1) No measurable degradation (highest level of compatibility);
 - 2) No degradation below the perceptible level of impairments for all services (standard or medium level of compatibility); or
 - 3) No degradation below the minimal standards accepted by the industry (for example, FCC for analog video services) or other service providers (minimal level of compatibility).

5.2.3 Fault isolation impact on other users

As downstream transmissions are on a shared-media, point-to-multipoint system, fault-isolation procedures should take into account the potential harmful impact of faults and fault-isolation procedures on numerous users of the data-over-cable, video, and other services.

For the interpretation of harmful impact, see clause 5.2.2 above.

5.3 Downstream plant assumptions

The DRFI specifications have been developed with the downstream plant assumptions of this clause.

5.3.1 Transmission levels

The nominal power level of the downstream RF signal(s) within a 6-MHz channel (average power) is targeted to be in the range: -10 dBc to -6 dBc, relative to analog video carrier level (peak power) and will normally not exceed analog video carrier level.

5.3.2 Frequency inversion

There will be no frequency inversion in the transmission path in either the downstream or the upstream directions (i.e., a positive change in frequency at the input to the cable network will result in a positive change in frequency at the output).

5.3.3 Analog and digital channel line-up

In developing this Recommendation, it was assumed that a maximum of 119 digital channels would be deployed in a headend. For the purposes of calculating CNR, protection for analog channels, it was assumed that analog channels are placed at lower frequencies in the channel line-up, versus digital channels.

5.3.4 Analog protection goal

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 60 dB for systems deploying up to 119 DRFI-compliant QAM channels.

This Recommendation assumes that the transmitted power level of the digital channels will be 6 dB below the peak envelope power of the visual signal of analog channels, which is the typical

condition for 256 QAM transmission. It is further assumed that the channel lineup will place analog channels at lower frequencies versus digital channels. An adjustment of $10 \cdot \log_{10}(6 \text{ MHz}/4 \text{ MHz})$ is used to account for the difference in noise bandwidth of digital channels, versus analog channels. With the assumptions above, for a 119 QAM channel system, the specification in item 5 of Table 6-5 equates to an analog CNR protection of 60 dB.

6 Physical media dependent sublayer specification

6.1 Scope

This clause applies to the first technology option referred to in clause 1. For the second option, refer to Annex A.

This Recommendation defines the electrical characteristics of the Downstream Radio Frequency Interface (DRFI) of a cable modem termination system (CMTS), or an edgeQAM (EQAM). It is the intent of this Recommendation to define an interoperable DRFI-compliant device, such that any implementation of a CM can work with any EQAM or CMTS. It is not the intent of this Recommendation to imply any specific implementation. Figure 6-1 shows the M-CMTS structure and interfaces.

Whenever there is a reference in this clause to spurious emissions, conflicts with any legal requirement for the area of operation, the latter shall take precedence.

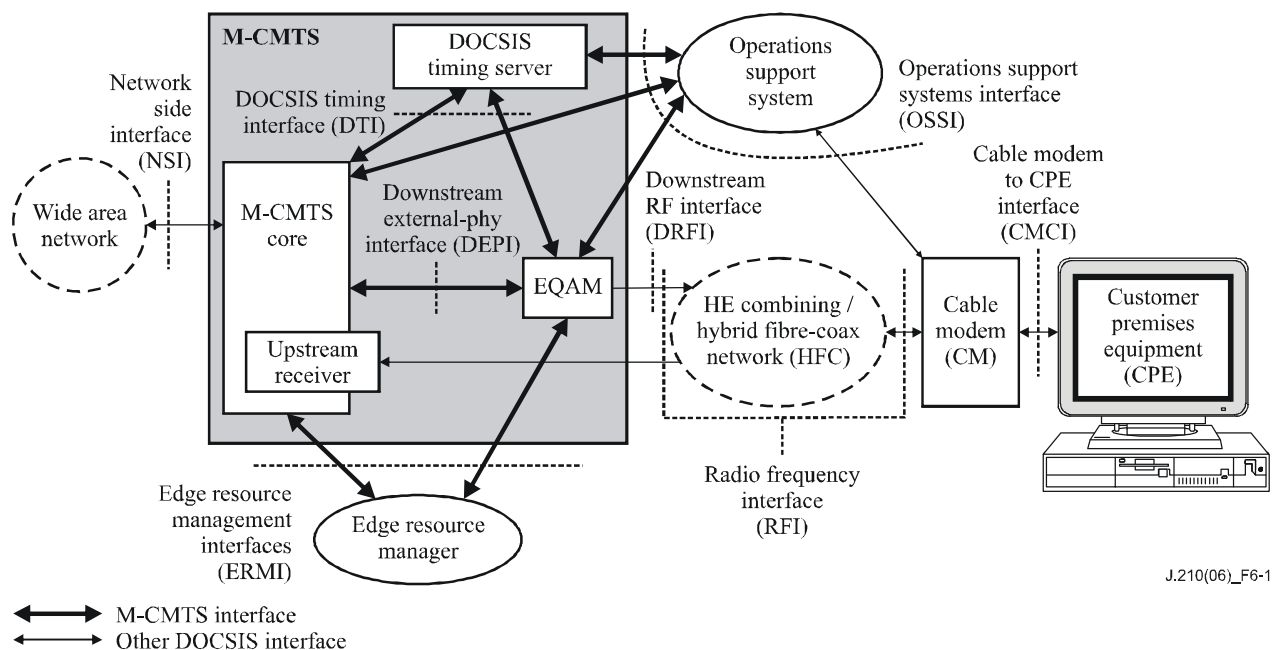


Figure 6-1 – Logical view of modular CMTS and interfaces

The CMTS network side interface (NSI), Modular CMTS operations support system interface (M-OSSI), radio frequency interface (RFI), and the cable modem to CPE interface (CMCI) are documented in existing DOCSIS Recommendations (see clause 2.2). The DOCSIS timing interface (DTI), downstream external-PHY interface [ITU-T J.212], downstream radio frequency interface (this Recommendation), and edge resource manager interface (ERMI) require new specifications specific to the M-CMTS in a next generation network architecture (NGNA) environment.

6.2 EdgeQAM (EQAM) differences from CMTS

The EQAM is primarily the RF modulation and transmission module extracted from a consolidated CMTS. Because the CMTS has been divided into constituent parts into the modules, the EQAM needs to have a new interface to the modular-CMTS (M-CMTS) MAC module. That new interface is an Ethernet interface, as specified in [ITU-T J.212], needed to communicate with the now remote EQAM. DEPI constructs, semantics, and syntax, as well as any new EQAM components and processing, are defined in the DEPI documentation.

EQAMs may also interface to video servers, via the Ethernet interface, and provide a downstream RF transmission to deliver digital video services. The protocols necessary to implement video services over EQAMs are out of the scope of this Recommendation.

Several new features are supported in this Recommendation. The DOCSIS 1.x and 2.0 Recommendations do not reflect the ability of vendors to support multiple RF channels per physical RF port. This Recommendation presents the requirements and optional functions that enable an EQAM, or a CMTS, with multiple channels per RF port to be tested, measured and, if successful, qualified.

For an M-CMTS, module synchronization is not as easy as with an integrated CMTS. A DRFI-compliant EQAM has a timing port on it that enables a high precision (DTI) to be used to distribute a common clock and timing signals. This permits the EQAM to be used in all modes, including S-CDMA mode, because of the high stability and low jitter of the external clock and distribution system. The DOCSIS Timing Interface is defined in [ITU-T J.211].

6.3 Downstream

6.3.1 Downstream protocol

The downstream PMD sublayer MUST conform to [ITU-T J.83-B], except for clause B.6.2. Interleaver depths are defined in clause 6.3.3. The applicability of a particular interleaver depth depends on the data service provided on a particular QAM RF channel. Applicability of interleaver depths for service delivery, other than DOCSIS high-speed data, is beyond the scope of this Recommendation.

6.3.2 Spectrum format

The downstream modulator for each QAM channel of the EQAM, or CMTS, MUST provide operation with the RF signal format of $S(t) = I(t) \cdot \cos(\omega t) + Q(t) \cdot \sin(\omega t)$, where t denotes time, ω denotes RF angular frequency, and where $I(t)$ and $Q(t)$ are the respective Root-Nyquist filtered baseband quadrature components of the constellation, as specified in [ITU-T J.83-B].

6.3.3 Scaleable interleaving to support video and high-speed data services

The CMTS or EQAM downstream PMD sublayer MUST support a variable-depth interleaver. [ITU-T J.83-B] defines the variable interleaver depths in Table B.2/J.83, Level 2 interleaving.

A CMTS or EQAM MUST support the set of interleaver depths described in Tables 6-1 and 6-2. Requirements for operational availability of interleaver depths are given in clause 6.3.5.1.2, item 1.

Table 6-1 – Low latency interleaver depths

Control word	Interleaver taps	Interleaver increment	64QAM 5.056941 Msym/s 6 bits per symbol		256QAM 5.360537 Msym/s 8 bits per symbol	
			Burst protection	Latency	Burst protection	Latency
Four bits	I	J				
1001	8	16	5.9 μ s	0.22 ms	4.1 μ s	0.15 ms
0111	16	8	12 μ s	0.48 ms	8.2 μ s	0.33 ms
0101	32	4	24 μ s	0.98 ms	16 μ s	0.68 ms
0011	64	2	47 μ s	2.0 ms	33 μ s	1.4 ms
0001	128	1	95 μ s	4.0 ms	66 μ s	2.8 ms

Table 6-2 – Long duration burst noise protection interleaver depths

Control word	Interleaver taps	Interleaver increment	64QAM 5.056941 Msym/s 6 bits per symbol		256QAM 5.360537 Msym/s 8 bits per symbol	
			Burst protection	Latency	Burst protection	Latency
Four bits	I	J				
0000	128	1	95 μ s	4.0 ms	66 μ s	2.8 ms
0010	128	2	190 μ s	8.0 ms	132 μ s	5.6 ms
0100	128	3	285 μ s	12 ms	198 μ s	8.4 ms
0110	128	4	380 μ s	16 ms	264 μ s	11 ms
1000	128	5	475 μ s	20 ms	330 μ s	14 ms
1010	128	6	570 μ s	24 ms	396 μ s	17 ms
1100	128	7	664 μ s	28 ms	462 μ s	20 ms
1110	128	8	759 μ s	32 ms	528 μ s	22 ms

The interleaver depth, which is coded in a 4-bit control word contained in the FEC frame synchronization trailer, always reflects the interleaving in the immediately following frame. In addition, errors are allowed while the interleaver memory is flushed after a change in interleaving is indicated.

Refer to [ITU-T J.83-B] for the control bit specifications required to specify which interleaving mode is used.

6.3.4 Downstream frequency plan

The downstream frequency plan SHOULD comply with the frequency plan being used by the cable system in which it will operate. For example, this might be a harmonic related carrier (HRC); incremental related carrier (IRC), or Standard (STD) North American frequency plan for digital QAM carriers. Operational frequencies MAY include all channels between, and including centre frequencies of 57 MHz to 999 MHz. Operational frequencies MUST include at least 91 MHz to 867 MHz.

6.3.5 DRFI output electrical

EQAMs and CMTSs may be available in two distinct versions:

- Single channel devices that can only generate one RF channel per physical RF port;
- Multiple channel devices capable of generating more than one channel simultaneously per physical RF port. A multiple channel device could be used to generate a single channel; even so, it is still defined as a multiple channel device.

An N -channel per RF port device **MUST** comply with all requirements operating with all N channels on the RF port, and **MUST** comply with all requirements for an N' -channel per RF port device operating with N' channels on the RF port for all even values of N' less than N , and for $N' = 1$. A single channel device **MUST** comply with all requirements for an N -channel device with $N = 1$.

These specifications assume that the DRFI device will be terminated with a 75 ohm load.

If more than one CMTS or EQAM is packaged in a chassis, each CMTS or EQAM **MUST** meet the appropriate parameters and definitions in this Recommendation, regardless of the number of other CMTSs or EQAMs, their location in the chassis, or their configuration.

6.3.5.1 CMTS or EQAM output electrical

A CMTS or EQAM **MUST** output an RF modulated signal with the characteristics defined in Tables 6-3, 6-4 and 6-5. The condition for these requirements is all N combined channels, commanded to the same average power, except for the single channel active phase noise and diagnostic carrier suppression (Table 6-4) requirements.

Table 6-3 – RF output electrical requirements

Parameter	Value
Centre Frequency (f_c) of any RF channel of a CMTS or EQAM	MAY be 57 MHz to 999 MHz ± 30 kHz (Note 1) MUST be at least 91 MHz to 867 MHz ± 30 kHz
Level	Adjustable. See Table 6-4
Modulation Type	64QAM, 256QAM
Symbol Rate (nominal) 64QAM 256QAM	5.056941 Msym/s 5.360537 Msym/s
Nominal Channel Spacing	6 MHz
Frequency response 64QAM 256QAM	~ 0.18 Square Root Raised Cosine Shaping ~ 0.12 Square Root Raised Cosine Shaping
Inband Spurious, Distortion and Noise	Unequalized MER (Note 2) > 35 dB Equalized MER > 43 dB
Inband Spurious and Noise	≤ -48 dBc; where channel spurious and noise includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, and other undesired transmitter products. Spurious and noise within ± 50 kHz of the carrier is excluded. When $N > 1$, noise outside the Nyquist bandwidth is excluded.
Out-of-Band Spurious and Noise	See Table 6-5

Table 6-3 – RF output electrical requirements

Parameter	Value
Phase Noise Single Channel Active, $N - 1$ Channels Suppressed (see clause 6.3.5.1.2 (6)) 64QAM and 256QAM	1 kHz-10 kHz: -33 dBc double sided noise power 10 kHz-50 kHz: -51 dBc double sided noise power 50 kHz-3 MHz: -51 dBc double sided noise power
All N Channels Active, (see clause 6.3.5.1.2 (7)) 64QAM and 256QAM	1 kHz-10 kHz: -33 dBc double sided noise power 10 kHz-50 kHz: -51 dBc double sided noise power
Output Impedance	75 ohms
Output Return Loss (Note 3)	> 14 dB within an active output channel from 88 MHz to 750 MHz (Note 4) > 13 dB within an active output channel from 750 MHz to 870 MHz (Note 5) > 12 dB in every inactive channel from 54 MHz to 870 MHz > 10 dB in every inactive channel from 870 MHz to 1002 MHz
Connector	F connector per [IEC 60169-24]
<p>NOTE 1 – 30 kHz includes an allowance of 25 kHz for the largest frequency offset normally built into upconverters.</p> <p>NOTE 2 – MER (modulation error ratio) is determined by the cluster variance caused by the transmit waveform at the output of the ideal receive matched filter. MER includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, distortion, and other undesired transmitter products. Unequalized MER also includes linear filtering distortion, which is compensated by a receive equalizer. Phase noise up to ± 50 kHz of the carrier is excluded from inband specification, to separate the phase noise and inband spurious requirements as much as possible. In measuring MER, record length or carrier tracking loop bandwidth may be adjusted to exclude low frequency phase noise from the measurement. For equalized MER, receive equalizer coefficients are computed and applied with receiver operating with device under test. For unequalized MER, receive equalize coefficients may be computed to flatten receiver response, if necessary, then are held fixed when device under test is connected. MER requirements assume measuring with a calibrated test instrument with its residual MER contribution removed.</p> <p>NOTE 3 – Frequency ranges are edge-to-edge.</p> <p>NOTE 4 – If the EQAM or CMTS provides service to a centre frequency of 57 MHz (see line 1 in the table), then the EQAM or CMTS MUST provide a return loss of > 14 dB within an active output channel, from 54 MHz to 750 MHz (f_{edge}).</p> <p>NOTE 5 – If the EQAM or CMTS provides service to a centre frequency of 999 MHz (see line 1 in the table), then the EQAM or CMTS MUST provide a return loss of > 12 dB within an active output channel, from 870 MHz to 1002 MHz (f_{edge}).</p>	

6.3.5.1.1 Power per channel CMTS or EQAM

An EQAM or CMTS MUST generate an RF output with power capabilities as defined in Table 6-4. Channel RF power MAY be adjustable on a per channel basis with each channel independently meeting the power capabilities defined in Table 6-4. If the EQAM or CMTS has independent modulation capability on a per channel basis, then the channel RF power MUST be adjustable on a per channel basis, with each channel independently meeting the power capabilities defined in Table 6-4.

Table 6-4 – DRFI device output power

Parameter	Value
Range of commanded transmit power per channel	≥ 8 dB below required power level specified below maintaining full fidelity over the 8 dB range.
Commanded power per channel step size	≤ 0.2 dB strictly monotonic
Power difference between any two adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 0.5 dB
Power difference between any two non-adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 1 dB
Power per channel absolute accuracy	± 2 dB
Diagnostic carrier suppression (3 modes) Mode 1: One channel suppressed Mode 2: All channels suppressed except one Mode 3: All channels suppressed	<ol style="list-style-type: none"> 1) ≥ 50 dB carrier suppression within the Nyquist bandwidth in any one 6 MHz channel in the block. This MUST be accomplished without discontinuity or detriment to the other channels in the block. 2) 50 dB carrier suppression within the Nyquist bandwidth in every 6 MHz channel in the block except one. This MUST be accomplished without discontinuity or detriment to the remaining channel in the block. 3) 50 dB carrier suppression within the Nyquist bandwidth in every 6 MHz channel in the block.
RF block muting	≥ 73 dB below the unmuted aggregate power of the block, in every 6 MHz channel in the block
Required power per channel for N channels combined onto a single RF port. ' N ' = number of combined channels: $N = 1$ $N = 2$ $N = 3$ $N = 4$ $N > 4$	Required power in dBmV per channel 60 dBmV 56 dBmV 54 dBmV 52 dBmV $60 - \text{ceil}[3.6 * \log_2(N)]$ dBmV

6.3.5.1.2 Independence of individual channel within the multiple channels on a single RF port

A potential use of a CMTS or an EQAM is to provide a universal platform that can be used for high-speed data services or for video services. For this reason, it is essential that interleaver depth be set on a per channel basis to provide a suitable transmission format for either video or data as needed in normal operation. Any N -channel block of a CMTS or EQAM MUST be configurable with at least two different interleaver depths, using any of the interleaver depths shown in Tables 6-1 and 6-2. Although not as critical as per-channel interleaver depth control, there are strong benefits for the operator if the EQAM is provided with the ability to set RF power, centre frequency, and modulation type on a per-channel basis.

- 1) A multiple-channel CMTS or EQAM MUST be configurable with at least two different interleaver depths among the N channels on an RF output port, with each channel using one of the two (or more) interleaver depths, on a per channel basis, see Tables 6-1 and 6-2 for information on interleaver depths.

- 2) A multiple-channel CMTS or EQAM MUST provide for 3 modes of carrier suppression of RF power for diagnostic and test purposes, see Table 6-4 for mode descriptions and carrier RF power suppression level.
- 3) A multiple-channel CMTS or EQAM MAY provide for independent adjustment of RF power in a per channel basis with each RF carrier independently meeting the requirements defined in Table 6-4.
- 4) A multiple-channel CMTS or EQAM MAY provide for independent selection of centre frequency on a per channel basis, thus providing for non-contiguous channel frequency assignment with each channel independently meeting the requirements in Table 6-3.
- 5) A multiple-channel CMTS or EQAM MAY provide for independent selection of modulation order, either 64QAM or 256QAM, on a per channel basis, with each channel independently meeting the requirements in Table 6-3.
- 6) A CMTS or EQAM MUST provide a test mode of operation, for out-of-service testing, configured for N channels but generating one-CW-per-channel, one channel at a time at the centre frequency of the selected channel; all other combined channels are suppressed. One purpose for this test mode is to support one method for testing the phase noise requirements of Table 6-3. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance; for example, repeated selection of a constellation symbol with power close to the constellation RMS level would seemingly exercise much of the modulation and up-conversion chain in a realistic manner. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table 6-3.
- 7) A CMTS or EQAM MUST provide a test mode of operation, for out-of-service testing, generating one-CW-per-channel, at the centre frequency of the selected channel, with all other $N - 1$ of the combined channels active and containing valid data modulation at operational power levels. One purpose for this test mode is to support one method for testing the phase noise requirements of Table 6-3. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance. For example, a repeated selection of a constellation symbol, with power close to the constellation RMS level, would seemingly exercise much of the modulation and upconversion chain in a realistic manner. For this test mode, it is acceptable that all channels operate at the same average power, including each of the $N - 1$ channels in valid operation, and the single channel with a CW tone at its centre frequency. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table 6-3.

If either centre frequency 4) or modulation type 5), or both are independently adjustable on a per channel basis, then the CMTS or EQAM MUST provide for independent adjustment of RF power 3) on a per channel basis, with each RF carrier independently meeting the requirements defined in Table 6-3.

6.3.5.1.3 Out-of-band noise and spurious requirements for CMTS or EQAM

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 60 dB for systems deploying up to 119 DRFI compliant QAM channels.

This Recommendation assumes that the transmitted power level of the digital channels will be 6 dB below the peak envelope power of the visual signal of analog channels, which is the typical condition for 256 QAM transmission. It is further assumed that the channel lineup will place analog channels at lower frequencies than digital channels. An adjustment of $10 \cdot \log_{10}(6 \text{ MHz}/4 \text{ MHz})$ is used to account for the difference in noise bandwidth of digital channels, versus analog channels.

With the assumptions above, for a 119 QAM channel system, the specification in item 5 of Table 6-5 equates to an analog CNR protection of 60 dB.

Table 6-5 lists the out-of-band spurious requirements. In cases where the N combined channels are not commanded to the same power level, "dBc" denotes decibels relative to the strongest carrier in the channel block. The out-of-band spurious emission requirements assume a test condition with a contiguous block of N combined channels commanded to the same power level, and for this test condition "dBc" should be interpreted as the average channel power, averaged over the block, to mitigate the variation in channel power across the block (see Table 6-4) which is allowed with all channels commanded to the same power.

Items 1 through 4 list the requirements in channels adjacent to the commanded channels.

Item 5 lists the requirements in all other channels further from the commanded channels. Some of these "other" channels are allowed to be excluded from meeting the Item 5 specification. All the exclusions, such as 2nd and 3rd harmonics of the commanded channel, are fully identified in the table.

Item 6 lists the requirements on the $2N$ 2nd harmonic channels and the $3N$ 3rd harmonic channels.

Table 6-5 – EQAM or CMTS output out-of-band noise and spurious emissions requirements

Item	Band	N , number of combined channels per RF port				
		1	2	3	4	$N > 4$ All equations are ceiling (power, 0.5) dBc
1	Adjacent channel up to 750 kHz from channel block edge	< -58 dBc	< -58 dBc	< -58 dBc	< -58 dBc	$< 10 * \log_{10} [10^{-58/10} + (0.75/6) * (10^{-65/10} + (N-2) * 10^{-73/10})]$
2	Adjacent channel (750 kHz from channel block edge to 6 MHz from channel block edge)	< -62 dBc	< -60 dBc	< -60 dBc	< -60 dBc	$< 10 * \log_{10} [10^{-62/10} + (5.25/6) * (10^{-65/10} + (N-2) * 10^{-73/10})]$
3	Next-adjacent channel (6 MHz from channel block edge to 12 MHz from channel block edge)	< -65 dBc	< -64 dBc	< -63.5 dBc	< -63 dBc	$< 10 * \log_{10} [10^{-65/10} + (N-1) * 10^{-73/10}]$
4	Third-adjacent channel (12 MHz from channel block edge to 18 MHz from channel block edge)	< -73 dBc	< -70 dBc	< -67 dBc	< -65 dBc	For $N = 5$: -64.5 dBc; For $N = 6$: -64 dBc; For $N \geq 7$: $< -73 + 10 * \log_{10}(N)$ dBc
5	Noise in other channels (47 MHz to 1000 MHz) Measured in each 6 MHz channel excluding the following: a) Desired channel(s) b) 1st, 2nd and 3rd adjacent channels (see Items 1, 2, 3, 4 in this table) c) Channels coinciding with 2nd and 3rd harmonics (see Item 6 in this table)	< -73 dBc	< -70 dBc	< -68 dBc	< -67 dBc	$< -73 + 10 * \log_{10}(N)$
6	In each of $2N$ contiguous 6 MHz channels or in each of $3N$ contiguous 6 MHz channels coinciding with 2nd harmonic and with 3rd harmonic components respectively (up to 1000 MHz)	$< -73 + 10 * \log_{10}(N)$, or -63 dBc, whichever is greater				

6.3.5.2 CMTS or EQAM master clock fitter for asynchronous operation

An EQAM MUST implement a DTI client and client interface per [ITU-T J.211]. The master clock specifications are defined in [ITU-T J.211]. The DTI client provides the master clock. An integrated CMTS not actively serviced by a DTI server MUST include a master clock with the specifications as follows:

The 10.24 MHz Master Clock MUST have, over a temperature range of 0 to 40 degrees C and for up to ten years from date of manufacture (see Note below):

- A frequency accuracy of $\leq \pm 5$ ppm;
- A drift rate $\leq 10^{-8}$ per second; and
- An edge jitter of ≤ 10 ns peak-to-peak (± 5 ns).

NOTE – This specification MAY also be met by synchronizing the DRFI device master clock oscillator to an external frequency reference source. If this approach is used, the internal DRFI device master clock MUST have a frequency accuracy of ± 20 ppm over a temperature range of 0 to 40 degrees C, up to 10 years from date of manufacture, when no frequency reference source is connected. The drift rate and edge jitter MUST be as specified above.

The drift rate and jitter requirements on the DRFI device Master Clock implies that the duration of two adjacent segments of 10'240'000 cycles will be within 30 ns, due to 10 ns jitter on each segments' duration, and 10 ns due to frequency drift. Durations of other counter lengths also may be deduced: adjacent 1'024'000 segments, ≤ 21 ns; 1'024'000 length segments separated by one 10'240'000-cycle segment, ≤ 30 ns; adjacent 102'400'000 segments, ≤ 120 ns. The DRFI device master clock MUST meet such test limits in 99% or more measurements.

6.3.5.3 CMTS or EQAM master clock jitter for synchronous operation

In addition to the requirements in clause 6.3.5.2, the 10.24 MHz CMTS master clock MUST meet the following double sideband phase noise requirements over the specified frequency ranges:

- $< [-50 + 20 \cdot \log(f_{MC}/10.24)]$ dBc (*i.e.*, < 0.05 ns RMS) 10 Hz to 100 Hz
- $< [-58 + 20 \cdot \log(f_{MC}/10.24)]$ dBc (*i.e.*, < 0.02 ns RMS) 100 Hz to 1 kHz
- $< [-50 + 20 \cdot \log(f_{MC}/10.24)]$ dBc (*i.e.*, < 0.05 ns RMS) 1 kHz to 10 kHz
- $< [-50 + 20 \cdot \log(f_{MC}/10.24)]$ dBc (*i.e.*, < 0.05 ns RMS) 10 kHz to $f_{MC}/2$

f_{MC} is the frequency of the measured master clock in MHz. The value of f_{MC} MUST be either an integral multiple or divisor of 10.24 MHz. For example, if a 20.48 MHz oscillator is used as the master clock frequency source, and there is no explicit 10.24 MHz clock to test, the 20.48 MHz clock may be used with f_{MC} equal to 20.48 in the above expressions.

Specifications for EQAM master clock jitter in synchronous operation are contained in [ITU-T J.211].

6.3.5.4 CMTS or EQAM master clock frequency drift for synchronous operation

The frequency of the CMTS master clock MUST NOT drift more than 10^{-8} per second.

Specifications for EQAM master clock frequency drift in synchronous operation is contained in [ITU-T J.211].

6.3.6 CMTS or EQAM clock generation

When the 10.24 MHz master clock is provided by the DTI interface, a DRFI-compliant device MUST lock the downstream symbol clock to the 10.24 MHz master clock using the M/N divisors provided in Table 6-6.

6.3.6.1 CMTS clock generation

The CMTS MUST lock the downstream symbol clock to the CMTS master clock using the M/N divisors provided in Table 6-6.

6.3.6.2 EQAM clock generation

Because it operates with an active DTI interface, an EQAM MUST lock the downstream symbol clock to the master clock using the M/N divisors provided in Table 6-6.

6.3.6.3 Downstream symbol rate

Let f'_b represent the rate of the downstream symbol clock which is locked to the master clock and let f'_m represent the rate of the master clock locked to the downstream symbol clock. Let f_b represent the nominal specified downstream symbol rate and let f_m represent the nominal master clock rate (10.24 MHz). With the downstream symbol clock locked to the master clock, the following equation MUST hold:

$$f'_b = f_m * M/N$$

With the master clock locked to the downstream symbol clock, the following equation MUST hold:

$$f'_m = f_b * N/M$$

Note that M and N in Table 6-6 are unsigned integer values, each representable in 16 bits and result in a value of f'_b or f'_m that is not more than ± 1 ppm from its specified nominal value.

The standard deviation of the timing error of the EQAM/CMTS RF symbol clock, referenced to the DTI server master clock, MUST be less than 1.5 ns measured over 100 seconds.

Table 6-6 lists the downstream modes of operation, their associated nominal symbol rates, f_b , values for M and N, the resulting synchronized clock rates and their offsets from their nominal values.

Table 6-6 – Downstream symbol rates & parameters for synchronization with master clock

Downstream mode	Nominal specified symbol rate, f_b (MHz)	M/N	Master clock rate, f'_m (MHz)	Downstream symbol rate, f'_b (MHz)	Offset from nominal
Annex B, 64QAM	5.056941	401/812	10.239990...	5.056945...	0.95 ppm
Annex B, 256QAM	5.360537	78/149	10.240000...	5.360536...	0.02 ppm

6.3.7 Downstream symbol clock jitter for synchronous operation

The downstream symbol clock MUST meet the following double sideband phase noise requirements over the specified frequency ranges:

- $< [-53 + 20 * \log (f_{DS}/5.057)]$ dBc (*i.e.*, < 0.07 ns RMS) 10 Hz to 100 Hz
- $< [-53 + 20 * \log (f_{DS}/5.057)]$ dBc (*i.e.*, < 0.07 ns RMS) 100 Hz to 1 kHz
- $< [-53 + 20 * \log (f_{DS}/5.057)]$ dBc (*i.e.*, < 0.07 ns RMS) 1 kHz to 10 kHz
- $< [-36 + 20 * \log (f_{DS}/5.057)]$ dBc (*i.e.*, < 0.5 ns RMS) 10 kHz to 100 kHz
- $< [-30 + 20 * \log (f_{DS}/5.057)]$ dBc (*i.e.*, < 1 ns RMS) 100 kHz to $(f_{DS} / 2)$

f_{DS} is the frequency of the measured clock in MHz. The value of f_{DS} MUST be an integral multiple or divisor of the downstream symbol clock. For example, an $f_{DS} = 20.227764$ MHz clock may be measured if there is no explicit 5.056941 MHz clock available.

A DRFI-compliant device MUST provide a means for clock testing in which:

- The device provides test points for direct access to the master clock and the downstream symbol clock.

Alternatively, a DRFI conformant device MUST provide a test mode in which:

- The downstream QAM symbol sequence is replaced with an alternating binary sequence (1, -1, 1, -1, 1, -1...) at nominal amplitude, on both I and Q.
- The device generates the downstream symbol clock from the 10.24 MHz reference clock as in normal synchronous operation.

If an explicit downstream symbol clock, which is capable of meeting the above phase noise requirements, is available (e.g., a smooth clock without clock domain jitter), this test mode is not required.

6.3.8 Downstream symbol clock drift for synchronous operation

The frequency of the downstream symbol clock MUST NOT drift more than 10^{-8} per second.

7 Downstream transmission convergence sublayer

7.1 Introduction

The downstream transmission convergence layer used in the M-CMTS is defined as a continuous series of 188-byte MPEG [ITU-T H.222.0] packets. These packets consist of a 4-byte header followed by 184 bytes of payload. The header identifies the payload as belonging to the data-over-cable MAC. Other values of the header may indicate other payloads. The mixture of MAC payloads and those of other services is optional and is controlled by the CMTS.

Figure 7-1 illustrates the interleaving of DOCSIS MAC bytes with other digital information (digital video in the example shown).

header=DOC	DOC MAC payload
header=video	digital video payload
header=video	digital video payload
header=DOC	DOC MAC payload
header=video	digital video payload
header=DOC	DOC MAC payload
header=video	digital video payload
header=video	digital video payload
header=video	digital video payload

Figure 7-1 – Example of interleaving MPEG packets in downstream

7.2 MPEG packet format

The format of an MPEG packet carrying DOCSIS data is shown in Figure 7-2. The packet consists of a 4-byte MPEG header, a pointer_field (not present in all packets), and the DOCSIS payload.

MPEG header (4 bytes)	pointer_field (1 byte)	DOCSIS payload (183 or 184 bytes)
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Figure 7-2 – Format of an MPEG packet

7.3 MPEG header for DOCSIS data-over-cable

The format of the MPEG transport stream header is defined in clause 2.4 of [ITU-T H.222.0]. The particular field values that distinguish data-over-cable MAC streams are defined in Table 7-1. Field names are from [ITU-T H.222.0].

The MPEG header consists of 4 bytes that begin the 188-byte MPEG packet. The format of the header for use on a DOCSIS data-over-cable PID MUST be as shown in Table 7-1. The header format conforms to the MPEG standard, but its use is restricted in this Recommendation to not allow inclusion of an adaptation_field in the MPEG packets.

Table 7-1 – MPEG header format for DOCSIS data-over-cable packets

Field	Length (bits)	Description
sync_byte	8	0x47; MPEG Packet Sync byte
transport_error_indicator	1	Indicates that an error has occurred in the reception of the packet. This bit is reset to zero by the sender, and set to one whenever an error occurs in the transmission of the packet
payload_unit_start_indicator	1	A value of one indicates the presence of a pointer_field as the first byte of the payload (fifth byte of the packet)
transport_priority	1	Reserved; set to zero
PID	13	DOCSIS Data-Over-Cable well-known PID (0x1FFE)
transport_scrambling_control	2	Reserved, set to '00'
adaptation_field_control	2	'01'; use of the adaptation_field is not allowed on the DOCSIS PID
continuity_counter	4	cyclic counter within this PID

7.4 MPEG payload for DOCSIS data-over-cable

The MPEG payload portion of the MPEG packet will carry the DOCSIS MAC frames. The first byte of the MPEG payload will be a 'pointer_field' if the payload_unit_start_indicator (PUSI) of the MPEG header is set.

7.4.1 stuff_byte

This standard defines a stuff_byte pattern having a value (0xFF) that is used within the DOCSIS payload to fill any gaps between the DOCSIS MAC frames. This value is chosen as an unused value for the first byte of the DOCSIS MAC frame. The 'FC' byte of the MAC header will be defined to never contain this value. (FC_TYPE = '11' indicates a MAC-specific frame, and FC_PARM = '11111' is not currently used and, according to this Recommendation, is defined as an illegal value for FC_PARM.)

7.4.2 pointer_field

The pointer_field is present as the fifth byte of the MPEG packet (first byte following the MPEG header) whenever the PUSI is set to one in the MPEG header. The interpretation of the pointer_field is as follows:

The pointer_field contains the number of bytes in this packet that immediately follow the pointer_field that the CM decoder will skip past before looking for the beginning of a DOCSIS MAC Frame. A pointer field MUST be present if it is possible to begin a data-over-cable MAC frame in the packet, and MUST point to either:

- 1) the beginning of the first MAC frame to start in the packet; or
- 2) to any stuff_byte preceding the MAC frame.

7.5 Interaction with the MAC sublayer

MAC frames may begin anywhere within an MPEG packet. MAC frames may span MPEG packets, and several MAC frames may exist within an MPEG packet.

The following figures show the format of the MPEG packets that carry DOCSIS MAC frames. In all cases, the PUSI flag indicates the presence of the pointer_field as the first byte of the MPEG payload.

Figure 7-3 shows a MAC frame that is positioned immediately after the pointer_field byte. In this case, the pointer_field is zero, and the DOCSIS decoder will begin searching for a valid FC byte, at the byte immediately following the pointer_field.

MPEG header (PUSI = 1)	pointer_field (= 0)	MAC frame (up to 183 bytes)	stuff_byte(s) (0 or more)
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Figure 7-3 – Packet format where a MAC frame immediately follows the pointer_field

Figure 7-4 shows the more general case, where a MAC frame is preceded by the tail of a previous MAC frame and a sequence of stuffing bytes. In this case, the pointer_field still identifies the first byte after the tail of Frame #1 (a stuff_byte) as the position where the decoder should begin searching for a legal MAC sublayer FC value. This format allows the multiplexing operation in the CMTS to immediately insert a MAC frame that is available for transmission if that frame arrives after the MPEG header and pointer_field have been transmitted.

In order to facilitate multiplexing of the MPEG packet stream carrying DOCSIS data with other MPEG-encoded data, the CMTS SHOULD NOT transmit MPEG packets with the DOCSIS PID which contain only stuff_bytes in the payload area. MPEG null packets SHOULD be transmitted instead.

NOTE – There are timing relationships implicit in the DOCSIS MAC sublayer which must also be preserved by any MPEG multiplexing operation.

MPEG header (PUSI = 1)	pointer_field (= M)	Tail of MAC frame #1 (M bytes)	stuff_byte(s) (0 or more)	Start of MAC frame #2
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Figure 7-4 – Packet format with MAC frame preceded by stuffing bytes

Figure 7-5 shows that multiple MAC frames may be contained within the MPEG packet. The MAC frames may be concatenated one after the other or be separated by an optional sequence of stuffing bytes.

MPEG header (PUSI = 1)	pointer_field (= 0)	MAC frame #1	MAC frame #2	stuff_byte(s) (0 or more)	MAC frame #3
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Figure 7-5 – Packet format showing multiple MAC frames in a single packet

Figure 7-6 shows the case where a MAC frame spans multiple MPEG packets. In this case, the pointer_field of the succeeding frame points to the byte following the last byte of the tail of the first frame.

MPEG header (PUSI = 1)	pointer_field (= 0)	stuff_byte(s) (0 or more)	Start of MAC frame #1 (up to 183 bytes)
MPEG header (PUSI = 0)	Continuation of MAC frame #1 (184 bytes)		
MPEG header (PUSI = 1)	pointer_field (= M)	Tail of MAC frame #1 (M bytes)	Start of MAC frame #2 (M bytes)

Figure 7-6 – Packet format where a MAC frame spans multiple packets

The transmission convergence sublayer must operate closely with the MAC sublayer in providing an accurate timestamp to be inserted into the time synchronization message.

7.6 Interaction with the physical layer

The MPEG-2 packet stream MUST be encoded according to [ITU-T J.83-B], including MPEG-2 transport framing using a parity checksum as described in [ITU-T J.83-B].

Annex A

Additions and modifications for European specification

(This annex forms an integral part of this Recommendation)

This annex applies to the second technology option referred to in clause 1. For the first option, refer to clauses 5, 6 and 7.

This annex describes the physical layer specifications required for the EuroDOCSIS integrated CMTS and EuroDOCSIS EQAM. This is an optional annex and in no way affects certification of equipment adhering to the North American technology option described in the clauses referenced above.

The numbering of the paragraphs has been maintained such that the suffix after the letter for the annex refers to the part of the Recommendation where the described changes apply. As a consequence, some heading numbers might be missing in this annex, since no change is required to the relevant paragraph in the main body of the Recommendation.

A.1 Scope and purpose

See clause 1.

A.2 References

See clause 2.

A.3 Terms and definitions

See clause 3.

A.4 Acronyms and abbreviations

See clause 4.

A.5 Functional assumptions

This clause describes the characteristics of a cable television plant, assumed to be for the purpose of operating a data-over-cable system. It is not a description of EQAM or CMTS parameters. The data-over-cable system **MUST** be interoperable within the environment described in this clause.

Whenever there is a reference to frequency plans or to compatibility with other services in this clause, conflicts with a legal requirement for the area of operation, the latter shall take precedence. Any reference to analog TV signals in a particular frequency band does not imply that such signals are physically present.

A.5.1 Broadband access network

A coaxial-based broadband access network is assumed. This may take the form of either an all-coax or hybrid fibre/coax (HFC) network. The generic term "cable network" is used here to cover all cases.

A cable network uses a shared-medium, "tree-and-branch" architecture, with analog transmission. The key functional characteristics assumed in this annex are the following:

- Two-way transmission.
- A maximum optical/electrical spacing between the DRFI-compliant device and the most distant CM of 160 km (route meters) in each direction.

- A maximum differential optical/electrical spacing between the DRFI-compliant device and the closest and most distant modems of 160 km (route meters) in each direction.

At a propagation velocity in fibre of approximately 5 ns/m, 160 km of fibre in each direction results in a round-trip delay of approximately 1.6 ms. For further information, see Appendix VIII of [ITU-T J.122].

A.5.2 Equipment assumptions

A.5.2.1 Frequency plan

In the downstream direction, the cable system is assumed to have a passband with a typical lower edge between 47 and 87.5 MHz, and an upper edge that is implementation-dependent, but is typically in the range of 300 to 862 MHz. Within that passband, PAL/SECAM analog television signals in 7/8 MHz channels and FM radio signals are assumed to be present, as well as other narrow-band and wideband digital signals. 8 MHz channels are used for data communication.

A.5.2.2 Compatibility with other services

The CM and EQAM or CMTS MUST coexist with the other services on the cable network, for example:

– *Annex A*

They MUST be interoperable in the cable spectrum assigned for EQAM- or CMTS-CM interoperation while the balance of the cable spectrum is occupied by any combination of television and other signals; and

– *Annex B*

They MUST NOT cause harmful interference to any other services that are assigned to the cable network in spectrum outside of that allocated to the EQAM or CMTS. The latter is understood as:

- No measurable degradation (highest level of compatibility);
- No degradation below the perceptible level of impairments for all services (standard or medium level of compatibility); or
- No degradation below the minimal standards accepted by the industry or other service provider (minimal level of compatibility).

A.5.2.3 Fault isolation impact on other users

See clause 5.2.3.

A.5.3 Downstream plant assumptions

See clause 5.3.

A.5.3.1 Transmission levels

The nominal average power level of the downstream RF signal(s) within an 8 MHz channel is targeted to be in the range of –13 dBc to 0 dBc, relative to analog peak video carrier level and will normally not exceed analog peak video carrier level (typically between –10 to –6 dBc for 64QAM, and between –6 to –4 dBc for 256QAM).

A.5.3.2 Frequency inversion

See clause 5.3.2.

A.5.3.3 Analog and digital channel line-up

In developing this technology option, it was assumed that a maximum of 85 digital channels would be deployed in a headend. For the purposes of calculating CNR protection for analog channels, it

was assumed that analog channels are placed at lower frequencies in the channel line-up than digital channels.

A.5.3.4 Analog protection goal

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 59 dB measured in a 5.08 MHz wide frequency band for systems deploying up to 85 DRFI compliant QAM channels.

For purposes of calculation, it is assumed that the transmitted power level of the digital channels will be 5 dB below the peak envelope power of the visual signal of analog channels, which is within the range of typical conditions for 256QAM transmission. It is further assumed, for the purpose of calculation, that the channel line-up will place analog channels at lower frequencies than digital channels. An adjustment of $10 \cdot \log_{10}(8 \text{ MHz}/5.08 \text{ MHz})$ is used to account for the difference in bandwidth used to define the noise requirements for DRFI-compliant digital QAM channels, versus analog PAL channels. With the assumptions above, for an 85 QAM channel system, the specification in item 5 of Table A.4 equates to an analog CNR protection of 59 dB.

A.6 Physical media dependent sublayer specification

A.6.1 Scope

This clause applies to the second technology option referred to in clause 1 (1.1 Scope). In cases where the requirements for both technology options are identical, a reference is provided to the main text.

For the remainder of this clause, see clause 6.1.

A.6.2 EdgeQAM (EQAM) differences from CMTS

See clause 6.2.

A.6.3 Downstream

A.6.3.1 Downstream protocol

The downstream PMD sublayer MUST conform to [ETSI EN 300 429].

A.6.3.2 Spectrum format

The downstream modulator for each QAM channel of the EQAM or CMTS MUST provide operation with an RF signal format transmitted in the cable system whose baseband quadrature components form a constellation diagram, as specified in [ETSI EN 300 429].

A.6.3.3 Scaleable interleaving to support video and high-speed data services

The CMTS or EQAM downstream PMD sublayer MUST support the interleaver with the characteristics defined in Table A.1. This interleaver mode fully complies with [ETSI EN 300 429].

Table A.1 – Interleaver characteristics

Interleaver taps	Interleaver increment	64QAM 6.952 Msym/s 6 bits per symbol		256QAM 6.952 Msym/s 8 bits per symbol	
		Burst protection	Latency	Burst protection	Latency
I	J				
12	17	18 μ s	0.43 ms	14 μ s	0.32 ms

A.6.3.4 Downstream frequency plan

It is up to the operator to decide which frequencies to use to meet national and network requirements.

A.6.3.5 DRFI output electrical

See clause 6.3.5.

A.6.3.5.1 CMTS or EQAM output electrical

A CMTS or EQAM MUST output an RF modulated signal with the characteristics defined in Tables A.2, A.3 and A.4.

A.6.3.5.1.1 Output electrical per RF port

The condition for the requirements in Table A.2 is that all N channels delivered to the RF port are commanded to the same average power. That condition does not apply to the requirement on single channel active phase noise.

Table A.2 – Output electrical requirements per RF port

Parameter	Value
Centre Frequency (f_c) of any RF channel of a CMTS or EQAM	MAY be 85 MHz to 999 MHz ± 30 kHz at 250 kHz increments MUST be 112 MHz to 858 MHz ± 30 kHz at 250 kHz increments
Level	Adjustable. See Table A.3
Modulation Type	64QAM, 256QAM
Symbol Rate (nominal) 64QAM 256QAM	6.952 Msym/s 6.952 Msym/s
Nominal Channel Spacing	8 MHz
Frequency response 64QAM 256QAM	~ 0.15 Square Root Raised Cosine Shaping ~ 0.15 Square Root Raised Cosine Shaping
Inband Spurious, Distortion and Noise	Unequalized MER (Note 1) > 35 dB Equalized MER > 43 dB
Inband Spurious and Noise ($f_c \pm 4$ MHz)	≤ -46.7 dBc; where channel spurious and noise includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, and other undesired transmitter products. Spurious and noise within ± 50 kHz of the carrier is excluded. When $N > 1$, noise outside the Nyquist bandwidth is excluded.
Out-of-Band Spurious and Noise	See Table A.4
Phase Noise Single Channel Active, $N - 1$ Channels Suppressed (see clause 6.3.5.1.2 (6)), 64QAM and 256QAM	1 kHz – 10 kHz: –33 dBc double sided noise power 10 kHz – 50 kHz: –51 dBc double sided noise power 50 kHz – 3 MHz: –51 dBc double sided noise power
All N Channels Active (see clause 6.3.5.1.2 (7)), 64QAM and 256QAM	1 kHz – 10 kHz: –33 dBc double sided noise power 10 kHz – 50 kHz: –51 dBc double sided noise power
Output Impedance	75 ohms

Table A.2 – Output electrical requirements per RF port

Parameter	Value
Output Return Loss	> 14 dB within an active output channel in the frequency range from 108 MHz to 862 MHz (Note 2) > 12 dB in every inactive channel from 81 MHz to 862 MHz > 10 dB in every inactive channel above 862 MHz
Connector	F connector per [IEC 60169-24]
<p>NOTE 1 – MER (modulation error ratio) is determined by the cluster variance caused by the transmit waveform at the output of the ideal receive matched filter. MER includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, distortion, and other undesired transmitter products. Unequalized MER also includes linear filtering distortion, which is compensated by a receive equalizer. Phase noise up to ± 50 kHz of the carrier is excluded from inband specification, to separate the phase noise and inband spurious requirements as much as possible. In measuring MER, record length or carrier tracking loop bandwidth may be adjusted to exclude low frequency phase noise from the measurement. For equalized MER, receive equalizer coefficients are computed and applied with receiver operating with device under test. For unequalized MER, receive equalize coefficients may be computed to flatten receiver response, if necessary, then are held fixed when device under test is connected. MER requirements assume measuring with a calibrated test instrument with its residual MER contribution removed.</p> <p>NOTE 2 – If the EQAM or CMTS provides service to a centre frequency of 85 MHz (see line 1 in the table) and above, then the EQAM or CMTS MUST provide a return loss > 14 dB within an active output channel in the frequency range from 81 MHz to 108 MHz. If the EQAM or CMTS provides service to a centre frequency of 999 MHz (see line 1 in the table) and below, then the EQAM or CMTS MUST provide a return loss > 14 dB within an active output channel in the frequency range from 862 MHz to 1003 MHz.</p>	

A.6.3.5.1.2 Power per channel CMTS or EQAM

An EQAM or CMTS MUST generate an RF output with power capabilities as defined in Table A.3. Channel RF power MAY be adjustable on a per channel basis with each channel independently meeting the power capabilities defined in Table A.3. If the EQAM or CMTS has independent modulation capability on a per channel basis, then the channel RF power MUST be adjustable on a per channel basis with each channel independently meeting the power capabilities defined in Table A.3.

Table A.3 – DRFI device output power

Parameter	Value
Range of commanded transmit power per channel	≥ 8 dB below required power level specified below maintaining full fidelity over the 8 dB range.
Commanded power per channel step size	≤ 0.2 dB strictly monotonic
Power difference between any two adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 0.5 dB
Power difference between any two non-adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 1 dB
Power per channel absolute accuracy	± 2 dB

Table A.3 – DRFI device output power

Parameter	Value
Diagnostic carrier suppression (3 modes) Mode 1: One channel suppressed Mode 2: All channels suppressed except one Mode 3: All channels suppressed	1) ≥ 50 dB carrier suppression within the Nyquist bandwidth in any one 8 MHz channel in the block. This MUST be accomplished without discontinuity or detriment to the other channels in the block. 2) ≥ 50 dB carrier suppression within the Nyquist bandwidth in every 8 MHz channel in the block except one. This MUST be accomplished without discontinuity or detriment to the remaining channel in the block. 3) ≥ 50 dB carrier suppression within the Nyquist bandwidth in every 8 MHz channel in the block.
RF block muting	≥ 71.5 dB below the unmuted aggregate power of the block, in every 8 MHz channel in the block
Required power per channel for N channels combined onto a single RF port: $N = 1$ $N = 2$ $N = 3$ $N = 4$ $N > 4$	Required power in dBmV per channel 60 dBmV 56 dBmV 54 dBmV 52 dBmV $60 - \text{ceil} [3.6 * \log_2(N)]$ dBmV

A.6.3.5.1.3 Independence of individual channel within the multiple channels on a single RF port

A potential use of a CMTS or an EQAM is to provide a universal platform that can be used for high-speed data services or for video services. There are strong benefits for the operator if the multiple-channel CMTS or the EQAM is provided with the ability to set RF power, centre frequency and modulation type on a per-channel basis.

- 1) A multiple-channel CMTS or EQAM MUST provide for 3 modes of carrier suppression of RF power for diagnostic and test purposes. See Table A.3 for mode descriptions and carrier RF power suppression levels.
- 2) A multiple-channel CMTS or EQAM MAY provide for independent adjustment of RF power in a per channel basis with each RF carrier independently meeting the requirements defined in Table A.3.
- 3) A multiple-channel CMTS or EQAM MAY provide for independent selection of centre frequency on a per channel basis, thus providing for non-contiguous channel frequency assignment, with each channel independently meeting the requirements in Table A.3.
- 4) A multiple-channel CMTS or EQAM MAY provide for independent selection of modulation order, either 64QAM or 256QAM, on a per channel basis, with each channel independently meeting the requirements in Table A.2.
- 5) A CMTS or EQAM MUST provide a test mode of operation, for out-of-service testing, configured for N channels, but generating one-CW-per-channel, one channel at a time at the centre frequency of the selected channel; all other of the combined channels are suppressed. One purpose for this test mode is to support one method for testing the phase noise requirements of Table A.2. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance; for example, repeated

selection of a constellation symbol with power close to the constellation RMS level would seemingly exercise much of the modulation and up-conversion chain in a realistic manner. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table A.2.

- 6) A CMTS or EQAM MUST provide a test mode of operation for out-of-service testing, generating one-CW-per-channel, at the centre frequency of the selected channel, with all other $N - 1$ of the combined channels active and containing valid data modulation at operational power levels. One purpose for this test mode is to support one method for testing the phase noise requirements of Table A.2. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance. For example, a repeated selection of a constellation symbol, with power close to the constellation RMS level, would seemingly exercise much of the modulation and up conversion chain in a realistic manner. For this test mode, it is acceptable that all channels operate at the same average power, including each of the $N - 1$ channels in valid operation, and the single channel with a CW tone at its centre frequency. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table A.2.

If either centre frequency 3) or modulation type 4), or both are independently adjustable on a per channel basis, then the CMTS or EQAM MUST provide for independent adjustment of RF power 2) on a per channel basis, with each RF carrier independently meeting the requirements defined in Table A.2.

A.6.3.5.1.4 Out-of-band noise and spurious requirements for CMTS or EQAM

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 59 dB measured in a 5.08 MHz wide frequency band for systems deploying up to 85 DRFI-compliant QAM channels.

For purposes of calculation, it is assumed that the transmitted power level of the digital channels will be 5 dB below the peak envelope power of the visual signal of analog channels, which is within the range of typical conditions for 256QAM transmission. It is further assumed, for the purpose of calculation, that the channel line-up will place analog channels at lower frequencies than digital channels. An adjustment of $10 \cdot \log_{10}(8 \text{ MHz}/5.08 \text{ MHz})$ is used to account for the difference in bandwidth used to define the noise requirements for DRFI-compliant digital QAM channels, versus analog PAL channels. With the assumptions above, for a 85 QAM channel system, the specification in item 5 of Table A.4 equates to an analog CNR protection of 59 dB.

Table A.4 lists the out-of-band spurious requirements. In cases where the N combined channels are not commanded to the same power level, "dBc" denotes the logarithmic power ratio relative to the strongest carrier in the channel block. The out-of-band spurious emission requirements assume a test condition with a contiguous block of N combined channels commanded to the same power level, and for this test condition "dBc" should be interpreted as the average channel power, averaged over the block, to mitigate the variation (see Table A.3) in channel power across the block, which is allowed with all channels commanded to the same power.

Items 1 through 4 list the requirements in channels adjacent to the commanded channels.

Item 5 lists the requirements in all other channels further from the commanded channels. Some of these "other" channels are allowed to be excluded from meeting the Item 5 specification. All the exclusions, such as 2nd and 3rd harmonics of the commanded channel, are fully identified in the table.

Item 6 lists the requirements on the $2N$ 2nd harmonic channels and the $3N$ 3rd harmonic channels.

Table A.4 – EQAM or CMTS output out-of-band noise and spurious emissions requirements

Item	Band	N, Number of combined channels per RF port				
		1	2	3	4	N > 4 All equations are ceiling (Power, 0.5) dBc
1	Adjacent channel up to 750 kHz from channel block edge	< -58 dBc	< -58 dBc	< -58 dBc	< -58 dBc	$< 10 * \log_{10} [10^{-58/10} + (0.75/8)^* (10^{-63.5/10} + (N-2) * 10^{-71.5/10})]$
2	Adjacent channel (750 kHz from channel block edge to 8 MHz from channel block edge)	< -60.5 dBc	< -59 dBc	< -58.5 dBc	< -58.5 dBc	$< 10 * \log_{10} [10^{-60.5/10} + (7.25/8)^* (10^{-63.5/10} + (N-2) * 10^{-71.5/10})]$
3	Next-adjacent channel (8 MHz from channel block edge to 16 MHz from channel block edge)	< -63.5 dBc	< -63 dBc	< -62.5 dBc	< -62 dBc	$< 10 * \log_{10} [10^{-63.5/10} + (N-1) * 10^{-71.5/10}]$
4	Third-adjacent channel (16 MHz from channel block edge to 24 MHz from channel block edge).	< -71.5 dBc	< -68.5 dBc	< -65.5 dBc	< -64 dBc	For N = 5: < -63 dBc; For N = 6: < -62.5 dBc; For N ≥ 7: < -71.5 + 10*log ₁₀ (N) dBc
5	Noise in other channels (80 MHz to 1003 MHz) Measured in each 8 MHz channel excluding the following: a) Desired channel(s) b) 1st, 2nd and 3rd adjacent channels (see Items 1, 2, 3, 4 in this table) c) Channels coinciding with 2nd and 3rd harmonics (see Item 6 in this table)	< -71.5 dBc	< -68.5 dBc	< -66.5 dBc	< -65.5 dBc	< -71.5 + 10*log ₁₀ (N) dBc
6	In each of 2N contiguous 8 MHz channels or in each of 3N contiguous 8 MHz channels coinciding with 2nd harmonic and with 3rd harmonic components respectively (up to 1000 MHz)	< -71.5 + 10*log ₁₀ (N), or -63 dBc, whichever is greater				

A.6.3.5.2 CMTS or EQAM master clock jitter for asynchronous operation

See clause 6.3.5.2.

A.6.3.5.3 CMTS or EQAM master clock jitter for synchronous operation

See clause 6.3.5.3.

A.6.3.5.4 CMTS or EQAM master clock frequency drift for synchronous operation

See clause 6.3.5.4.

A.6.3.6 CMTS or EQAM clock generation

When the 10.24 MHz master clock is provided by the DTI interface, a DRFI-compliant device MUST lock the downstream symbol clock to the 10.24 MHz master clock using the M/N divisors provided in Table A.5.

A.6.3.6.1 CMTS clock generation

The CMTS MUST lock the downstream symbol clock to the CMTS master clock using the M/N divisors provided in Table A.5.

A.6.3.6.2 EQAM clock generation

Because it operates with an active DTI interface, an EQAM MUST lock the downstream symbol clock to the master clock using the M/N divisors provided in Table A.5.

A.6.3.6.3 Downstream symbol rate

Let f'_b represent the rate of the downstream symbol clock which is locked to the master clock and let f'_m represent the rate of the master clock locked to the downstream symbol clock. Let f_b represent the nominal specified downstream symbol rate and let f_m represent the nominal master clock rate (10.24 MHz). With the downstream symbol clock locked to the master clock, the following equation MUST hold:

$$f'_b = f_m * M/N$$

With the master clock locked to the downstream symbol clock, the following equation MUST hold:

$$f'_m = f_b * N/M$$

Note that M and N in Table A.5 are unsigned integer values, each representable in 16 bits, and result in a value of f'_b or f'_m that is not more than ± 1 ppm from its specified nominal value.

The standard deviation of the timing error of the EQAM/CMTS RF symbol clock, referenced to the DTI server master clock, MUST be less than 1.5 ns measured over 100 seconds.

Table A.5 lists the downstream modes of operation, their associated nominal symbol rates, f_b , values for M and N, the resulting synchronized clock rates and their offsets from their nominal values.

Table A.5 – Downstream symbol rates & parameters for synchronization with the master clock

Downstream mode	Nominal specified symbol rate, f_b (MHz)	M/N	Master clock rate, f'_m (MHz)	Downstream symbol rate, f'_b (MHz)	Offset from nominal
ETSI EN 300 429, 64QAM	6.952	869/1280	10.240...	6.952	0 ppm
ETSI EN 300 429, 256QAM	6.952	869/1280	10.240...	6.952	0 ppm

A.6.3.7 Downstream symbol clock jitter for synchronous operation

The downstream symbol clock MUST meet the following double sideband phase noise requirements over the specified frequency ranges:

- $< [-53 + 20 * \log (f_{DS}/6.952)]$ dBc (i.e., < 0.07 ns RMS) 10 Hz to 100 Hz
- $< [-53 + 20 * \log (f_{DS}/6.952)]$ dBc (i.e., < 0.07 ns RMS) 100 Hz to 1 kHz
- $< [-53 + 20 * \log (f_{DS}/6.952)]$ dBc (i.e., < 0.07 ns RMS) 1 kHz to 10 kHz
- $< [-36 + 20 * \log (f_{DS}/6.952)]$ dBc (i.e., < 0.5 ns RMS) 10 kHz to 100 kHz
- $< [-30 + 20 * \log (f_{DS}/6.952)]$ dBc (i.e., < 1 ns RMS) 100 kHz to $(f_{DS} / 2)$

f_{DS} is the frequency of the measured clock in MHz. The value of f_{DS} MUST be an integral multiple or divisor of the downstream symbol clock. For example, an $f_{DS} = 27.808$ MHz clock may be measured if there is no explicit 6.952 MHz clock available.

A DRFI-compliant device MUST provide a means for clock testing in which:

- The device provides test points for direct access to the master clock and the downstream symbol clock.

Alternatively, a DRFI-conformant device MUST provide a test mode in which:

- The downstream QAM symbol sequence is replaced with an alternating binary sequence (1, -1, 1, -1, 1, -1...) at nominal amplitude, on both I and Q.
- The device generates the downstream symbol clock from the 10.24 MHz reference clock as in normal synchronous operation.

If an explicit downstream symbol clock, which is capable of meeting the above phase noise requirements, is available (e.g., a smooth clock without clock domain jitter), this test mode is not required.

A.6.3.8 Downstream symbol clock drift for synchronous operation

See clause 6.3.8.

A.7 Downstream transmission convergence sublayer

A.7.1 Introduction

See clause 7.1.

A.7.2 MPEG packet format

See clause 7.2.

A.7.3 MPEG header for DOCSIS data-over-cable

See clause 7.3.

A.7.4 MPEG payload for DOCSIS data-over-cable

See clause 7.4.

A.7.5 Interaction with the MAC sublayer

See clause 7.5.

A.7.6 Interaction with the physical layer

The MPEG-2 packet stream MUST be encoded according to [ETSI EN 300 429].

Annex B

Additions and modifications for Japanese specification

(This annex forms an integral part of this Recommendation)

This annex applies to the third technology option referred to in clause 1 (1.1 Scope). For the first option, refer to clauses 4.2, 6 and 7. For the second option, refer to Annex A.

This annex describes the physical layer specifications required for the integrated CMTS and EQAM for Japanese specification. This is an optional annex and in no way affects certification of equipment adhering to the North American technology option described in the clauses referenced above.

The numbering of the paragraphs has been maintained such that the suffix after the letter for the annex refers to the part of the specification where the described changes apply. As a consequence, some heading numbers might be missing in this annex, since no change is required to the relevant paragraph in the main body of the Recommendation.

B.1 Scope and purpose

See clause 1.

B.2 References

See clause 2.

B.3 Terms and definitions

See clause 3.

B.4 Acronyms and abbreviations

See clause 4.

B.5 Functional assumptions

See clause 5.

B.5.1 Broadband access network

See clause 5.1.

B.5.2 Equipment assumptions

B.5.2.1 Frequency plan

In the downstream direction, the cable system is assumed to have a passband with a lower edge 90 MHz and an upper edge that is implementation-dependent but is typically in the range of 350 to 770 MHz. Within that passband, NTSC analog television signals in 6-MHz channels are assumed present on the Japan frequency plan, as well as other narrow-band and wideband digital signals.

B.5.2.2 Compatibility with other services

See clause 5.2.2.

B.5.2.3 Fault isolation impact on other users

See clause 5.2.3.

B.5.3 Downstream plant assumptions

See clause 5.3.

B.5.3.1 Transmission levels

See clause 5.3.1.

B.5.3.2 Frequency inversion

See clause 5.3.2.

B.5.3.3 Analog and digital channel line-up

In developing this technology option, it was assumed that a maximum of 110 digital channels would be deployed in a headend. For the purposes of calculating CNR, protection for analog channels, it was assumed that analog channels are placed at lower frequencies in the channel line-up, versus digital channels.

B.5.3.4 Analog protection goal

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 60 dB for systems deploying up to 110 DRFI-compliant QAM channels.

The Recommendation assumes that the transmitted power level of the digital channels will be 6 dB below the peak envelope power of the visual signal of analog channels, which is more stringent than the typical condition of 10 dB below for digital transmission. It is further assumed that the channel line-up will place analog channels at lower frequencies versus digital channels. An adjustment of $10 \cdot \log_{10}(6 \text{ MHz}/4 \text{ MHz})$ is used to account for the difference in noise bandwidth of digital channels, versus analog channels. With the assumptions above, for a 110 QAM channel system, the specification in item 5 of Table B.4 equates to an analog CNR protection of 60 dB.

B.6 Physical media dependent sublayer specification

B.6.1 Scope

This clause applies to the third technology option referred to in clause 1 (1.1 Scope). In cases where the requirements for all technology options are identical, a reference is provided to the main text.

For the remainder of this clause, see clause 6.1.

B.6.2 EdgeQAM (EQAM) differences from CMTS

See clause 6.2.

B.6.3 Downstream

B.6.3.1 Downstream protocol

The downstream PMD sublayer MUST conform to [ITU-T J.83-C]. Interleaver depths are defined in clause B.6.3.3. The applicability of a particular interleaver depth depends on the data service provided on a particular QAM RF channel. Applicability of interleaver depths for service delivery, other than DOCSIS high-speed data, is beyond the scope of this annex.

B.6.3.2 Spectrum format

The downstream modulator for each QAM channel of the EQAM, or CMTS, MUST support operation with the 64 QAM constellation, as specified in [ITU-T J.83-C], and the 256 QAM constellation, as specified in clause J.6.3.4.2 of [ITU-T J.122].

B.6.3.3 Scaleable interleaving to support video and high-speed data services

The CMTS or EQAM downstream PMD sublayer MUST support the interleaver with the characteristics defined in Table B.1.

The interleave method for 256QAM is equal to [ITU-T J.83-C] except for interleave depth values. The interleave depth $I = 12$ MUST be supported by the CMTS or EQAM. The CMTS or EQAM MAY support interleaver depths of $I = 34$ or $I = 204$. Table B.1 shows the interleaver characteristics at 5.274 Msymb/s.

Table B.1 – Interleaver characteristics

I (Number of taps)	J (Increment)	Burst protection 64QAM/256QAM	Latency 64QAM/256QAM
12	17	24 μ s/18 μ s	0.57 ms/0.43 ms
34	6	-/51 μ s	-/1.28 ms
204	1	-/300 μ s	-/7.85 ms

B.6.3.4 Downstream frequency plan

The downstream frequency plan SHOULD comply with the frequency plan being used by the cable system in which it will operate. For example, this might be the Japan frequency plan for digital QAM carriers [Article 23-(1)]. Operational frequencies MUST include all channels between 90 MHz and 770 MHz.

B.6.3.5 DRFI output electrical

See clause 6.3.5.

B.6.3.5.1 CMTS or EQAM output electrical

A CMTS or EQAM MUST output an RF modulated signal with the characteristics defined in Tables B.2, B.3 and B.4. The condition for these requirements is all N combined channels, commanded to the same average power, except for the single channel active phase noise and diagnostic carrier suppression (Table B.3) requirements.

Table B.2 – RF output electrical requirements

Parameter	Value
Centre frequency (f_c) of any RF channel of a CMTS or EQAM	MUST be 93 MHz to 767 MHz
Level	Adjustable. See Table B.3
Modulation Type	64QAM, 256QAM
Symbol rate (nominal) 64QAM 256QAM	5.274 Msym/s 5.274 Msym/s
Nominal channel spacing	6 MHz
Frequency response 64QAM 256QAM	~ 0.13 Square Root Raised Cosine Shaping ~ 0.13 Square Root Raised Cosine Shaping

Table B.2 – RF output electrical requirements

Parameter	Value
Inband Spurious, Distortion and Noise	Unequalized MER (Note 2) > 35 dB Equalized MER > 43 dB
Inband Spurious and Noise	$\leq -48\text{dBc}$; where channel spurious and noise includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, and other undesired transmitter products. Spurious and noise within ± 50 kHz of the carrier is excluded. When $N > 1$, noise outside the Nyquist bandwidth is excluded.
Out-of-Band Spurious and Noise	See Table B.4
Phase Noise Single Channel Active, $N - 1$ Channels Suppressed (see clause B.6.3.5.1.2 (6)) 64QAM and 256QAM	1 kHz – 10 kHz: –33 dBc double sided noise power 10 kHz – 50 kHz: –51 dBc double sided noise power 50 kHz – 3 MHz: –51 dBc double sided noise power
All N Channels Active, (see clause B.6.3.5.1.2 (7)) 64QAM and 256QAM	1 kHz – 10 kHz: –33 dBc double sided noise power 10 kHz – 50 kHz: –51 dBc double sided noise power
Output Impedance	75 ohms
Output Return Loss (Note 3)	> 14 dB within an active output channel from 90 MHz to 770 MHz (Note 4)
Connector	F connector per [IEC 60169-24]]
<p>NOTE 1 – 30 kHz includes an allowance of 25 kHz for the largest FCC frequency offset normally built into upconverters.</p> <p>NOTE 2 – MER (modulation error ratio) is determined by the cluster variance caused by the transmit waveform at the output of the ideal receive matched filter. MER includes all discrete spurious, noise, carrier leakage, clock lines, synthesizer products, distortion, and other undesired transmitter products. Unequalized MER also includes linear filtering distortion, which is compensated by a receive equalizer. Phase noise up to ± 50 kHz of the carrier is excluded from inband specification, to separate the phase noise and inband spurious requirements as much as possible. In measuring MER, record length or carrier tracking loop bandwidth may be adjusted to exclude low frequency phase noise from the measurement. For equalized MER, receive equalizer coefficients are computed and applied with receiver operating with device under test. For unequalized MER, receive equalize coefficients may be computed to flatten receiver response, if necessary, then are held fixed when device under test is connected. MER requirements assume measuring with a calibrated test instrument with its residual MER contribution removed.</p> <p>NOTE 3 – Frequency ranges are edge-to-edge.</p> <p>NOTE 4 – If the EQAM or CMTS provides service to a centre frequency of 93 MHz (see line 1 in the table), then the EQAM or CMTS MUST provide a return loss of > 14 dB within an active output channel, from 90 MHz to 770 MHz (f_{edge}).</p>	

B.6.3.5.1.1 Power per channel CMTS or EQAM

An EQAM or CMTS MUST generate an RF output with power capabilities as defined in Table B.3. Channel RF power MAY be adjustable on a per channel basis with each channel independently meeting the power capabilities defined in Table B.3. If the EQAM or CMTS has independent modulation capability on a per channel basis, then the channel RF power MUST be adjustable on a per channel basis, with each channel independently meeting the power capabilities defined in Table B.3.

Table B.3 – DRFI device output power

Parameter	Value
Range of commanded transmit power per channel	≥ 8 dB below required power level specified below maintaining full fidelity over the 8 dB range.
Commanded power per channel step size	≤ 0.2 dB Strictly monotonic
Power difference between any two adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 0.5 dB
Power difference between any two non-adjacent channels in a block (with commanded power difference removed if channel power is independently adjustable)	≤ 1 dB
Power per channel absolute accuracy	± 2 dB
Diagnostic carrier suppression (3 modes) Mode 1: One channel suppressed Mode 2: All channels suppressed except one Mode 3: All channels suppressed	<ol style="list-style-type: none"> 1) ≥ 50 dB carrier suppression within the Nyquist bandwidth in any one 6 MHz channel in the block. This MUST be accomplished without discontinuity or detriment to the other channels in the block. 2) 50 dB carrier suppression within the Nyquist bandwidth in every 6 MHz channel in the block except one. This MUST be accomplished without discontinuity or detriment to the remaining channel in the block. 3) 50 dB carrier suppression within the Nyquist bandwidth in every 6 MHz channel in the block.
RF block muting	≥ 73 dB below the unmuted aggregate power of the block, in every 6 MHz channel in the block
Required power per channel for N channels combined onto a single RF port. ' N ' = number of combined channels: $N = 1$ $N = 2$ $N = 3$ $N = 4$ $N > 4$	Required power in dB μ V per channel 120 dB μ V 116 dB μ V 114 dB μ V 112 dB μ V 120 – ceil [3.6*log ₂ (N)] dB μ V

B.6.3.5.1.2 Independence of individual channel within the multiple channels on a single RF port

A potential use of a CMTS or an EQAM is to provide a universal platform that can be used for high-speed data services or for video services. For this reason, it is essential that interleaver depth be set on a per channel basis to provide a suitable transmission format for either video or data as needed in normal operation. Any N -channel block of a CMTS or EQAM MUST be configurable with at least interleaver depth $I = 12$. Although not as critical as per-channel interleaver depth control, there are strong benefits for the operator if the EQAM is provided with the ability to set RF power, centre frequency, and modulation type on a per-channel basis.

- 1) A multiple-channel CMTS or EQAM MUST be configurable with at least interleaver depth $I = 12$ among the N channels on an RF output port, with each channel using one of the two (or more) interleaver depths, on a per channel basis, see Table B.1 for information on interleaver depths.

- 2) A multiple-channel CMTS or EQAM MUST provide for 3 modes of carrier suppression of RF power for diagnostic and test purposes, see Table B.3 for mode descriptions and carrier RF power suppression level.
- 3) A multiple-channel CMTS or EQAM MAY provide for independent adjustment of RF power in a per channel basis with each RF carrier independently meeting the requirements defined in Table B.3.
- 4) A multiple-channel CMTS or EQAM MAY provide for independent selection of centre frequency on a per channel basis, thus providing for non-contiguous channel frequency assignment with each channel independently meeting the requirements in Table B.2.
- 5) A multiple-channel CMTS or EQAM MAY provide for independent selection of modulation order, either 64QAM or 256QAM, on a per channel basis, with each channel independently meeting the requirements in Table B.2.
- 6) A CMTS or EQAM MUST provide a test mode of operation, for out-of-service testing, configured for N channels but generating one-CW-per-channel, one channel at a time at the centre frequency of the selected channel; all other combined channels are suppressed. One purpose for this test mode is to support one method for testing the phase noise requirements of Table B.2. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance; for example, repeated selection of a constellation symbol with power close to the constellation RMS level would seemingly exercise much of the modulation and up-conversion chain in a realistic manner. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table B.2.
- 7) A CMTS or EQAM MUST provide a test mode of operation, for out-of-service testing, generating one-CW-per-channel, at the centre frequency of the selected channel, with all other $N - 1$ of the combined channels active and containing valid data modulation at operational power levels. One purpose for this test mode is to support one method for testing the phase noise requirements of Table B.2. As such, the generation of the CW test tone SHOULD exercise the signal generation chain to the fullest extent practicable, in such manner as to exhibit phase noise characteristics typical of actual operational performance. For example, a repeated selection of a constellation symbol, with power close to the constellation RMS level, would seemingly exercise much of the modulation and upconversion chain in a realistic manner. For this test mode, it is acceptable that all channels operate at the same average power, including each of the $N - 1$ channels in valid operation, and the single channel with a CW tone at its centre frequency. The test mode MUST be capable of generating the CW tone over the full range of centre frequency in Table B.2.

If either centre frequency 4) or modulation type 5), or both are independently adjustable on a per channel basis, then the CMTS or EQAM MUST provide for independent adjustment of RF power 3) on a per channel basis, with each RF carrier independently meeting the requirements defined in Table B.2.

B.6.3.5.1.3 Out-of-band noise and spurious requirements for CMTS or EQAM

One of the goals of the DRFI Recommendation is to provide the minimum intended analog channel CNR protection of 60 dB for systems deploying up to 110 DRFI-compliant QAM channels.

This Recommendation assumes that the transmitted power level of the digital channels will be 6 dB below the peak envelope power of the visual signal of analog channels, which is more stringent than the typical condition of 10 dB below for digital transmission. It is further assumed that the channel line-up will place analog channels at lower frequencies than digital channels. An adjustment of $10 \cdot \log_{10}(6 \text{ MHz}/4 \text{ MHz})$ is used to account for the difference in noise bandwidth of digital

channels, versus analog channels. With the assumptions above, for a 110 QAM channel system, the specification in item 5 of Table B.4 equates to an analog CNR protection of 60 dB.

Table B.4 lists the out-of-band spurious requirements. In cases where the N combined channels are not commanded to the same power level, "dBc" denotes decibels relative to the strongest carrier in the channel block. The out-of-band spurious emissions requirements assume a test condition with a contiguous block of N combined channels commanded to the same power level, and for this test condition "dBc" should be interpreted as the average channel power, averaged over the block, to mitigate the variation in channel power across the block (see Table B.3) which is allowed with all channels commanded to the same power.

Items 1 through 4 list the requirements in channels adjacent to the commanded channels.

Item 5 lists the requirements in all other channels further from the commanded channels. Some of these "other" channels are allowed to be excluded from meeting the Item 5 specification. All the exclusions, such as 2nd and 3rd harmonics of the commanded channel, are fully identified in the table.

Item 6 lists the requirements on the $2N$ 2nd harmonic channels and the $3N$ 3rd harmonic channels.

Table B.4 – EQAM or CMTS output out-of-band noise and spurious emissions requirements

Item	Band	N , Number of combined channels per RF port				
		1	2	3	4	$N > 4$ All equations are ceiling (Power, 0.5) dBc
1	Adjacent channel up to 750 kHz from channel block edge	< -58 dBc	< -58 dBc	< -58 dBc	< -58 dBc	$< 10 * \log_{10} [10^{-58/10} + (0.75/6) * (10^{-65/10} + (N-2) * 10^{-73/10})]$
2	Adjacent channel (750 kHz from channel block edge to 6 MHz from channel block edge)	< -62 dBc	< -60 dBc	< -60 dBc	< -60 dBc	$< 10 * \log_{10} [10^{-62/10} + (5.25/6) * (10^{-65/10} + (N-2) * 10^{-73/10})]$
3	Next-adjacent channel (6 MHz from channel block edge to 12 MHz from channel block edge)	< -65 dBc	< -64 dBc	< -63.5 dBc	< -63 dBc	$< 10 * \log_{10} [10^{-65/10} + (N-1) * 10^{-73/10}]$
4	Third-adjacent channel (12 MHz from channel block edge to 18 MHz from channel block edge)	< -73 dBc	< -70 dBc	< -67 dBc	< -65 dBc	For $N = 5$: -64.5 dBc; For $N = 6$: -64 dBc; For $N \geq 7$: $< -73 + 10 * \log_{10}(N)$ dBc
5	Noise in other channels (80 MHz to 780 MHz) Measured in each 6 MHz channel excluding the following: a) Desired channel(s) b) 1st, 2nd, and 3rd adjacent channels (see Items 1, 2, 3, 4 in this table) c) Channels coinciding with 2nd and 3rd harmonics (see Item 6 in this table)	< -73 dBc	< -70 dBc	< -68 dBc	< -67 dBc	$< -73 + 10 * \log_{10}(N)$

Table B.4 – EQAM or CMTS output out-of-band noise and spurious emissions requirements

Item	Band	N, Number of combined channels per RF port				
		1	2	3	4	N > 4 All equations are ceiling (Power, 0.5) dBc
6	In each of 2N contiguous 6 MHz channels or in each of 3N contiguous 6 MHz channels coinciding with 2nd harmonic and with 3rd harmonic components respectively (up to 1000 MHz)	$< -73 + 10 \cdot \log_{10}(N)$, or -63 dBc, whichever is greater				

B.6.3.5.2 CMTS or EQAM master clock jitter for asynchronous operation

An EQAM MUST implement a DTI client and client interface per [ITU-T J.211]. The master clock specifications are defined in [ITU-T J.211]. The DTI client provides the master clock. An integrated CMTS not actively serviced by a DTI server MUST include a master clock with the specifications as follows:

The 9.216 MHz master clock MUST have, over a temperature range of 0 to 40 degrees C and for up to ten years from date of manufacture (see Note below):

- A frequency accuracy of $\leq \pm 5$ ppm;
- A drift rate $\leq 10^{-8}$ per second; and
- An edge jitter of ≤ 10 ns peak-to-peak (± 5 ns).

NOTE – This specification MAY also be met by synchronizing the DRFI device master clock oscillator to an external frequency reference source. If this approach is used, the internal DRFI device master clock MUST have a frequency accuracy of ± 20 ppm over a temperature range of 0 to 40 degrees C, up to 10 years from date of manufacture, when no frequency reference source is connected. The drift rate and edge jitter MUST be as specified above.

The drift rate and jitter requirements on the DRFI device master clock implies that the duration of two adjacent segments of 9'216'000 cycles will be within 30 ns, due to 10 ns jitter on each segments' duration, and 10 ns due to frequency drift. Durations of other counter lengths also may be deduced: adjacent 921'600 segments, ≤ 21 ns; 921'600 length segments separated by one 9'216'000-cycle segment, ≤ 30 ns; adjacent 92'160'000 segments, ≤ 120 ns. The DRFI device master clock MUST meet such test limits in 99% or more measurements.

B.6.3.5.3 CMTS or EQAM master clock jitter for synchronous operation

In addition to the requirements in clause B.6.3.5.2, the 9.216 MHz CMTS master clock MUST meet the following double sideband phase noise requirements over the specified frequency ranges:

- $< [-50 + 20 \cdot \log(f_{MC}/9.216)]$ dBc (*i.e.*, < 0.05 ns RMS) 10 Hz to 100 Hz
- $< [-58 + 20 \cdot \log(f_{MC}/9.216)]$ dBc (*i.e.*, < 0.02 ns RMS) 100 Hz to 1 kHz
- $< [-50 + 20 \cdot \log(f_{MC}/9.216)]$ dBc (*i.e.*, < 0.05 ns RMS) 1 kHz to 10 kHz
- $< [-50 + 20 \cdot \log(f_{MC}/9.216)]$ dBc (*i.e.*, < 0.05 ns RMS) 10 kHz to $f_{MC}/2$

f_{MC} is the frequency of the measured master clock in MHz. The value of f_{MC} MUST be either an integral multiple or divisor of 9.216 MHz. For example, if a 18.432 MHz oscillator is used as the master clock frequency source, and there is no explicit 9.216 MHz clock to test, the 18.432 MHz clock may be used with f_{MC} equal to 18.432 in the above expressions.

Specifications for EQAM master clock jitter in synchronous operation are contained in [ITU-T J.211].

B.6.3.5.4 CMTS or EQAM master clock frequency drift for synchronous operation

See clause 6.3.5.4.

B.6.3.6 CMTS or EQAM clock generation

When the 9.216 MHz master clock is provided by the DTI interface, a DRFI-compliant device MUST lock the downstream symbol clock to the 9.216 MHz master clock using the M/N divisors provided in Table B.5.

B.6.3.6.1 CMTS clock generation

The CMTS MUST lock the downstream symbol clock to the CMTS master clock using the M/N divisors provided in Table B.5.

B.6.3.6.2 EQAM clock generation

Because it operates with an active DTI interface, an EQAM MUST lock the downstream symbol clock to the master clock using the M/N divisors provided in Table B.5.

B.6.3.6.3 Downstream symbol rate

Let f'_b represent the rate of the downstream symbol clock which is locked to the master clock and let f'_m represent the rate of the master clock locked to the downstream symbol clock. Let f_b represent the nominal specified downstream symbol rate and let f_m represent the nominal master clock rate (9.216 MHz). With the downstream symbol clock locked to the master clock, the following equation MUST hold:

$$f'_b = f_m * M/N$$

With the master clock locked to the downstream symbol clock, the following equation MUST hold:

$$f'_m = f_b * N/M$$

Note that M and N in Table B.5 are unsigned integer values, each representable in 16 bits and result in a value of f'_b or f'_m that is not more than ± 1 ppm from its specified nominal value.

The standard deviation of the timing error of the EQAM/CMTS RF symbol clock, referenced to the DTI server master clock, MUST be less than 1.5 ns measured over 100 seconds.

Table B.5 lists the downstream modes of operation, their associated nominal symbol rates, f_b , values for M and N , the resulting synchronized clock rates and their offsets from their nominal values.

**Table B.5 – Downstream symbol rates & parameters
for synchronization with master clock**

Downstream mode	Nominal specified symbol rate, f_b (MHz)	M/N	Master clock rate, f'_m (MHz)	Downstream symbol rate, f'_b (MHz)	Offset from nominal
64QAM and 256QAM	5.274	293/512	9.216	5.274	0 ppm

B.6.3.7 Downstream symbol clock jitter for synchronous operation

The downstream symbol clock MUST meet the following double sideband phase noise requirements over the specified frequency ranges:

- $< [-53 + 20 \cdot \log(f_{DS}/5.274)]$ dBc (*i.e.*, < 0.07 ns RMS) 10 Hz to 100 Hz
- $< [-53 + 20 \cdot \log(f_{DS}/5.274)]$ dBc (*i.e.*, < 0.07 ns RMS) 100 Hz to 1 kHz
- $< [-53 + 20 \cdot \log(f_{DS}/5.274)]$ dBc (*i.e.*, < 0.07 ns RMS) 1 kHz to 10 kHz
- $< [-36 + 20 \cdot \log(f_{DS}/5.274)]$ dBc (*i.e.*, < 0.5 ns RMS) 10 kHz to 100 kHz
- $< [-30 + 20 \cdot \log(f_{DS}/5.274)]$ dBc (*i.e.*, < 1 ns RMS) 100 kHz to $(f_{DS}/2)$

f_{DS} is the frequency of the measured clock in MHz. The value of f_{DS} MUST be an integral multiple or divisor of the downstream symbol clock. For example, an $f_{DS} = 21.096$ MHz clock may be measured if there is no explicit 5.274 MHz clock available.

A DRFI-compliant device MUST provide a means for clock testing in which:

- The device provides test points for direct access to the master clock and the downstream symbol clock.

Alternatively, a DRFI-conformant device MUST provide a test mode in which:

- The downstream QAM symbol sequence is replaced with an alternating binary sequence (1, -1, 1, -1, 1, -1...) at nominal amplitude, on both I and Q.
- The device generates the downstream symbol clock from the 9.216 MHz reference clock as in normal synchronous operation.

If an explicit downstream symbol clock, which is capable of meeting the above phase noise requirements, is available (e.g., a smooth clock without clock domain jitter), this test mode is not required.

B.6.3.8 Downstream symbol clock drift for synchronous operation

See clause 6.3.8.

B.7 Downstream transmission convergence sublayer

B.7.1 Introduction

See clause 7.1.

B.7.2 MPEG packet format

See clause 7.2.

B.7.3 MPEG header for DOCSIS data-over-cable

See clause 7.3.

B.7.4 MPEG payload for DOCSIS data-over-cable

See clause 7.4.

B.7.5 Interaction with the MAC sublayer

See clause 7.5.

B.7.6 Interaction with the physical layer

The MPEG-2 packet stream MUST be encoded according to [ITU-T J.83-C], including MPEG-2 transport framing using a parity checksum as described in [ITU-T J.83-C].

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