

Waveguide modes for Terabit transmission on ordinary wiring

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Drivers for Higher-Speed DSLs

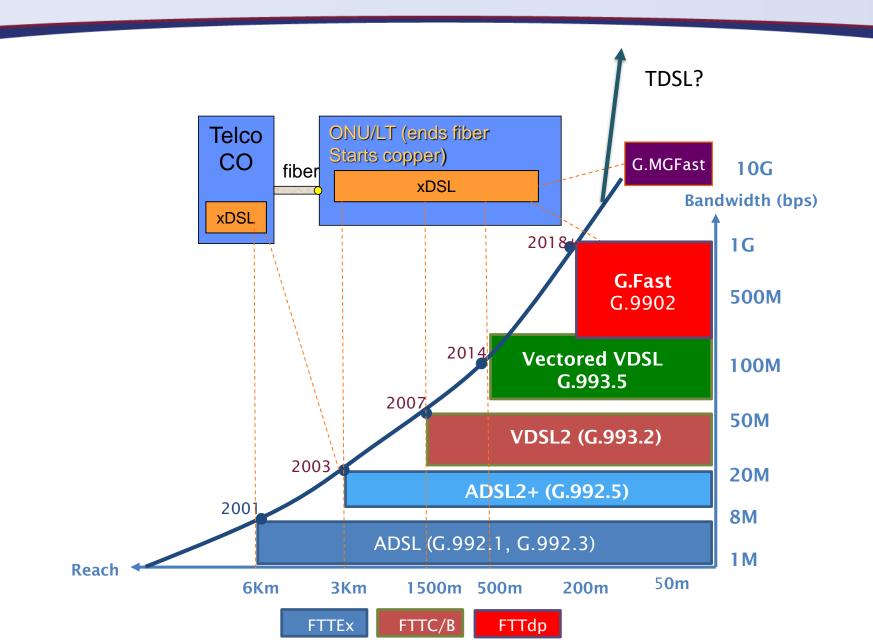
- MULTITUDE of 5G smaller cells
 - high-speed low-latency wired support
 - New 5G-fiber cost = 400B euros (for europe, DT CTO, 2016)
- Fiber theoretical capacity ~ 500 Tbps
 - Today supports 1 Gbps to 100 Gbps (access-network)
- BUT
 - FTTP install costs \$2000-\$3000/home (average)
- The copper twisted pairs are there (1.3B)
 - Run fiber part of the way (\$3000/10 homes is a better business case)
 - Continues x in xDSL, so can x=T?



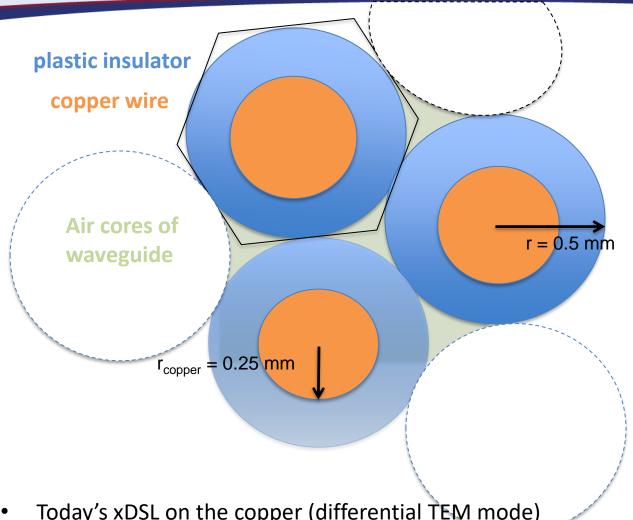
Yes, we think so



Current xDSL progression



Single Pair: Cable Cross Section





And/or the air gaps? (green)



Surface Wave Transmission

(1909 Sommerfeld wave)





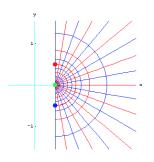
THE G-LINE

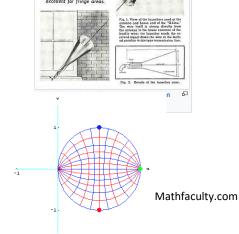
ANTENNA

LEAD-IN

featuring low losses over long lengths at u.h.f. is

- Surface Mode (or TM10)
 - Waves use single wire in TM mode as guide
 - · E.g. Goubau antenna or "G-line"
 - See also AT&T "AirGig"
 - Effectively wireless transmission
 - Works reasonably well (no atmosphere inside cable)
 - Dielectric (plastic) can help (see [Wiltske]), p. 971) keep energy close
 - Tube with non-uniform dilectric constant
 - Conformal mapping of 1/r dimension
- Energy still leaks off wire if bent









Surface-Wave Measurements 2006

Single wire TM01

- Wiltse's surface-wave measurements are 2mm wire core, not 0.5mm)
- Measures attenuation/m

Wiltse Extrapolation

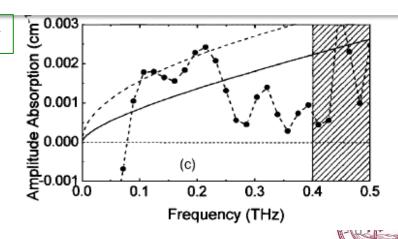
- .8 dB/m @ .1-.3 THz
- Fatter wires
- Grischkowsky has .5 db/m
 - For .52mm diameter Cu wire
 - 2nd wire would probably improve transfer
 - Like in twisted pair
- 100m should see 50-80 dB
- Bending is less of a problem
 - Each wire has a TM mode
 - Between wires is a TEM plasmon polariton mode
 - 2nd TEM "plasmonic" (weaker?) to other pairs somewhat like phantoms/split-pairs
 - TIR mode
 - Surface mode (maybe same as TM ...)
 - 3 -- 4 modes per pair



Table II Attenuation of Single-wire Transmission Lines

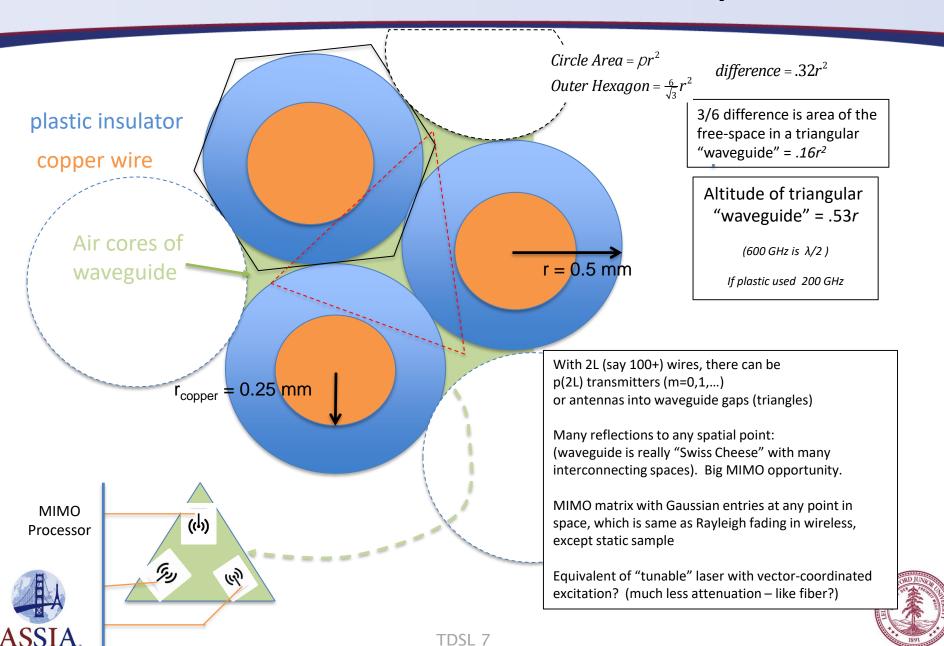
1	147	-		
	Wire	Frequency	Calculated	Measured
	Diameter	GHz	Attenuation	Attenuation
	mm		dB/m	dB/m
	2.032	105	0.23	0.46
	3.251	105	0.16	0.33

Grischkowsky



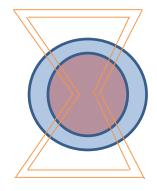


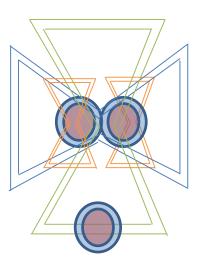
Cross Section Geometry



Vectoring = Massive MIMO

- Lets try m=1 with TM antenna(s) wirelessly exciting each wire end
 - Photoconductive antennas perhaps?
 - Both polarizations (TM and TE) for each wire
 - Or possibly for pairs of wires
- There is also a TEM plasmon polariton mode
 - At least one, really two
 - Could think of this as dual polarization, but not quite really
 - There is also at least one TIR mode (total internal reflection) with sheath
- Nominally intersections would introduce crosstalk between the TMs and TEMs
 - Use MIMO or MISO (just like in mmW wireless 5G, except mmW/10)
 - Will tend toward log normal
- "Swiss Cheese" Waveguide
 - ULTRA rich scattering (exactly what massive MIMO needs)
- Coupling (splicing) is open to innovations, but photoconductive and other types of antennas/lasers/detectors do exist in these frequency ranges today.

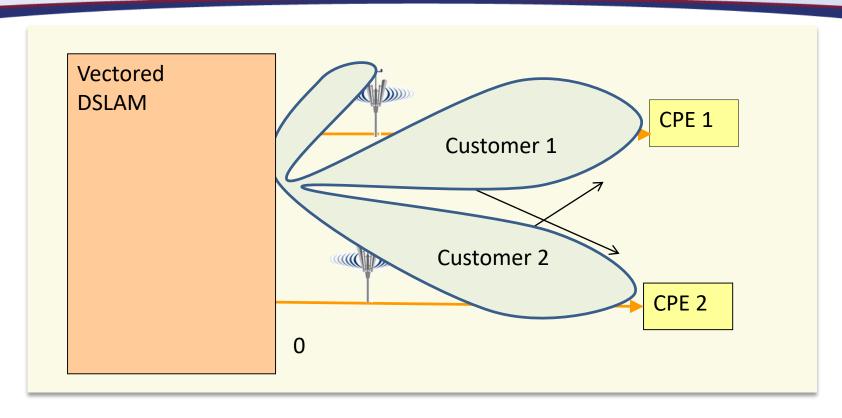








Vectoring ~ Channel Hardening



- Say from 100 GHz to 300 GHz
 - Use 4096 tones, so roughly 50 MHz wide each
 - Two wires in a pair, and two polarizations
- Its conceivable that even 2.5 bits/ tone average, so 1 Tbps





Model

Channel (Grischkowsky)

$$H(f) = e^{-0.05 \cdot \left(\frac{f}{10^{11}}\right) \cdot d}$$

$$X(f) = 10^{k/10} \cdot e^{-0.05 \cdot \left(\frac{f}{10^{11}}\right) \cdot d}$$

- Xtalk (this paper)
 - Log normal

$$X(f) = 10^{k/10} \cdot e^{-0.05 \cdot \left(\frac{f}{10^{11}}\right) \cdot d}$$

Mean k=0 dB Std. Dev. = 6 dB

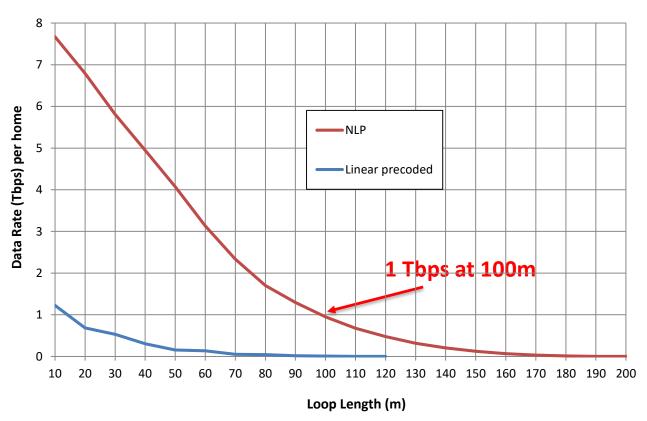
- 20 dBm total transmit power, flat transmit PSD
- 4096 subcarriers from 100 GHz to 300 GHz, 48.8 MHz subcarrier spacing
 - Bit loading from 1 to 12 bits/Hz
 - 10% phy-layer overhead removed before presenting results
 - 6.0 dB coding gain, 1.5 dB implementation loss
 - Carriers from 50 GHz to 150 GHz were used for the 10 Gbps results
- 50 pairs, vector precoded with either zero-forcing linear precoder or Non-Linear Precoder (NLP) using Generalized Decision Feedback Equalization (GDFE); ideal channel estimation assumed.
- -160 dBm/Hz background AWGN.
- We also add in a TM2 and TEM2 mode for 400 GHz to 500 GHz (same parameters)





Results in Tbps [down+up]/pair

100-300 GHz TDSL, per Home data rates



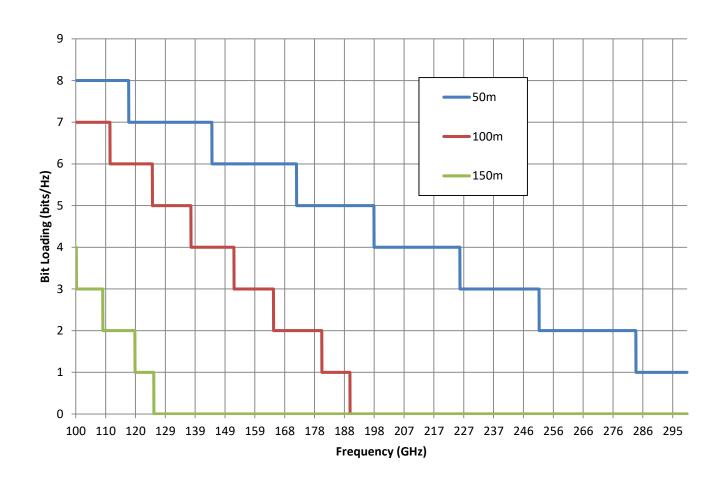
Can any PON get 1 Tbps to each customer?



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Bit Loading

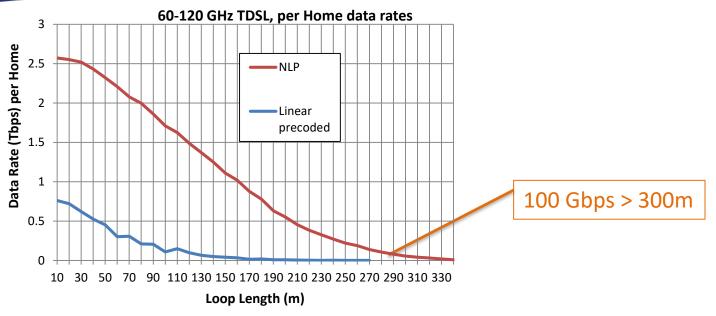
(each polarization of 1 wire)

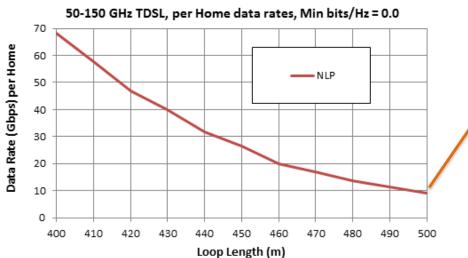






Longer Range, Lower Speed?







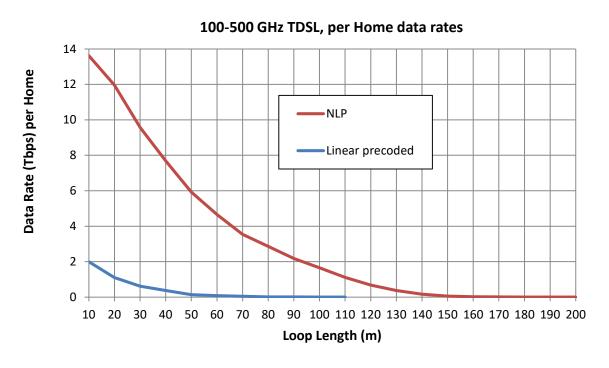


10 Gbps > 500m

 $(^{\sim}0.5km)$

Very-high speed TDSL

Adding TE2 and TEM2 modes from 400-500 GHz



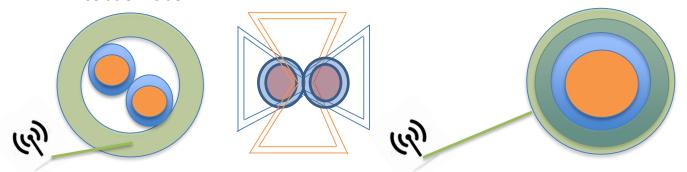
100 meters	300 meters	500 meters
2 Tbps	100 Gbps	10 Gbps





Antennas (analog processing)?

- What would antennas look like?
 - Annular rings around each wire end
 - Also at CPE side



- Possibly multiple co-centric rings at CPE side
- Combinations
- Catch as much drifting energy on CPE receiver as possible (dual for upstream transmitter)
- What would coupling to waveguides look like? (photoconductive, photodetect)
 - It may be feasible to have on die a coupler in this 200-400 GHz frequency range.
 - · Coupling losses?
- Have not included "nested MIMO" over the 4 (or more) antennas per home in results
 - This will be a large improvement (like vector-bonding in multi-line DSL, but perhaps better)
 - Current plots ignore this improvement
 - However, we were optimistic on the energy loss after the sheath-break on the surface waves
 - The two effects may offset





Digital Signal Processing?

- Conversion devices?
 - Might not use all 200 GHz, but still ...
 - ADC's running at 120 Gsamples/second exist (Jarittech)
 - Use Multicarrier (AMT instead of DMT)?
 - Each tone could have its own ADC/DAC (so easily available, but many in parallel)
- Processing Capabilities
 - Vector Engine, even at per tone of 50 MHz
 - .1-.25 Giga-ops per tone
 - Tera-ops for a full system
 - Current Nvidia Tesla GP100 has 5 Teraflops in 16nm CMOS
 - 4-7 nm on immediate horizon and should allow cost reduction
 - Within emerging capabilities
 - Start at 100 Gbps instead (1/10 the cost)?





Opportunities – Measurements

- Real <u>cable</u> measurements would help
 - These modes certainly exist, but what exactly is attenuation?
 - Best/reasonable antenna/interface designs?
 - Real-world impairments: bends, splices, taps, etc.
- How good is the log-normal model for waveguide modes' xtalk?
 - Might this xtalk be larger?
- Even 10% of projected data rates >> "G.mgfast"

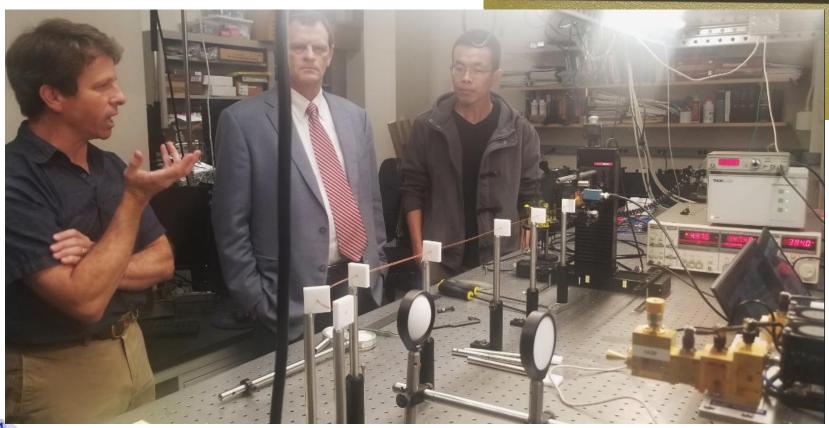




Brown University – current NSF work

Professor Dan Mittleman









What about yet-higher-order modes?

- They exist!
- Higher bandwidths, but attenuation?
 - Unknown for now
 - Likely need even more antennas/wire (MIMO)
 - 10 meters (instead of 100m) might work
 - Not clear if waves could be focused like surface waves by MIMO processing to "hug the wires" as they separate and go to individual homes
 - Grounded shield would contain them though
 - PDSL? (P=Petabit or 10^{15})
- TDSL will probably be enough for now





Conclusions

- TDSL is technically feasible with 100 pairs and phantoms used for backhaul
 - Also roughly 1 Tbps @100m, 100Gbps at 300m, 10 Gbps at 500m
 - But of course on ALL 100 pairs used together
 - Could be very useful for 5G cell multitude
- Terabit/s DSL per home (or small cell) also appears feasible
 - Using waveguide modes and vectoring SINGLE pair
 - Measurements of attenuation would help refine rate/range
 - Probably with MIMO-channel characterization used
 - Could be expensive to prototype
 - Because of processing/converter speeds involved
- Is it worth it? (or should we spend \$4 Trillion to replace all the copper with fiber instead, say in the next 3-5 years Or century?)
 - Would 5G small cells be accelerated since this would reduce deployment cost?



Acknowledgements and Refs

- Prof. Joe Kahn, Stanford U.
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- Dr. S. Galli, Huawei
- Prof. Leonid Kazovsky, Stanford U.
- J.C. Wiltse, "Surface-Wave Propagation on a Single Metal Wire or Rod at Millimeter-Wave and Terahertz Frequencies," <u>Microwave Symposium Digest, 2006. IEEE MTT-S International</u>, 11-16 June 2006.
- R.E. Collin, "Hertzian Dipole Radiating Over a Lossy Earth or Sea: Some Early and Late 20 th-Century Controversies" IEEE Ant and Prp. Magazine, Vol 46, No. 2, April 2004.
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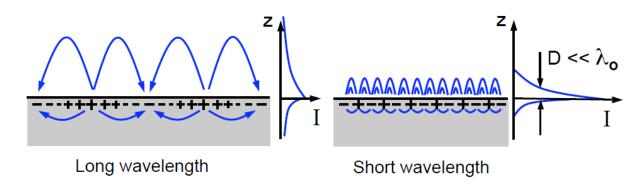


Back UP





Bare metal wire waveguide



Surface plasmon polariton

- EM surface wave that travels along an interface between metal (negative permittivity) and dielectric (positive permittivity) based on surface electron density changes below metal's plasma frequency
 - Phase velocity and group velocity is same (like free space) → no dispersion if frequency is way below plasma frequency
- E-field decays exponentially vertical to the wire
 - → energy is confied near the conductor so no 1/r type of path loss. Only small ohmic loss due to electron scattering → small in materials with high conductivity and high frequency
 - \rightarrow Loss about 0.1%~0.25% of field strength in 1cm \rightarrow 0.86dB/m ~ 2.1dB/m @ 0.25THz
- Problems
 - TM mode → Hard to generate radially polarized EM wave & low coupling coefficient
 - Need to be straight → lose energy due to bending
 - Connecting two metal wires are not easy





Splicer

