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G.728

Annex H

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SERIES G: TRANSMISSION SYSTEMS AND MEDIA,
DIGITAL SYSTEMS AND NETWORKS

Digital transmission systems – Terminal equipments –
Coding of analogue signals by methods other than PCM

Coding of speech at 16 kbit/s using low-delay code
excited linear prediction

**Annex H: Variable bit rate LD-CELP operation
mainly for DCME at rates less than 16 kbit/s**

ITU-T Recommendation G.728 – Annex H
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ITU-T RECOMMENDATION G.728

CODING OF SPEECH AT 16 kbit/s USING LOW-DELAY CODE EXCITED LINEAR PREDICTION

ANNEX H

**Variable bit rate LD-CELP operation mainly
for DCME at rates less than 16 kbit/s**

Source

ITU-T Recommendation G.728, Annex H was prepared by ITU-T Study Group 16 (1997-2000) and was approved under the WTSC Resolution No. 1 procedure on the 10th of July 1997.

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FOREWORD

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Recommendation G.728

CODING OF SPEECH AT 16 kbit/s USING LOW-DELAY CODE EXCITED LINEAR PREDICTION

ANNEX H¹

Variable bit rate LD-CELP operation mainly for DCME at rates less than 16 kbit/s

(Geneva, 1997)

H.1 Introduction

This Annex contains the modifications to Recommendation G.728 LD-CELP speech coding algorithm needed to reduce the coding bit rate down to 12.8 and 9.6 kbit/s. These modifications include the modifications to the shape and gain codebooks.

This Annex assumes that the reader is already familiar with the specifications in Recommendation G.728. The G.728 algorithm will not be discussed here; only the modifications to Recommendation G.728 will be described.

This annex consists of four subclauses. Subclause H.2 describes the principle of the operation for bit rate reduction. Subclause H.3 describes the modifications required for 12.8 kbit/s operation. Subclause H.4 describes the modifications required for 9.6 kbit/s operation.

H.2 Principles of operation

H.2.1 Procedure to reduce coding bit rate

Reducing the bit rate without significantly changing the coding algorithm can be accomplished by reducing the size of the codebook. In the main body of Recommendation G.728, the codebook index consists of 10 bits corresponding to 1024 codebook vectors. A 2-bit reduction out of this 10-bit codebook index reduces the coding bit rate from 16 kbit/s to 12.8 kbit/s; a 4-bit reduction reduces the coding bit rate to 9.6 kbit/s.

The codebook index (bit 9 to bit 0) is divided into two sections: 7 bits (bit 9 to bit 3) for the shape codebook and 3 bits (bit 2 to bit 0) for the gain codebook. The 7-bit shape codebook consists of 128 code vectors; the 3-bit gain codebook consists of 8 scalar values that are symmetric with respect to zero.

The number of vectors in the shape codebook can be reduced as follows. The probability of vectors in the shape codebook obtained from normal speech samples does not have a uniform distribution. The occurrence probability of the codebook index numbers from 65 to 128 is observed to be much higher than those from 1 to 64. By taking advantage of this non-uniform distribution, it is possible to restrict the number of codebook vectors without introducing very much degradation in speech quality. For example, bit 9 in the shape codebook index is a candidate for reduction.

¹ This Annex is an optional addition to Recommendation G.728 and is not required in order to correctly implement the encoder and decoder. This Annex is intended to enhance the performance of Recommendation G.728 in some specific applications, such as in those to be used with digital circuit multiplication equipment. It is left to the implementers to choose whether to use this Annex.

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Another way to reduce the size of the shape codebook is to redesign the optimized codebook for each reduced bit rate operation. However, this would require more memory locations and a significant modification in the implementation.

The number of gain values in the gain codebook can be reduced as follows. To reduce the number of bits used for the gain codebook, it is better to redesign the reduced gain codebook, optimizing it for the reduced bit operations, because the gain codebook will occupy only a small amount of memory. Each reduced size gain codebook should be optimized based on the probability distribution of the gain values before quantization.

Reduction in the number of codebook index bits for 12.8 kbit/s operation is described in H.2.2 and that for 9.6 kbit/s operation is described in H.2.3.

H.2.2 Principle of 12.8 kbit/s operation

Two bits must be eliminated from the 10-bit codebook index to achieve 12.8 kbit/s operation. Bit 9 of the shape codebook index is not used and a reduced gain codebook with four values is selected. Eliminating bit 9 from the shape codebook index restricts the codebook index numbers to the range 65 to 128. The four values of the gain codebook and the related values are optimized for 12.8 kbit/s operation; they are shown in Table H.1/G.728 (for floating-point calculation) and in Table H.2/G.728 (for fixed-point calculation) in H.3.2.

H.2.3 Principle of 9.6 kbit/s operation

Reducing the 10-bit codebook index by 4 bits is necessary to achieve 9.6 kbit/s operation. Bits 9, 8 and 5 of the shape codebook index are eliminated, and the gain codebook is reduced from 8 to 4 values.

Eliminating bits 9, 8 and 5 from the shape codebook index limits the codebook index numbers to the ranges 97 to 100, 105 to 108, 113 to 116 and 121 to 124. The four values of the gain codebook and the related values are optimized for 9.6 kbit/s operation; they are shown in Table H.4/G.728 (for floating-point calculation) and Table H.5/G.728 (for fixed-point calculation) in H.4.2.

H.3 Modifications for 12.8 kbit/s operation

H.3.1 Pseudo-code

Only the block execution sequences are shown; the low level details of parameter passing are not described.

H.3.1.1 Blocks 17 and 18 – Error calculator and the best codebook index selector

Both the floating-point and fixed-point pseudo-codes for Blocks 17 and 18 are given in this subclause. These codes have been modified for 12.8 kbit/s operation and should be substituted for the original Blocks 17 and 18 described in 5.11 of Recommendation G.728. The floating-point pseudo-code is presented first.

Initialize DISTM to the largest number representable in the hardware

$N1 = NG_128/2$

For $J=65,66,\dots,NCWD$, do the following

$J1 = (J-1) * IDIM$

$COR = 0$.

For $K=1,2,\dots,IDIM$, do the next line

$COR = COR + PN(K) * Y(J1+K)$ | compute inner product p_j

If $COR > 0$, then do the next 3 lines

$IDXG = N1$

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If $COR < GB_128(1)*Y2(J)$, do the next line
 $IDXG=1$ | Best positive gain found

If $COR \leq 0$, then do the next 3 lines
 $IDXG=NG_128$

If $COR > GB_128(3)*Y2(J)$, do the next line
 $IDXG=3$ | Best negative gain found

$D=-G2_128(IDXG)*COR+GSQ_128(IDXG)*Y2(J)$ | Compute distortion \hat{D}

If $D < DISTM$, do the next 3 lines
 $DISTM=D$ | Save the lowest distortion
 $IG=IDXG$ | and the best codebook indices
 $IS=J$ | so far.

Repeat the above indented section for the next J

$IS1=IS-NCWD/2$

$ICHAN=(IS1-1)*NG_128+(IG-1)$ | Concatenate shape and gain
| codebook indices.

Transmit ICHAN through communication channel. For bit-serial transmission, the most significant bit of ICHAN should be transmitted first. If ICHAN is represented by the 8-bit word b7, b6, b5, b4, b3, b2, b1, b0, then the order of the transmitted bits shall be b7, followed by b6, b5, b4, b3, b2, b1, b0 (b7 is the most significant bit).

The fixed-point version of the same module is given here. This fixed-point pseudo-code replaces the original Block 17 fixed-point code in G.3.9 of Recommendation G.728.

$DISTM=2147483647$

For $J=65,66,\dots,NCWD$, do the following

$J1=(J-1)*IDIM$

$AA0=0$

For $K=1,2,\dots,IDIM$, do the next 2 lines

$P=PN(K)*Y(J1+K)$ | compute inner product Pj
 $AA0=AA0+P$ | NLS for AA0 is 7+11=18

If $AA0 < 0$, set $AA0=-AA0$ | take absolute value

$IDXG=1$

$P=GB_128(1)*Y2(J)$ | NLS for P is 13+5=18

If $AA0 \geq P$, set $IDXG=IDXG+1$

$AA0=AA0 \gg 14$ | NLS for AA0=4

If $AA0 > 32767$, set $AA0=32767$ | clip AA0; AA0 in saturation mode

$AA1=GSQ_128(IDXG)*Y2(J)$ | NLSGSQ_128=11, NLSY2=5, so NLSAA1=16

$P=G2_128(IDXG)*AA0$ | NLSG2_128=12, NLSAA0=4, so NLSP=16

$AA1=AA1-P$

If $AA1 < DISTM$, do the next 3 lines

$DISTM=AA1$ | double precision DISTM

$IG=IDXG$

$IS=J$

Repeat the above indented section for the next J

$AA0=0$ | Now find the sign bit

$J1=(IS-1)*IDIM$

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For $K=1,2,\dots, \text{IDIM}$, do the next 2 lines

$P = \text{PN}(K) * Y(J1+K)$ | compute inner product

$\text{AA0} = \text{AA0} + P$

If $\text{AA0} \leq 0$, set $\text{IG} = \text{IG} + 2$

$\text{IS1} = \text{NCWD}$

$\text{IS1} = \text{IS1} \gg 1$

$\text{IS1} = \text{IS} - \text{IS1}$

$\text{ICHAN} = (\text{IS1} - 1) * \text{NG}_{128} + (\text{IG} - 1)$

In the above code, we used the following four lines:

$\text{AA0} = \text{AA0} \gg 14$ | NLS for $\text{AA0} = 4$

If $\text{AA0} > 32767$, set $\text{AA0} = 32767$ | clip AA0

$\text{AA1} = \text{GSQ}_{128}(\text{IDXG}) * Y2(J)$ | $\text{NLSGSQ}_{128} = 11$, $\text{NLSY2} = 5$, so $\text{NLSAA1} = 16$

$P = \text{G2}_{128}(\text{IDXG}) * \text{AA0}$ | $\text{NLSG2}_{128} = 12$, $\text{NLSAA0} = 4$, so $\text{NLSP} = 16$

In DSP chips which have a "clipping" function, these lines can be replaced with the following code to give exactly the same results.

$\text{AA0} = \text{AA0} \ll 2$ | NLS for $\text{AA0} = 20$

$\text{AA0} = \text{CLIP}(\text{AA0})$ | AA0 is in saturation mode

$\text{AA0} = \text{AA0} \gg 16$ | take high word; NLS for $\text{AA0} = 4$

$\text{AA1} = \text{GSQ}_{128}(\text{IDXG}) * Y2(J)$ | $\text{NLSGSQ}_{128} = 11$, $\text{NLSY2} = 5$, so $\text{NLSAA1} = 16$

$P = \text{G2}_{128}(\text{IDXG}) * \text{AA0}$ | $\text{NLSG2}_{128} = 12$, $\text{NLSAA0} = 4$, so $\text{NLSP} = 16$

The CLIP function and saturation mode refer to the concept of not allowing AA0 to overflow when the $\ll 2$ operation is performed. Instead of overflow, AA0 is set to the maximum positive or negative number, depending on its original sign. In this case, AA0 is always positive. This alternative is DSP dependent and may require more than a 32-bit accumulator. The alternative in the main pseudo-code can always be implemented.

H.3.1.2 Block 19 – Excitation VQ codebook and Block 21 – Gain scaling unit

Both the floating-point and fixed-point pseudo-code for Blocks 19 and 21 are given in this subclause. This code has been modified for the case of 12.8 kbit/s operation and should be substituted for the original Block 19 described in 5.12 of Recommendation G.728. This is the floating-point version of the pseudo-code for Block 19, the excitation VQ codebook.

$\text{NN} = (\text{IS} - 1) * \text{IDIM}$

For $K=1,2,\dots, \text{IDIM}$, do the next line

$\text{YN}(K) = \text{GQ}_{128}(\text{IG}) * Y(\text{NN} + K)$

The floating-point version of the pseudo-code for Block 21, the gain scaling unit is given below.

For $K=1,2,\dots, \text{IDIM}$, do the next line

$\text{ET}(K) = \text{GAIN} * \text{YN}(K)$

The fixed-point version of the same module is given here. This fixed-point pseudo-code replaces the original Blocks 19 and 21 fixed-point code in G.3.10 of Recommendation G.728.

For the fixed-point pseudo-code, we combine Blocks 19 and 21 into a single module. Both Y and GQ_{128} have fixed Q formats, Q11 and Q13, respectively. The value of GAIN has associated with it NLSGAIN . To get the maximum accuracy, the product $\text{GQ}_{128}(\text{IG}) * \text{GAIN}$ is normalized to 32 bits before rounding to the upper 16 bits is performed. Let $\text{NNGQ}_{128}(I)$ be (1+the number of left shifts needed to normalize the

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Q13 $GQ_{128}(I)$, $NNGQ_{128}(I)=3$ for $I=1,3$ and $NNGQ_{128}(I)=2$ for $I=2,4$. The pseudo-code can thus be written as follows:

$AA0 = GQ_{128}(IG) * GAIN$ $AA0 = AA0 \ll NNGQ_{128}(IG)$ $TMP = RND(AA0)$ $NLSAA0 = 13 + NLSGAIN$ $NLSTMP = NLSAA0 + NNGQ_{128}(IG) - 16$ $NN = (IS - 1) * IDIM$ Call $VSCALE(Y(NN+1), IDIM, IDIM, 14, TEMP, NLS$	$AA0$ has $NNGQ_{128}(IG)$ leading zeros left shift $NNGQ_{128}(IG)$ bits to normalize $AA0$ round to upper 16 bits and assign to TMP Q format of the product $GQ_{128}(IG) * GAIN$ Q format of TMP , because $AA0$ left shift by $NNGQ_{128}(IG)$ bits then round and take upper 16 bits normalize selected shape codevector to 16 bits; put in $TEMP$
For $K=1,2,\dots,IDIM$, do the next 2 lines $AA0 = TMP * TEMP(K)$ $ET(K) = RND(AA0)$ $NLSET = NLSTMP + 11 + NLS - 16$	TMP and $TEMP$ both normalized to 16 bits, so the product has 1 leading zero. Directly rounding to high work gives us a 15-bit ET array. calculate the NLS for ET .

H.3.1.3 Block 29 – Decoder excitation VQ codebook and Block 31 – Decoder gain scaling unit

Both the floating-point and fixed-point pseudo-code for Blocks 29 and 31 are given in this subclause. This code has been modified for 12.8 kbit/s operation and should be substituted for the original Block 29 described in 5.14 of Recommendation G.728. This is the floating-point version of the pseudo-code for Block 29, the decoder excitation VQ codebook.

This block first extracts the 2-bit gain codebook index IG and the 6-bit shape codebook index IS from the received 8-bit channel index. The remainder of the operation is exactly the same as for Block 19 of the encoder.

$ITMP = \text{integer part of } (ICHAN / NG_{128})$
 $IG = ICHAN - ITMP * NG_{128} + 1$
 $ITMP = ITMP + NCWD / 2$

$NN = ITMP * IDIM$

For $K=1,2,\dots,IDIM$, do the next line

$YN(K) = GQ_{128}(IG) * Y(NN+K)$

The operation of Block 31, the decoder gain scaling unit is exactly the same as for Block 21 of the encoder.

The fixed-point version of the same module is given here.

For the fixed-point pseudo-code, we combine Blocks 29 and 31 into a single module. Both Y and GQ_{128} have fixed Q formats, Q11 and Q13, respectively. The value of $GAIN$ has associated with it $NLSGAIN$. To get the maximum accuracy, the product $GQ_{128}(IG) * GAIN$ is normalized to 32 bits before rounding to the upper 16 bits is performed. Let $NNGQ_{128}(I)$ be (1+the number of left shifts needed to normalize the Q13 $GQ_{128}(I)$), so $NNGQ_{128}(I) = 3$ for $I = 1, 3$ and $NNGQ_{128}(I) = 2$ for $I = 2, 4$. The pseudo-code can thus be written as follows:

$IS = ICHAN \gg 2$
 $IG = ICHAN - IS * NG_{128} + 1$
 $IS1 = NCWD$

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IS1=IS1 >> 1
IS=IS+IS1+1

AA0=GQ_128(IG)*GAIN          | AA0 has NNGQ_128(IG) leading zeros
AA0=AA0 << NNGQ_128(IG)      | left shift NNGQ_128(IG) bits to normalize AA0
TMP=RND(AA0)                  | round to upper 16 bits and assign to TMP

NLSAA0=13+NLSGAIN             | Q format of the product GQ_128(IG)*GAIN
NLSTMP=NLSAA0+NNGQ_128(IG)-16 | Q format of TMP, because AA0 left
                                | shift by NNGQ_128(IG) bits then round
                                | and take upper 16 bits

NN=(IS-1)*IDIM                | normalize selected
Call VSCALE(Y(NN+1), IDIM, IDIM, 14 ,TEMP, NLS) | shape codevector to
                                | 16 bits; put in TEMP

For K=1,2,...,IDIM, do the next 2 lines | TMP and TEMP both normalized
AA0=TMP*TEMP(K)                | to 16 bits, so the product
ET(K)=RND(AA0)                 | has 1 leading zero. Directly
                                | rounding to high work gives
                                | us a 15-bit ET array.

NLSET=NLSTMP+11+NLS-16        | calculate the NLS for ET.

```

H.3.2 Additional new gain table

This subclause gives the values for the gain codebook for 12.8 kbit/s operation. The floating-point values are given first. See Table H.1.

Table H.1/G.728 – Floating-point values of gain codebook-related arrays

Array index	1	2	3	4
GQ_128	0.525824	1.562449	-0.525824	-1.562449
GB_128	0.869912	*	-0.869912	*
G2_128	1.051648	3.124898	-1.051648	-3.124898
GSQ_128	0.276491	2.441247	0.276491	2.441247

The fixed-point values are given next. See Table H.2.

Table H.2/G.728 – Fixed-point values of gain codebook-related arrays

Array index	1	2	3	4
GQ_128(Q13)	4 308	12 800	-4 308	-12 800
GB_128(Q13)	7 126	*	-7 126	*
G2_128(Q12)	4 308	12 800	-4 308	-12 800
GSQ_128(Q11)	566	5 000	566	5 000
NNGQ_128(Q0)	3	2	3	2

H.3.3 Change of coder parameter

This subclause contains the new parameter, NG_128. This parameter has been changed against NG (value = 8) of the G.728 for 12.8 kbit/s operation. See Table H.3.

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Table H.3/G.728 – Basic coder parameters of LD-CELP

Name	Value	Description
NG_128	4	Gain codebook size (number of gain levels)

H.4 Modifications for 9.6 kbit/s operation

H.4.1 Pseudo-code

Only the block execution sequences are shown; the low level detail of parameter passing is not described.

H.4.1.1 Blocks 17 and 18 – Error calculator and best codebook index selector

Both the floating-point and fixed-point pseudo-codes for Blocks 17 and 18 are given in this subclause. These codes have been modified for 9.6 kbit/s operation and should be substituted for the original Blocks 17 and 18 described in 5.11 of Recommendation G.728. The floating-point pseudo-code is presented first.

Initialize DISTM to the largest number representable in the hardware

$N1 = NG_96 / 2$

For $K=1,2,3,4$, do the following

For $K1=97,98,99,100$, do the following

$J = (K-1) * 8 + K1$

$J1 = (J-1) * IDIM$

$COR = 0$.

For $K2=1,2,\dots,IDIM$, do the next line

$COR = COR + PN(K2) * Y(J1 + K2)$ | compute inner product p_j

If $COR > 0$, then do the next 3 lines

$IDXG = N1$

If $COR < GB_96(1) * Y2(J)$, do the next line

$IDXG = 1$ | Best positive gain found

If $COR \leq 0$, then do the next 3 lines

$IDXG = NG_96$

If $COR > GB_96(3) * Y2(J)$, do the next line

$IDXG = 3$ | Best negative gain found

$D = -G2_96(IDXG) * COR + GSQ_96(IDXG) * Y2(J)$ | Compute distortion \hat{D}

If $D < DISTM$, do the next 3 lines

$DISTM = D$ | Save the lowest distortion

$IG = IDXG$ | and the best codebook

$IS = J$ | indices so far

Repeat the above indented section for the next $K1$.

Repeat the above indented section for the next K .

$IS1 = IS - (NCWD / 2 + NCWD / 4)$

$IS2 =$ integer part of $(IS1 / 8)$

$IS2 = IS2 * 4$

$IS3 = IS1 - IS2 * 2$

$IS1 = IS2 + IS3$

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ICHAN=(IS1-1)*NG_96+(IG-1) | Concatenate shape and
| gain codebook indices.

Transmit ICHAN through communication channel. For bit-serial transmission, the most significant bit of ICHAN should be transmitted first. If ICHAN is represented by the 6-bit word b5, b4, b3, b2, b1, b0, then the order of the transmitted bits shall be b5, and then b4, b3, b2, b1, b0 (b5 is the most significant bit).

The fixed-point version of the same module is given here. This fixed-point pseudo-code replaces the original Block 17 fixed-point code in G.3.9 of Recommendation G.728.

DISTM=2147483647

For K=1,2,3,4, do the following

For K1=97,98,99,100, do the following

J=(K-1)*8+K1

J1=(J-1)*IDIM

AA0=0

For K2=1,2,...,IDIM, do the next 2 lines

P=PN(K2)*Y(J1+K2) | compute inner product Pj

AA0=AA0+P | NLS for AA0 is 7+11=18

If AA0 < 0, set AA0=-AA0 | take absolute value

IDXG=1

P=GB_96(1)*Y2(J) | NLS for P is 13+5=18

If AA0 ≥ P, set IDXG=IDXG+1

AA0=AA0 >> 14 | NLS for AA0=4

If AA0 > 32767, set AA0=32767 | clip AA0; AA0 in saturation
| mode

AA1=GSQ_96(IDXG)*Y2(J) | NLSGSQ_96=11, NLSY2=5, so NLSAA1=16

P=G2_96(IDXG)*AA0 | NLSG2_96=12, NLSAA0=4, so NLSAA1=16

AA1=AA1-P

If AA1 < DISTM, do the next 3 lines

DISTM=AA1 | double precision DISTM

IG=IDXG

IS=J

Repeat the above indented section for the next K1.

Repeat the above indented section for the next K.

AA0=0 | Now find the sign bit

J1=(IS-1)*IDIM

For K=1,2,..., IDIM, do the next 2 lines

P=PN(K)*Y(J1+K) | compute inner product

AA0=AA0+P

If AA0 ≤ 0, set IG=IG+2

IS2=NCWD

IS1=IS2 >> 1

IS2=IS2 >> 2

IS1=IS-(IS1+IS2)

IS2=IS1 >> 3

IS2=IS2 << 2

IS3=IS2 << 1

IS3=IS1-IS3

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IS1=IS2+IS3

ICHAN=(IS1-1)*NG_96+(IG-1)

In the above code, we used the following four lines:

AA0=AA0 >> 14	NLS for AA0=4
If AA0 > 32767, set AA0=32767	clip AA0
AA1=GSQ_96(IDXG)*Y2(J)	NLSGSQ_96=11, NLSY2=5, so NLSAA1=16
P=G2_96(IDXG)*AA0	NLSG2_96=12, NLSAA0=4, so NLSP=16

In DSP chips which have a "clipping" function, these lines can be replaced with the following code to give exactly the same results:

AA0=AA0 << 2	NLS for AA0=20
AA0=CLIP(AA0)	AA0 is in saturation mode
AA0=AA0 >> 16	take high word; NLS for AA0=4
AA1=GSQ_96(IDXG)*Y2(J)	NLSGSQ_96=11, NLSY2=5, so NLSAA1=16
P=G2_96(IDXG)*AA0	NLSG2_96=12, NLSAA0=4, so NLSP=16

The CLIP function and saturation mode refer to the concept of not allowing AA0 to overflow when the << 2 operation is performed. Instead of overflow, AA0 is set to the maximum positive or negative number, depending on its original sign. In this case, AA0 is always positive. This alternative is DSP dependent and may require more than a 32-bit accumulator. The alternative in the main pseudo-code can always be implemented.

H.4.1.2 Block 19 – Excitation VQ codebook and Block 21 – Gain scaling unit

Both the floating-point and fixed-point pseudo-code for Blocks 19 and 21 are given in this subclause. This code has been modified for 9.6 kbit/s operation and should be substituted for the original Block 19 described in 5.12 of Recommendation G.728. This is the floating-point version of the pseudo-code for Block 19, the excitation VQ codebook. First, the floating-point version of pseudo-code for Block 19 is presented.

```
NN=(IS-1)*IDIM
For K=1,2,..., IDIM, do the next line
  YN(K)=GQ_96(IG)*Y(NN+K)
```

The floating-point version of pseudo-code for Block 21, the gain scaling unit is given below.

```
For K=1,2,..., IDIM, do the next line
  ET(K)=GAIN*YN(K)
```

The fixed-point version of the same module is given here. This fixed-point pseudo-code replaces the original Blocks 19 and 21 fixed-point code in G.3.10 of Recommendation G.728.

For the fixed-point pseudo-code, we combine both Blocks 19 and 21 into a single module. Both Y and GQ_96 have fixed Q formats, Q11 and Q13, respectively. The value of GAIN has associated with it NLSGAIN. To get the maximum accuracy, the product GQ_96(IG)*GAIN is normalized to 32 bits before rounding to the upper 16 bits is performed. Let NNGQ_96(I) be (1+the number of left shifts needed to normalize the Q13 GQ_96(I)). NNGQ_96(I)=3 for I=1,3 and NNGQ_96(I)=2 for I=2,4. The pseudo-code can thus be written as follows:

AA0=GQ_96(IG)*GAIN	AA0 has NNGQ_96(IG) leading zeros
AA0=AA0 << NNGQ_96(IG)	left shift NNGQ_96(IG) bits to normalize AA0
TMP=RND(AA0)	round to upper 16 bits and assign to TMP
NLSAA0=13+NLSGAIN	Q format of the product GQ_96(IG)*GAIN
NLSTMP=NLSAA0+NNGQ_96(IG)-16	Q format of TMP, because AA0 left shift

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$IS1=IS2 \gg 1$
 $IS2=IS2 \gg 2$
 $IS=IS+IS1+IS2+1$
 $AA0=GQ_96(IG)*GAIN$ | AA0 has NNGQ_96(IG) leading zeros
 $AA0=AA0 \ll NNGQ_96(IG)$ | left shift NNGQ_96(IG) bits to normalize AA0
 $TMP=RND(AA0)$ | round to upper 16 bits and assign to TMP
 $NLSAA0=13+NLSGAIN$ | Q format of the product $GQ_96(IG)*GAIN$
 $NLSTMP=NLSAA0+NNGQ_96(IG)-16$ | Q format of TMP, because AA0 left shift
| by NNGQ_96(IG) bits then round and take
| upper 16 bits
 $NN=(IS-1)*IDIM$ | normalize selected shape
Call $VSCALE(Y(NN+1), IDIM, IDIM, 14, TEMP, NLS)$ | codevector to 16
| bits; put in TEMP
For $K=1,2,\dots,IDIM$, do the next 2 lines | TMP and TEMP both normalized
| to 16 bits, so the product
 $AA0=TMP*TEMP(K)$ | has 1 leading zero.
 $ET(K)=RND(AA0)$ | Directly rounding to high
| work gives us a 15-bit ET
| array.
 $NLSET=NLSTMP+11+NLS-16$ | calculate the NLS for ET.

H.4.2 Additional new gain table

This subclause gives the values for the gain codebook for 9.6 kbit/s operation. The floating-point values are given first. See Table H.4.

Table H.4/G.728 – Floating-point values of gain codebook-related arrays

Array index	1	2	3	4
GQ_96	0.564657	1.937714	-0.564657	-1.937714
GB_96	1.007492	*	-1.007492	*
G2_96	1.129314	3.875428	-1.129314	-3.875428
GSQ_96	0.318838	3.754736	0.318838	3.754736

The fixed-point values are given next. See Table H.5.

Table H.5/G.728 – Fixed-point values of the gain codebook-related arrays

Array index	1	2	3	4
GQ_96(Q13)	4 626	15 874	-4 626	-15 874
GB_96(Q13)	8 253	*	-8 253	*
G2_96(Q12)	4 626	15 874	-4 626	-15 874
GSQ_96(Q11)	653	7 690	653	7 690
NNGQ_96(Q0)	3	2	3	2

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H.4.3 Change of coder parameter

This subclause contains the new parameter, NG_96. This parameter has been changed to NG (value = 8) of the G.728 for the case of 9.6 kbit/s operation. See Table H.6.

Table H.6/G.728 – Basic coder parameters of LD-CELP

Name	Value	Description
NG_96	4	Gain codebook size (number of gain values)

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