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Radiocommunication Sector of ITU

**Report ITU-R BT.2215-4**  
(11/2014)

**Measurements of protection ratios  
and overload thresholds  
for broadcast TV receivers**

**BT Series**  
**Broadcasting service**  
**(television)**

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## Foreword

The role of the Radiocommunication Sector is to ensure the rational, equitable, efficient and economical use of the radio-frequency spectrum by all radiocommunication services, including satellite services, and carry out studies without limit of frequency range on the basis of which Recommendations are adopted.

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## REPORT ITU-R BT.2215-4

**Measurements of protection ratios and overload thresholds  
for broadcast TV receivers**

(2011-2011-2012-2012-2014)

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## 1 Executive summary

This Report documents measurements of protection ratio (PR) and overload threshold ( $O_{th}$ ) against interference from other broadcasts or mobile broadband services in the 800 MHz band.

The types of interference used in the tests and the actual tests themselves varies with the different broadcasting systems and mobile technologies used around the world.

The aim of the Report is to establish test procedures together with measurement results to assist in network planning and sharing studies for the co-existence of TV broadcasting, with either mobile services, or other services and applications.

The original test details and measurement data, together with information on which particular Recommendation and its version number was updated with this data are contained in the following annexes:

Annex 1 – DVB-T receiver performance in the presence of interfering signals from DVB-T, UMTS, and LTE

Annex 2 – DVB-T2 receiver performance in the presence of interfering signals from DVB-T2 and LTE

Annex 3 – ATSC receiver performance in the presence of interfering signals from ATSC

## 2 Abbreviations

ACLR	Adjacent channel leakage ratio
ACS	Adjacent channel selectivity
AGC	Automatic gain control
BER	Bit error rate
BS	Base station
DTT	Digital terrestrial television
ESR <sub>5</sub>	Erroneous-Second Ratio with one (1) out of 20 seconds (5%) with errors
LTE	Long Term Evolution, 4 <sup>th</sup> generation mobile standard
MPEG	Motion Picture Expert Group
OFDMA	Orthogonal frequency-division multiple access – a multi-carrier modulation system used for the long term evolution (LTE) downlink.
$O_{th}$	Overload threshold
PR	Protection ratio
PSD	Power spectral density
QEF	Quasi-error-free
RB	Resource block – a unit of data transmission in LTE, represented by a certain number of carriers in an uplink or downlink symbol in the frequency domain.
SC-FDMA	Single carrier frequency division multiple access – a multi-carrier modulation system used for the LTE uplink.

SFP	Subjective failure point
TPC	Transmit power control
UE	User equipment – the mobile handset
UMTS	Universal Telecommunications System, 3 <sup>rd</sup> generation mobile standard

### 3 Useful definitions

#### 3.1 Radio frequency signal-to-interference ratio (*C/I*)

It is the ratio, generally expressed in dB, of the power of the wanted signal to the total power of interfering signals and noise, evaluated at the receiver input (see Recommendation ITU-R V.573).

The power of the wanted signal is measured in a bandwidth equal to the wanted signal bandwidth, while the total power of interfering signal and noise is measured in a bandwidth equal to the interfering signal bandwidth.

#### 3.2 Radio frequency protection ratio (PR)

It is the minimum value of the signal-to-interference ratio required to obtain a specified reception quality under specified conditions at the receiver input (note that this differs from the definition in Recommendation ITU-R V.573). In this Report, the “specified reception quality” and the “specified conditions” have been defined separately by each entity that has undertaken measurements.

Usually, PR is specified as a function of the frequency offset between the wanted and interfering signals over a wide frequency range. In this Report, PR specified in this way is referred to as “PR curve”. PR curves show the ability of a receiver to discriminate against interfering signals on frequencies differing from that of the wanted signal.

#### 3.3 Receiver (front-end) overload threshold

Overload threshold ( $O_{th}$ ) is the interfering signal level expressed in dBm, above which the receiver begins to lose its ability to discriminate against interfering signals at frequencies differing from that of the wanted signal (i.e., the onset of strong non-linear behaviour). Therefore, above the overload threshold the receiver will behave in a non-linear way, but does not necessarily fail immediately depending on the receiver and interference characteristics.

#### 3.4 Adjacent channel leakage power ratio

Adjacent channel leakage power ratio (ACLR) is the ratio of the filtered mean power centred on the assigned channel frequency to the filtered mean power centred on an adjacent channel frequency.

The requirements shall apply whatever the type of transmitter considered (single carrier or multi-carrier). It applies for all transmission modes foreseen by the manufacturer's specification. For a multi-carrier base station (BS), the requirement applies for the adjacent channel frequencies below the lowest carrier frequency transmitted by the BS and above the highest carrier frequency transmitted by the BS for each supported multi-carrier transmission configuration. The requirement applies during the transmitter ON period.

### 3.5 “Can” tuners

“Can” tuners are classical superheterodyne tuners housed in a metal enclosure containing discrete components. Classically, there are fixed and tunable circuits made up from discrete inductors and transistors usually with varactor diode frequency control. The metal enclosure should minimize RF interference and eliminate crosstalk and stray radiation.

### 3.6 “Silicon” tuners

“Silicon” tuners are IC-based tuners integrating all tuner circuitry into a small package directly to be fitted onto main boards. The tuned circuits may be completely absent or can be integrated onto the silicon. The silicon chip may be protected from external electromagnetic interference by a metallic cover. Silicon tuners have different characteristics to can tuners and their performance can be better and worse at some frequency offsets compared to can tuners.

## 4 References

### 4.1 Broadcasting technology characteristics

The following references explain the characteristics of the different broadcast systems including transmitter spectrum masks.

- DVB-T system characteristics: Recommendation ITU-R BT.1306, ETSI EN 300 744
- DVB-T2 system characteristics: Recommendation ITU-R BT.1877, ETSI EN 302 755
- ISDB-T system characteristics: Recommendation ITU-R BT.1306, ARIB STD-B31
- ATSC system characteristics: Recommendation ITU-R BT.1306, ATSC A/53
- DTMB system characteristics: Recommendation ITU-R BT.1306, GB20600-2006.

### 4.2 Mobile technology characteristics

The following references explain the characteristics of the different mobile broadband systems.

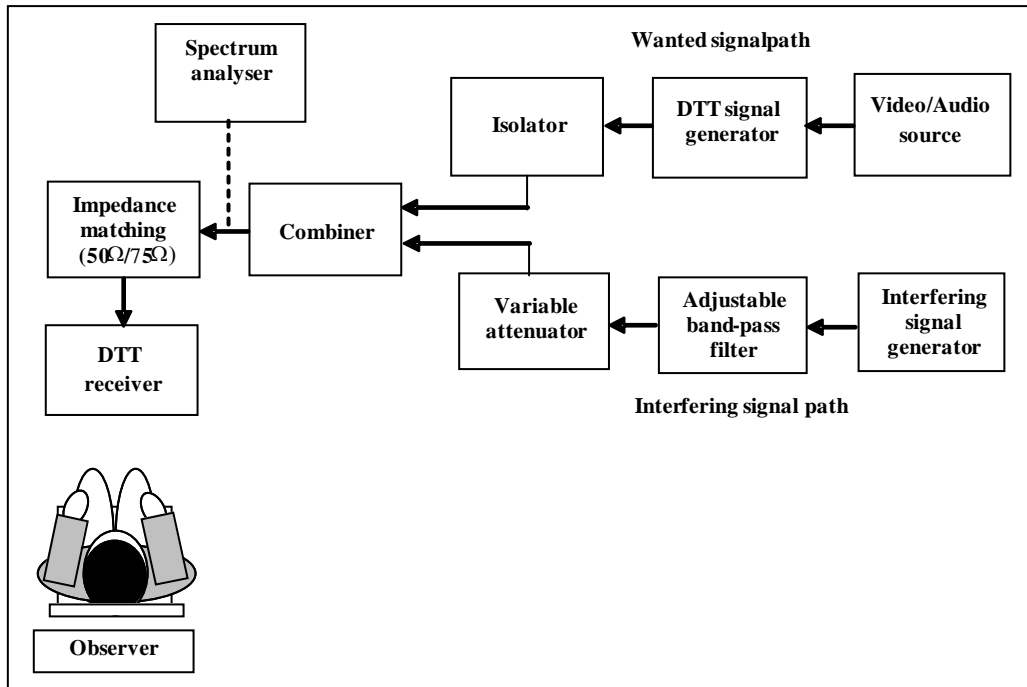
- UMTS system characteristics: ETSI TS 125.101, ETSI TS.125.104
- LTE system characteristics: ETSI TS 136.101, ETSI TS.136.104.

## 5 Measurement methodology

### 5.1 Example test setup

An example basic test setup for protection ratio and overload threshold measurements for digital terrestrial television (DTT) is depicted in Fig. 1.

FIGURE 1  
Example test setup

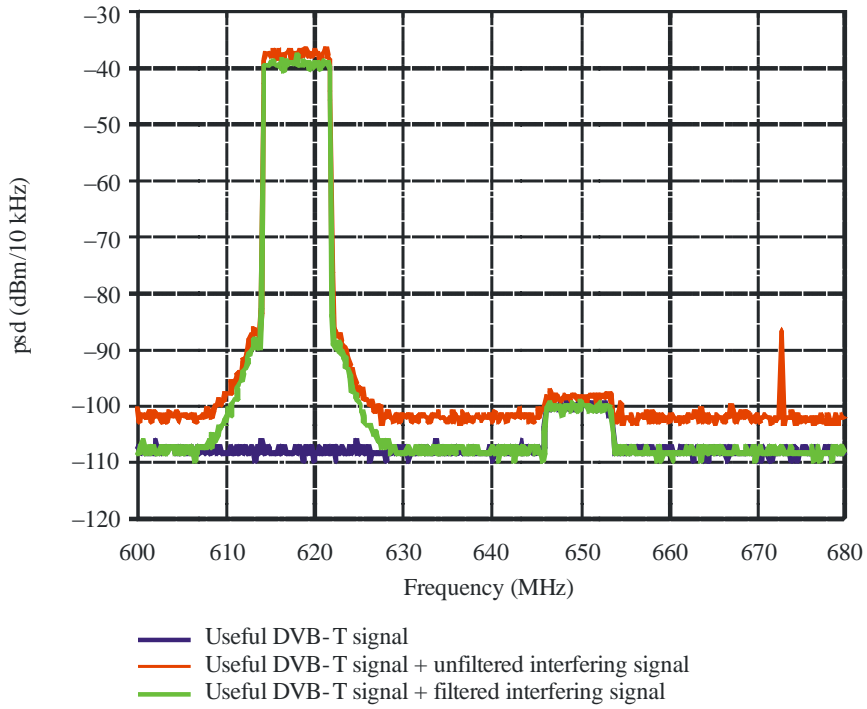


It is necessary to insert an adjustable band-pass filter between the interfering signal generator and the combiner. The objective of this filter is to eliminate the noise generated by the interfering signal generator and adjust the interfering signal to the correct interference transmission mask and ACLR values. In fact, most of the RF signal generators have a wide frequency range (from several hundred of kHz to several GHz) prohibiting the use of an internal adjustable RF channel filter over their whole frequency range. Consequently, depending on the generated signal level, a non-negligible wideband noise may be observed at the generator output. The higher the generated interfering signal level, the higher the noise level. The reduction of the undesired wideband noise by filtering at the output of interfering signal generator is shown as an example in Fig. 2. If this noise is not reduced by filtering, it is impossible to measure the actual protection ratios of the receiver under test. This is due to the wideband noise generated by the interfering signal generator, falling into the wanted signal channel, which cannot be reduced by the receiver filter. In this particular case, the receiver loses its ability to discriminate against interfering signals on frequencies differing from that it is tuned to. This phenomenon is shown in Fig. 3. It is also advisable to insert an isolator between the combiner and the DTT signal generator to keep the power from the interfering signal generator returning to the DTT signal generator output.



FIGURE 2

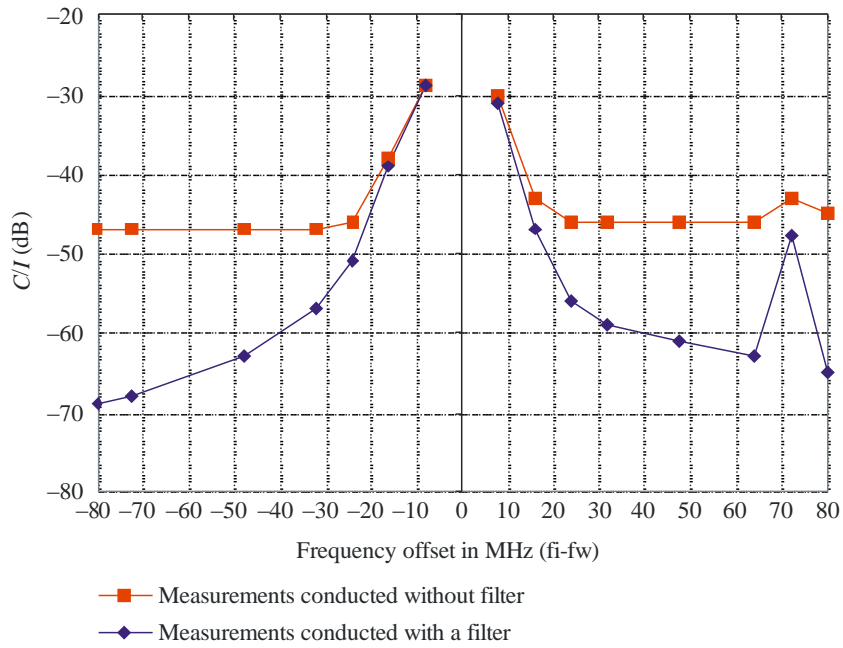
The benefit of band-pass filtering at the interfering signal generator output



Report BT.2215-02

FIGURE 3

The benefit of band-pass filtering at the output of the interfering signal generator (wanted signal level = Rx sensitivity +10 dB)



Report BT.2215-03

## 5.2 Wanted signal levels

Protection ratios and overload thresholds of a receiver are derived from its  $C(I)$  curves (see § 5.11). The measurements should be carried out by using different wanted signal levels to cover the range from weakest to strongest signals. The following wanted signal levels relative to the receiver sensitivity are advised as a possible range: receiver sensitivity +5, +10, +20, +30, +40, +50, +60, +70 and +80 dB. This range could be extended if the overload threshold of the receiver is not reached. At low wanted signal levels the protection ratio limit is usually reached before the overload threshold. Therefore it is necessary to use higher wanted signal levels to reach the onset of overload.

$$\Delta P = 10 \log (n_c - 1)$$

## 5.3 Frequency offsets between interfering signal and wanted signal

It is usual to use the following frequency offsets:

0,  $\pm N$ ,  $\pm(N+BWI)$ ,  $\pm(N+2 BWI)$ ,  $\pm(N+3 BWI)$ ,  $\pm(N+4 BWI)$ , ...

and 9 BWW (DVB-T/T2 image channel),

or  $\pm 14$  BWW and  $\pm 15$  BWW (ATSC image channels).

Where:

$$N : (BW_w + BW_i)/2$$

$BW_w$ : wanted signal bandwidth

$BW_i$ : interfering signal bandwidth

However, regional specific frequency offsets could also be used, and smaller steps where more detailed investigation is required.

## 5.4 Measurements in the presence of a time varying interfering signal

An important difference between existing interference by other broadcast signals, and mobile signals is that in many cases the mobile signal power can exhibit significant time variation which can degrade the PR and  $O_{th}$  performance of some DTT receivers due to interfering with automatic gain control (AGC) and channel estimation algorithms. It is important to test against such types of interference. Time variation occurs in (at least) the following circumstances.

## 5.5 UMTS uplink

The UE can use transmit power control (TPC) to improve performance in mobile reception conditions where the channel can be rapidly changing. The effect of this is for the UE to vary its transmit power rapidly over time in response to feedback messages from the BS.

## 5.6 LTE – downlink

The base station output power can vary over time if only some resource blocks (RB) are used in each OFDMA symbol, or if some OFDMA symbols are completely empty. This tends to happen when the BS traffic loading is zero or at low levels. Consequently, in the presence of a BS interfering signal, it is recommended to carry out the measurements with different network traffic loadings of 0% (idle), 50% and 100%.

## 5.7 LTE – uplink

The uplink signal can vary considerably in both the time and frequency domains depending upon the traffic loading required. In the frequency domain the number of RBs allocated for each SC-FDMA symbol can vary rapidly. In the time domain, there can be long periods where the UE does not transmit at all, leading to an irregular pulse like power profile.

Consequently, in the presence of a UE interfering signal, it is recommended to carry out the measurements with different data rates on the uplink. The modes should include both fully loaded continuous operation and time division multiplexed i.e., pulsed operation.

## 5.8 Interferer reference power level

Signal level variation can be from level reductions or time division occupancy. In order to be able to see the degradations caused by time variation in the interfering signal, it is necessary to set the appropriate rms power or power spectral density (psd) of the active portions of the time varying interference signal relative to the rms power or psd of the interferer with a 100% traffic loading (time invariant power condition).

## 5.9 Characterization of the interfering signal

Protection ratios and overload thresholds of a receiver strongly depend on the frequency and time domain characteristics of the interfering signal used in the measurements. Therefore, it is necessary to record the psd, the adjacent channel leakage power ratio as well as the amplitude as function of time of the interfering signal. These pieces of information allow comparisons of different measurement results from different measurement campaigns.

## 5.10 Failure point assessment methods

Initial studies of the protection ratios for the DVB-T system were based on a target bit error rate (BER) of  $2 \times 10^{-4}$  measured between the inner and outer codes, before Reed-Solomon decoding. For the case of a noise-like interferer, this has been taken to correspond to a quasi-error-free (QEF) picture quality with the BER  $< 1 \times 10^{-11}$  at the input of the MPEG-2 demultiplexer.

For domestic receivers it may not be possible to measure the BER and therefore a new method called the subjective failure point (SFP) method has been proposed in Recommendation ITU-R BT.1368 for protection ratio measurements in a unified manner. The quality criterion for protection ratio measurements is to find a limit for a just error-free picture at the TV screen. The RF protection ratio for the wanted DVB-T signal is a value of wanted-to-unwanted signal ratio at the receiver input, determined by the SFP method, and rounded to the next higher integer value.

The SFP method corresponds to the picture quality where no more than one error is visible in the picture for an average observation time of 20 s. The adjustment of the wanted and unwanted signal levels for the SFP method is to be carried out in small steps, usually in steps of 0.1 dB. For a “noise-like” interferer the difference in a value of wanted-to-unwanted signal ratio between the QEF method with a BER of  $2 \times 10^{-4}$  and the SFP method is less than 1 dB. It is proposed that the SFP method should be adopted for assessment of all DTT systems.

### 5.10.1 ESR<sub>5</sub> Criterion for the Assessment of DTT Transmission Quality

The “erroneous-second ratio” or ESR<sub>5</sub> criterion is used to assess the DTT (e.g., DVB-T) transmission quality at portable and mobile reception. It is used instead of the “quasi error-free” (QEF) or the “subjective failure point” (SFP) criterion.

The bit error rate (BER) after the Viterbi decoder given in the DVB-T standard for quasi error-free DVB-T reception is  $2 \cdot 10^{-4}$ . This transmission quality assessment criterion is suitable for fixed reception using a roof antenna, where the transmission channel transmitter – receiver is of type Gauss or Rice. The bit error rate is quite constant with time and can be easily measured.

The QEF criterion is not suitable for portable or mobile reception. The transmission channel is mostly of type Rayleigh (multipath). The BER fluctuation is very large and averaging over a longer time period is necessary. An average BER of  $2 \cdot 10^{-4}$  after the Viterbi decoder is nevertheless no guaranty for an error free reception.

The impairments at DVB-T transmission in Rayleigh channels occur in bursts. Even if the average BER after the Viterbi decoder is well below  $2 \cdot 10^{-4}$  (QEF), the BER peaks during error bursts are so high from time to time, that there are visible errors and also some losses of MPEG-2 transport stream synchronization. Synchronization losses generate serious video errors.

For transmission quality assessments at portable and mobile DVB-T reception other criterion than QEF have to be used. One such alternative criterion used is called “subjective failure point” (SFP). The TV program received is observed on a monitor. The quality of the received signal fulfils the SFP criterion if, in a 20-seconds time interval, there is no visible failure in the picture. The measurement has the drawback, that it cannot be made automatically, and a person has to monitor permanently the picture quality. It is not very accurate and gives different results for different observation times.

A more accurate criterion used to assess the transmission quality at portable and mobile DVB-T reception is the “erroneous second ratio” criterion, also called ESR<sub>5</sub> criterion. Compared to the SFP criterion it has the advantage, that its accomplishment can be checked automatically, without the need for a measuring engineer, and it can therefore be analyzed over a large time span. The errors in a MPEG 2 transport stream, which generate visible failures of the picture, are the packet uncorrectable errors signaled by a flag set by the Viterbi decoder, and the transport stream synchronization loss.

The ESR<sub>5</sub> criterion is fulfilled if, in a time interval of 20 seconds, there is at most one second with packet uncorrectable errors. If there is a transport stream synchronization loss, the ESR<sub>5</sub> criterion is not fulfilled for the corresponding 20 seconds time interval.

The ratio second with errors to all seconds in the 20-seconds time interval is 5 %, explaining the index 5 in ESR<sub>5</sub>.

The ESR<sub>5</sub> and the SFP criteria are somehow equivalent. The time interval for the basic assessment is 20 seconds for both. The smallest time interval for the error assessment at ESR<sub>5</sub> is the second. One or more errors in the transport stream during one second have the same subjective influence on the video quality and are perceived at a SFP measurement as one error. Not all transport-stream errors are producing visible errors, and for this reason, in contrast to SFP, at ESR<sub>5</sub> one erroneous second in a 20 second interval is allowed.

The description of the transmission quality is correct and accurate if the assessment is done for a larger time span, for example 15 to 60 minutes. In this case the measurement time is segmented in 20 seconds intervals, and in every interval the fulfillment of the ESR<sub>5</sub> criterion is tested. The ratio number of 20-second intervals fulfilling the ESR<sub>5</sub> criterion to the number of all 20-second intervals shows an average of the transmission quality. The averaging is important because a Rayleigh transmission channel has big fluctuations in time.

In a Gaussian or Rice channel the receiving quality is poor if the carrier to noise ( $C/N$ ) value is only so high, that the SFP or the ESR<sub>5</sub> criterion is just fulfilled. One possible error each 20-seconds interval is too much. The criteria are nevertheless suitable for quick measurements, and a 1-2 dB carrier power increase enhances the transmission quality considerably. This is due to the low power variation in time in Gaussian and Rice channels.

In a Rayleigh channel the power variation with time is remarkably higher. The DVB-T transmission quality is good, if the ESR<sub>5</sub> criterion is fulfilled for at least 99 % of time. The BER variation is quite high, and if only 1% of the 20-second intervals have more than one erroneous second within, most of the other intervals will have no erroneous second at all within. The assessment of the transmission quality takes now also its time variation into account.

The segmentation of the measurement time in 20 second intervals for the assessment of the compliance with the ESR<sub>5</sub> criterion is an arbitrary process. For short time intervals, the position of the starting point can influence the results, for large time intervals, there is a statistical smoothing of the results.

The ESR<sub>5</sub> criterion is less demanding than the QEF criterion. For the mode 16 QAM 2/3 the signal to noise ratio  $C/N$  in a Gaussian channel for 99% of time good quality transmission in accordance to the ESR<sub>5</sub> criterion is 1.2 dB lower than for the QEF-criterion.

In a Rayleigh channel, especially at mobile reception, the QEF criterion cannot be fulfilled for all the time, even for unrealistically high field strength levels. The only realistic requirement is the compliance to the ESR<sub>5</sub> criterion.

In conclusion, the “erroneous-second ratio”- or ESR<sub>5</sub>-criterion is a suitable criterion to assess the longtime, average DTT transmission quality at portable and mobile reception. It states that, if in a time interval of 20 seconds there is at most one erroneous second, the transmission quality for this 20-seconds interval is good. The average, longtime transmission quality is good, if the ESR<sub>5</sub> criterion is fulfilled for at least 99 % of time.

### 5.11 Method for determining protection ratios and overload thresholds

It should be stressed that the protection ratios are generally considered and used as independent of the wanted signal level. That is  $C(I)$  is supposed to be a linear function with unity slope (a straight line with unity slope). The protection ratio of the receiver is obtained by subtracting  $I$  from  $C(I)$  at any point on this line and can be used for all wanted signal levels. Fig. 4 illustrates the  $C/I$  curve of an ideal receiver in the presence of an interfering signal. The ideal receiver has perfectly linear behavior under all circumstances without overload. Fig. 5 illustrates the  $C(I)$  plots for the ideal receiver described in Fig. 4. The ideal receiver has identical  $C/I$  curves. Consequently, unity slope  $C(I)$  lines show that the receiver protection ratios depend neither on the wanted nor the interfering signal levels, but only  $C/I$ .

FIGURE 4  
**C/I curve of an ideal receiver in the presence of an interfering signal**  
 (Rx sensitivity = -80 dBm)

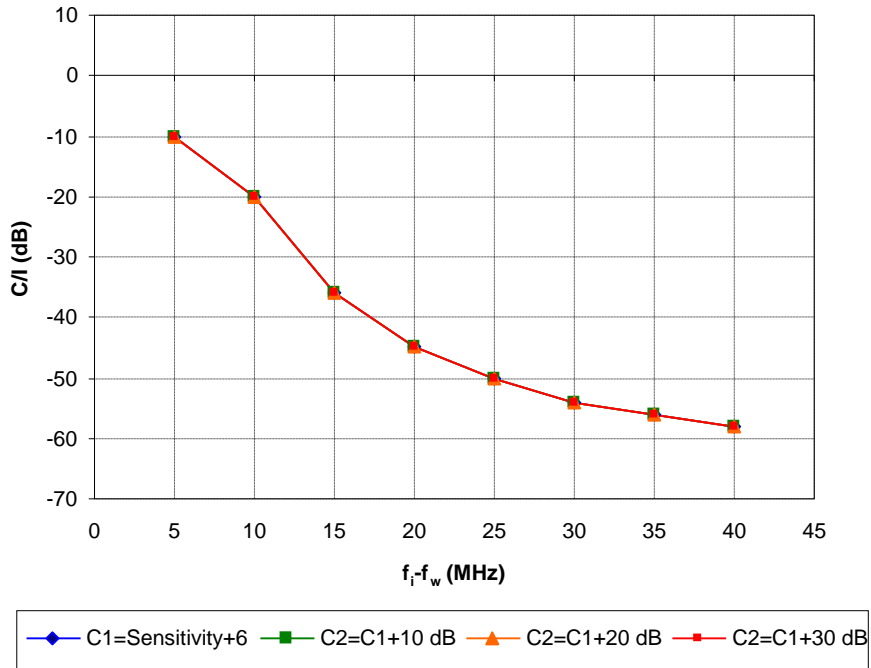
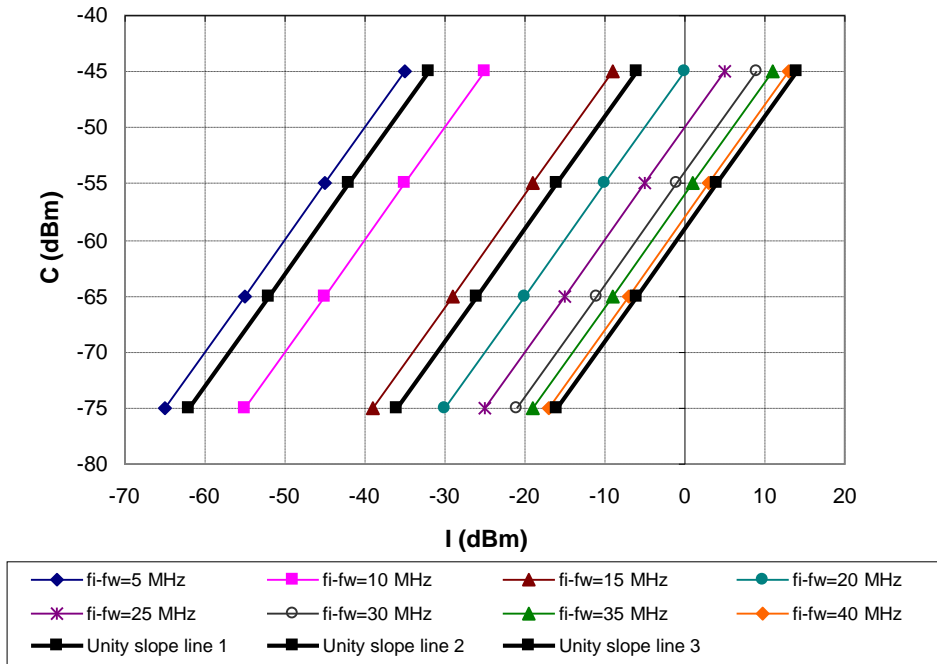


FIGURE 5  
**C(I) plots of an ideal receiver**  
 (Wanted signal levels = -75, -65, -55, and -45 dBm)



However, the measurement results show that in most cases the protection ratios of wideband TV receivers vary as a function of the wanted signal level. Consequently,  $C(I)$  is not a straight line with unity slope with some variation with the interfering signal strength. Fig. 6 illustrates a receiver which

behaves badly in the presence of a strong interfering signal. Fig. 7 illustrates the  $C(I)$  plots from the receiver shown in Fig. 6. Nevertheless, for interfering signals below the overload threshold such  $C(I)$  curves can always be approximated by a straight line with unity slope with an acceptable error. This method has been used in this report for determining the PR of DTT receivers.

FIGURE 6  
 **$C/I$  curves of a receiver behaving badly in the presence of a strong interfering signal  
 (Rx sensitivity = -80 dBm)**

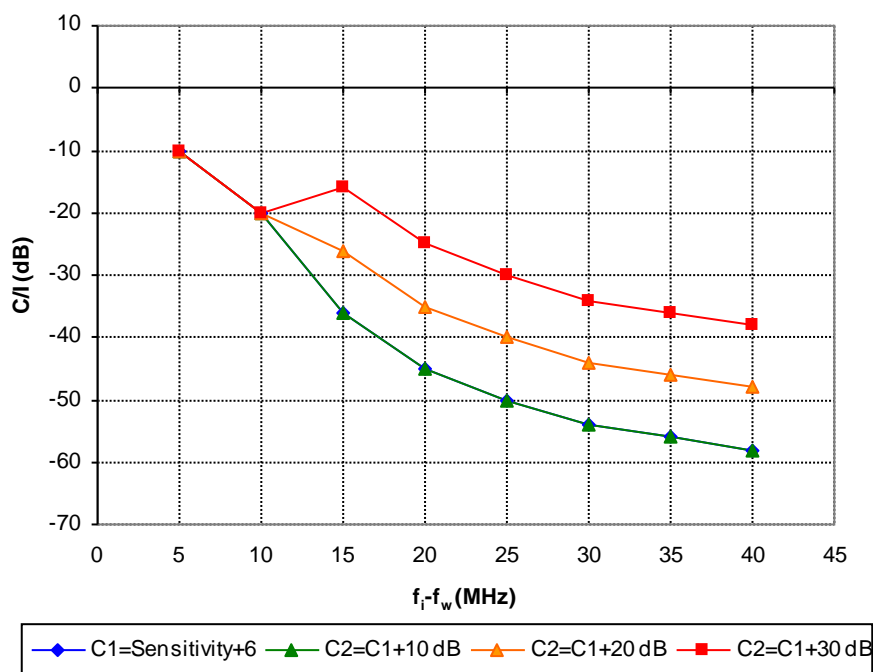
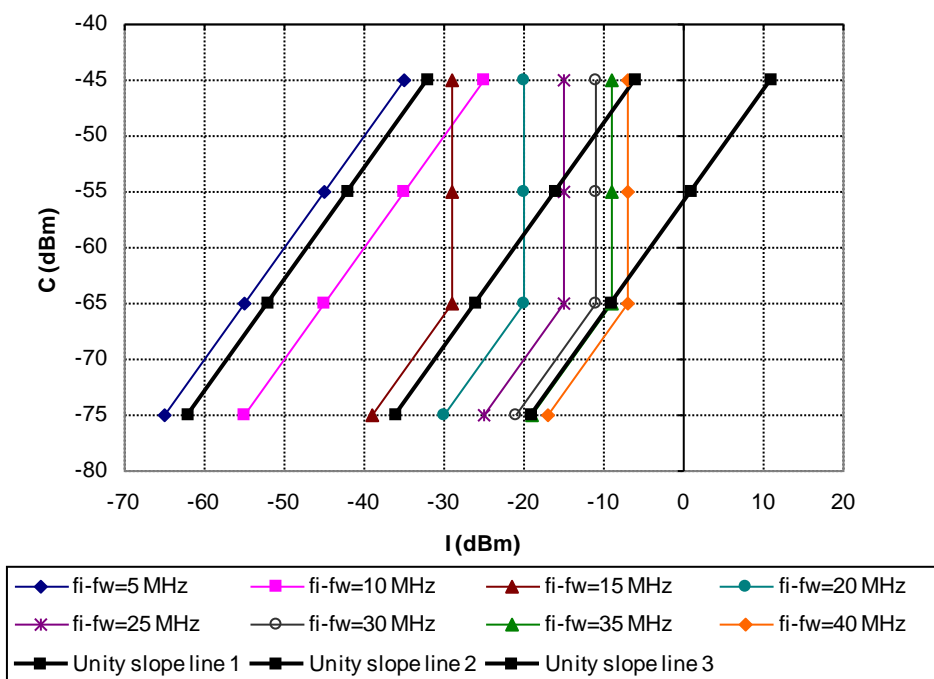


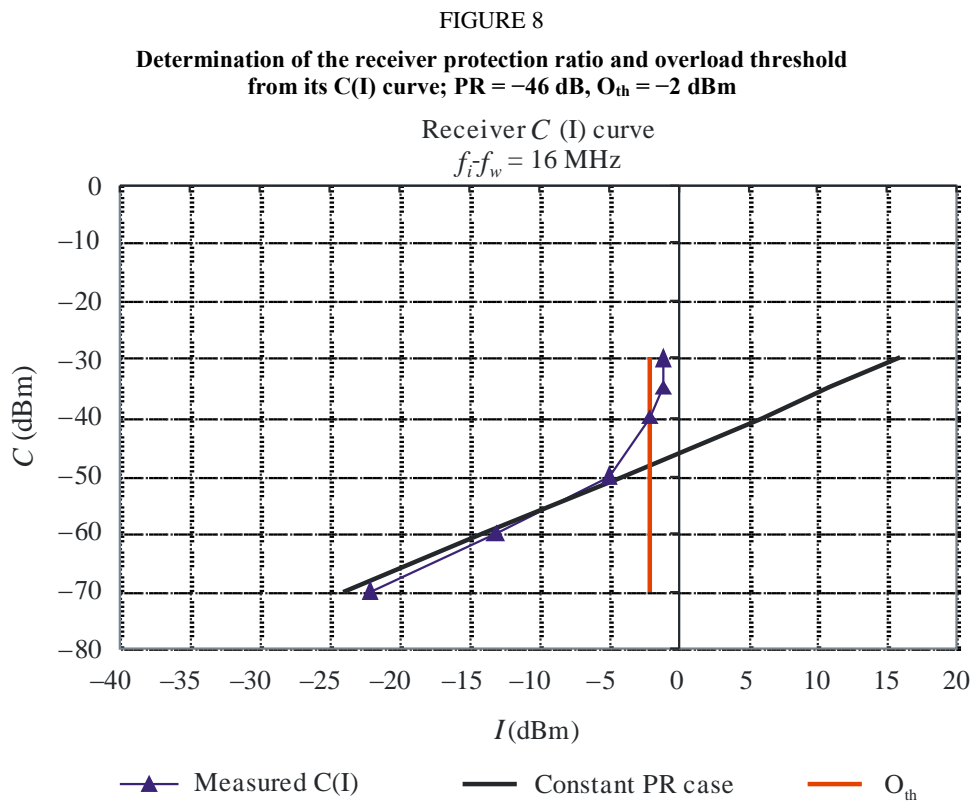
FIGURE 7  
 **$C(I)$  plots of the receiver described in Figure 6 at various wanted signal levels  
 (-75, -65, -55, -45 dBm)**



The method used for determining protection ratios and overload thresholds is composed of two steps:

- 1) The measured  $C(I)$  curve is approximated by a straight line with unity slope which represents the ideal linear behaviour of the receiver front-end (constant PR case). The protection ratio of the receiver is obtained by subtracting  $I$  from  $C(I)$  at any point on this line. The protection ratio obtained can be used for all wanted signal levels.
- 2) A strong deviation of the measured  $C(I)$  curve from the straight line with unity slope indicates where the interfering signal reaches the overload threshold; i.e., the onset of strong non-linear behaviour. The deviated segment of  $C(I)$  curve is approximated by a line vertical to  $I$ -axis (constant  $I$  case). The value of  $I$  at the point of intersection between the straight line with unity slope and the line vertical to  $I$ -axis is considered to be the receiver overload threshold ( $I = O_{th}$ ).

This two steps procedure is depicted in Fig. 8.



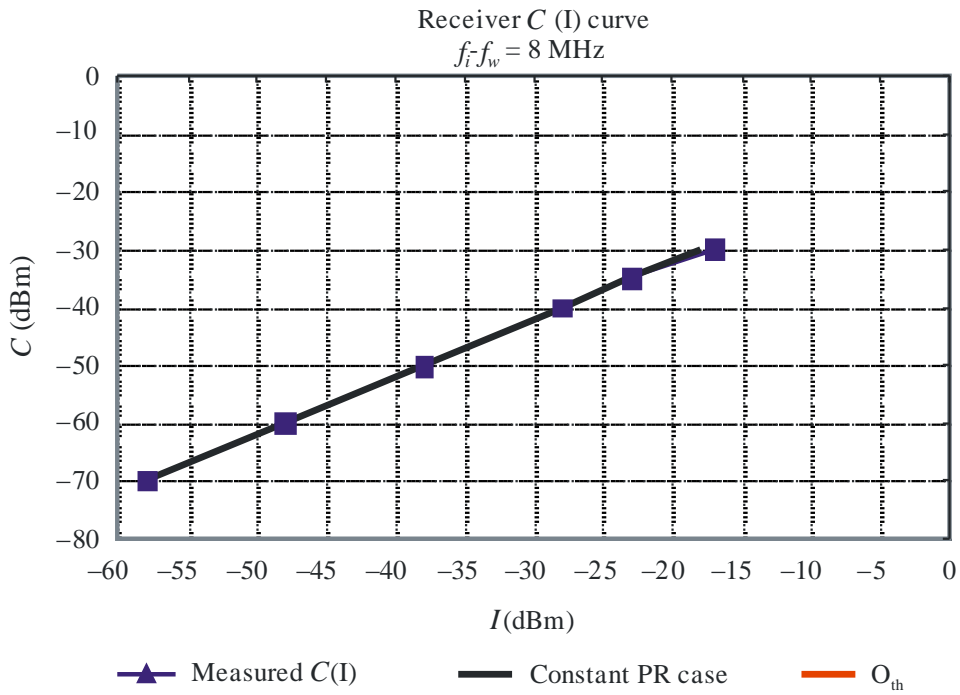
Report BT.2215-04

In some cases the approximation of a measured  $C(I)$  curve by a straight line with unity slope and a line vertical to  $I$ -axis may seem not to be very straight forward, but it is always possible to do it with an acceptable approximation error that should be in favour of the victim receiver.

Examples of approximations are shown in Figs 9 to 16. These examples use a wanted signal level range starting at  $-70$  dBm, but lower levels are possible depending upon the sensitivity of the receiver mode being tested.

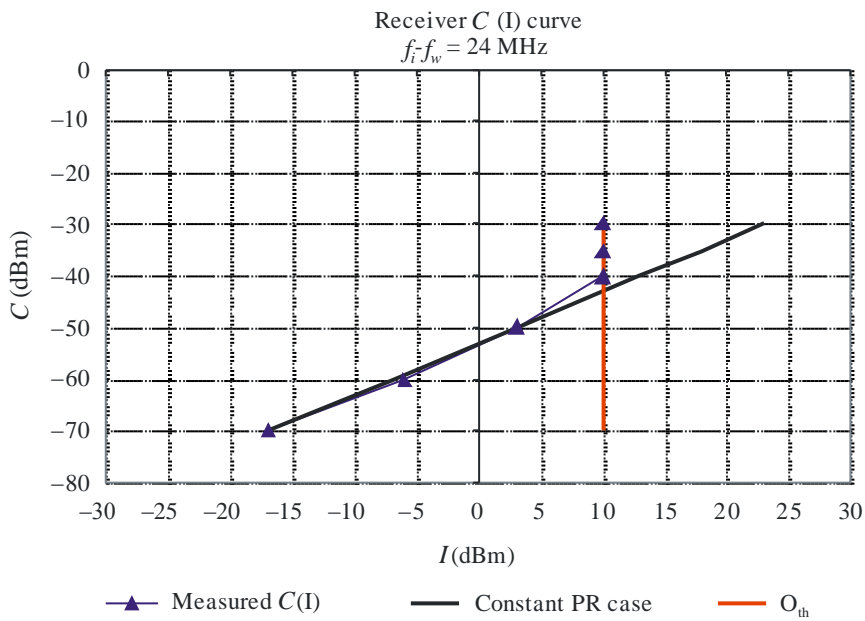


FIGURE 9  
 A well approximated C(I) curve; PR = -12 dB; overload  
 threshold is not reached



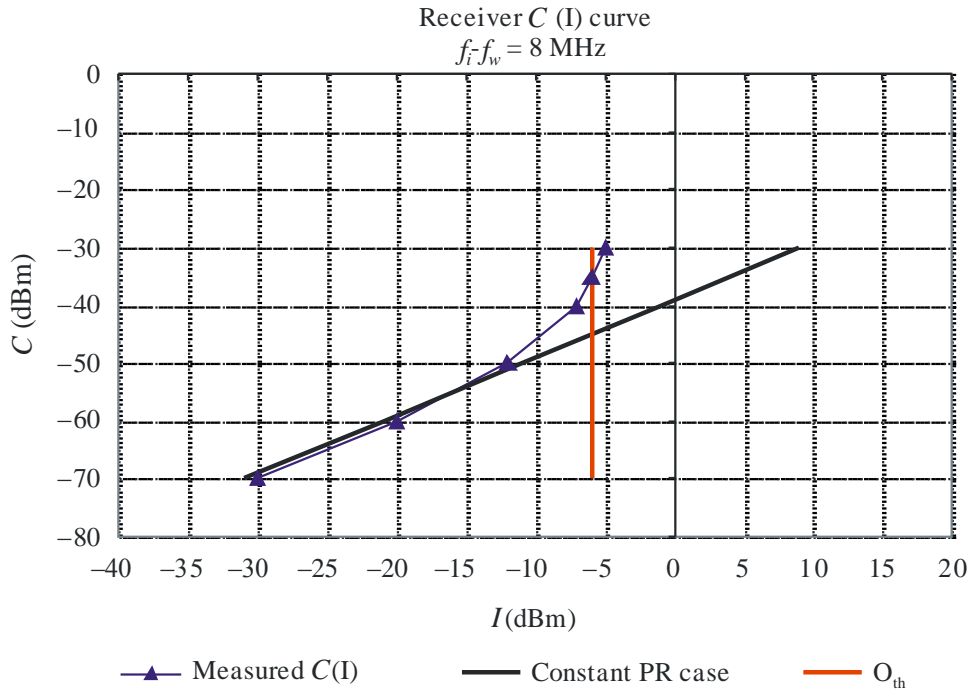
Report BT.2215-05

FIGURE 10  
 A well approximated C(I) curve; PR = -53 dB; O<sub>th</sub> = 10 dBm



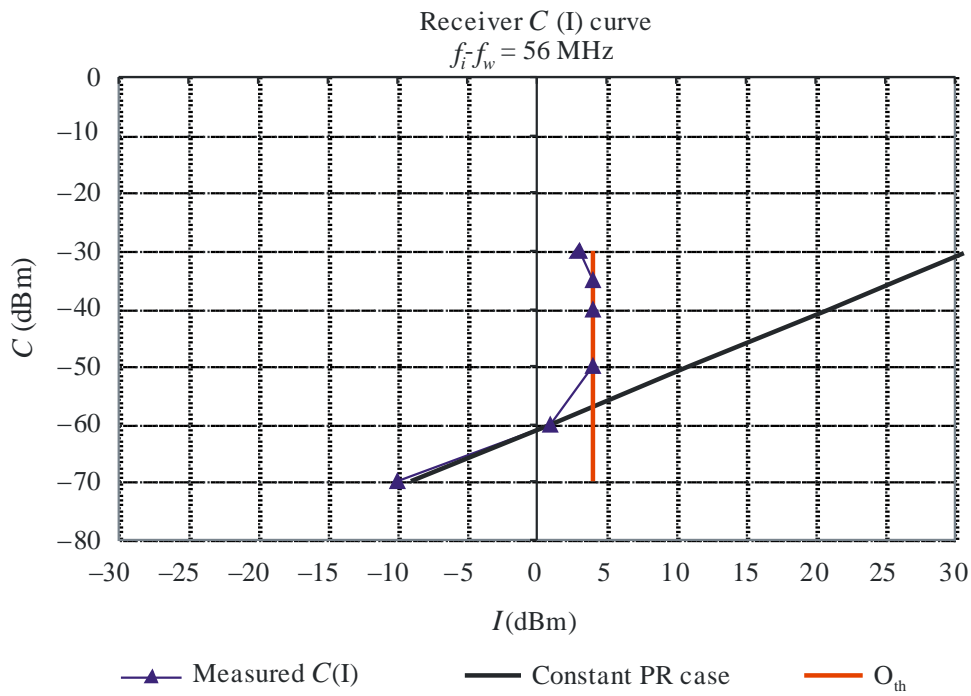
Report BT.2215-06

FIGURE 11  
 Some difficulties to determine the overload threshold;  
 PR = -39dB; O<sub>th</sub> = -6 dBm



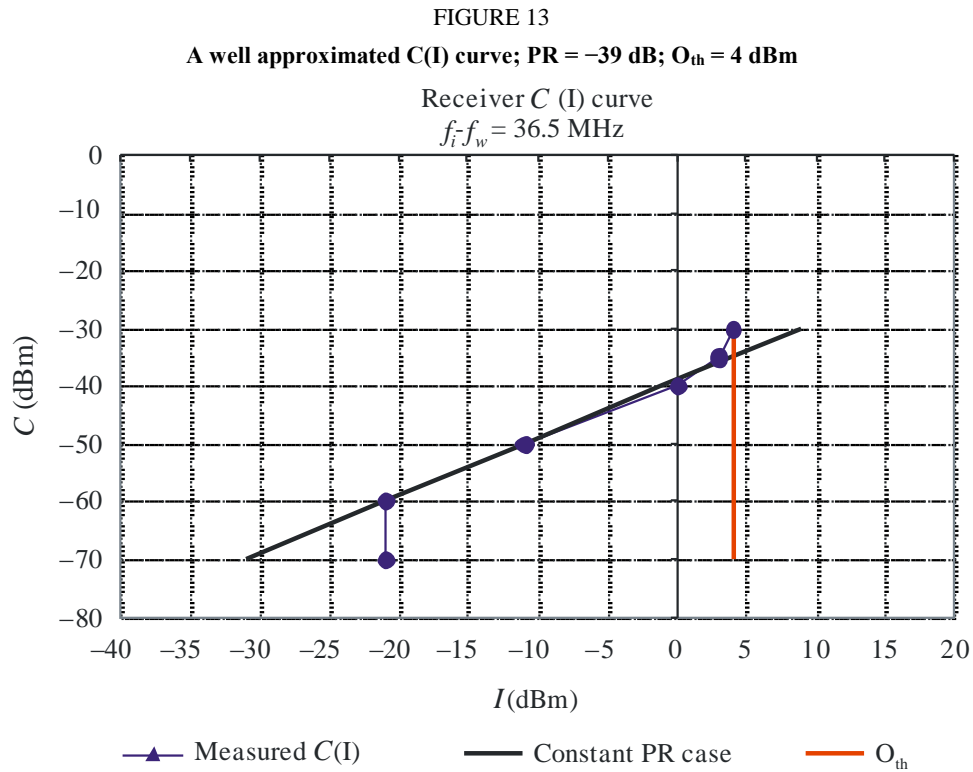
Report BT.2215-07

FIGURE 12  
 A well approximated C(I) curve; PR = -61 dB; O<sub>th</sub> = 4 dBm



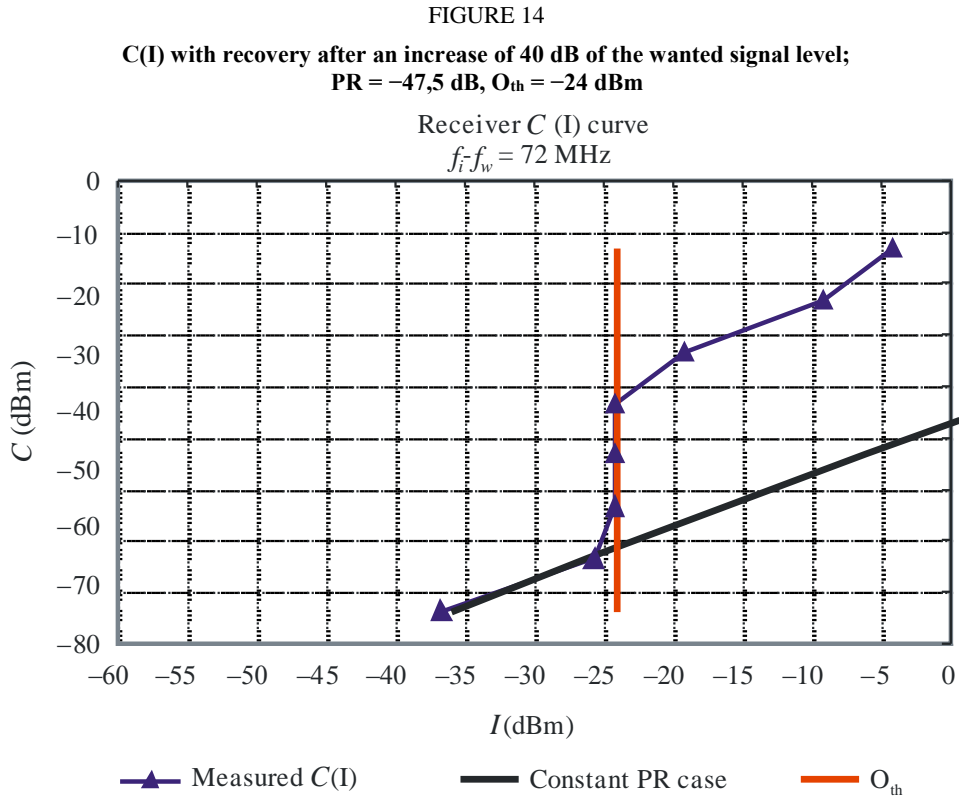
Report BT.2215-08

In the following example the receiver appears to behave in a non linear fashion when the interfering signal level reaches  $-21$  dBm, but is quite linear for higher interfering signal levels up to  $3$  dBm.



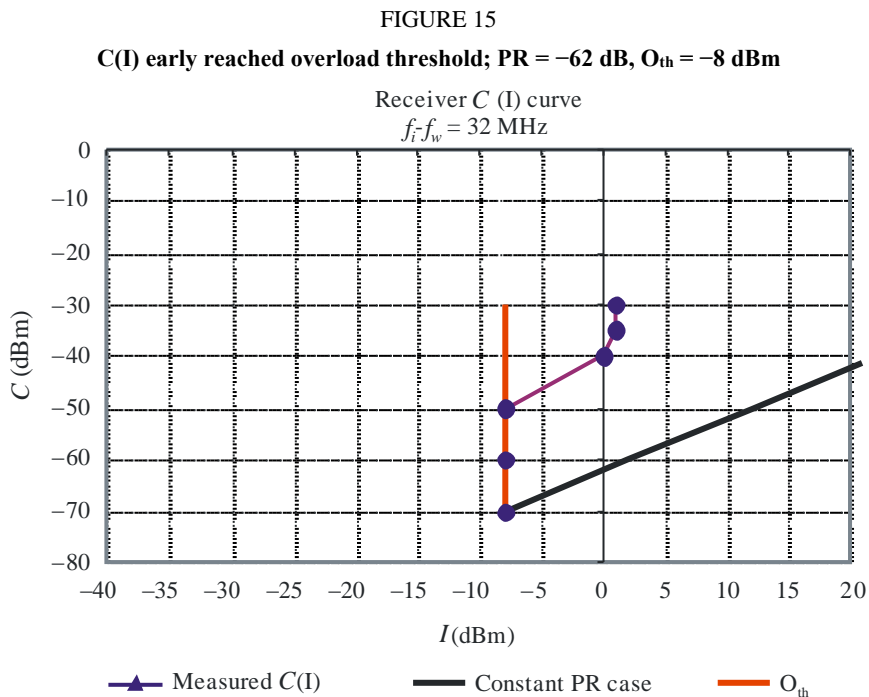
Report BT.2215-09

In the following example the  $O_{th}$  is reached at an interfering signal level of  $-24$  dBm. However, the measured  $C(I)$  curve shows that an increase of the wanted signal level by about  $30$  dB allows the receiver to behave linearly once again but with a reduced PR ( $-16$  dB instead of  $-48$  dB).

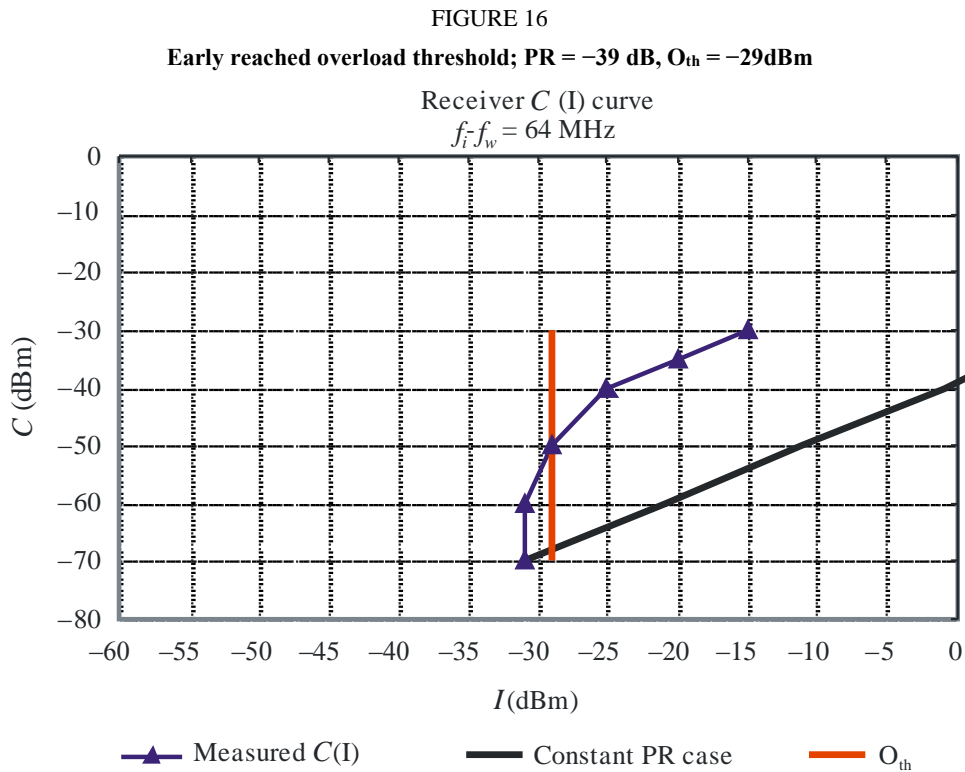


Report BT.2215-10

In the following examples the  $O_{th}$  is reached before the PR having been reached. In such cases the PR is obtained by subtracting  $I$  from  $C(I)$  at the lowest intersection point between the straight line with unity slope representing the constant PR case and the line vertical to  $I$ -axis representing the constant  $I$  case.



Report BT.2215-11



Report BT.2215-12

## 6 Conclusions and further work required

The results included in the annexes to this report are being used to improve and update Recommendation ITU-R BT.1368. They are also an opportunity to stimulate new areas of investigation for example the effects of time variation in interfering signals. Other Recommendations could also be considered as candidates for the material as required and a note added when required.

Where relevant the material in the annexes has a note indicating the version of Recommendation ITU-R BT.1368 which is amended, or another Recommendation where relevant. It is recommended that any future updates to this report add material together with similar notes of any particular version of a recommendation which relied on this information.

Additional results of receiver measurement tests are always welcome and are in some cases urgently needed for all digital TV broadcasting standards including DVB-T2, ISDB-T, ATSC, DTMB, so that the appropriate sections of Recommendation ITU-R BT.1368 can be filled with suitable PR and  $O_{th}$  data for assistance in network planning activities.

## Annex 1

### **DVB-T receiver performance in the presence of interfering signals from DVB-T, UMTS, and LTE**

Annex 1 contains measurements of protection ratios and overload thresholds for DVB-T receivers in the presence of various interfering signals including DVB-T, base stations (BS) and user equipment (UE) for UMTS, and BS and UE for LTE.

#### **A.1.1 Measurements of protection ratios and overload thresholds for DVB-T receivers under interference from DVB-T in adjacent channels**

Annex 1A contains measurements of protection ratios and overload thresholds for eight DVB-T receivers (four with “can” tuners and four with silicon tuners) under interference from DVB-T in adjacent channels. This information was used in the generation of Recommendation ITU-R BT.1368-8.



Annex 1A

#### **A.1.2 Measurements of protection ratio and overload threshold for DVB-T receivers under interference from UMTS BS and UE in co- and adjacent channels**

Annex 1B contains measurements of protection ratios and overload thresholds for seven DVB-T receivers (four with “can” tuners and three with silicon tuners) under interference from a UMTS BS and UE in adjacent channels. This information was used in the generation of Recommendation ITU-R BT.1368-8.



Annex 1B

Annex 1C contains graphical plots of protection ratios for DVB-T under interference from a UMTS BS and UE with and without TPC in co- and adjacent channels and with different TPC speed profiles. Only can tuners were tested. Note the test methodology may differ from the recommended methodology used for other measurements in this report, and for future measurements.



Annex 1C

Annex 1D contains measurements of protection ratios and overload thresholds for ten DVB-T receivers (all with “can” tuners) under interference from a UMTS BS and UE in co- and adjacent channels. Note the test methodology may differ from the recommended methodology used for other measurements in this report, and for future measurements.



Annex 1D

### A.1.3 Measurements of protection ratios and overload thresholds for DVB-T receivers under interference from LTE BS and UE in co- and adjacent channels

Annex 1E contains measurements of protection ratios and overload thresholds for 13 DVB-T receivers (nine with “can” tuners and four with silicon tuners) under interference from an LTE BS and UE in adjacent channels. These measurements use the recommended test methodology. This information was used in the generation of Recommendation ITU-R BT.1368-8.



Annex 1E

Annex 1F contains measurements of protection ratios and overload thresholds in adjacent channels for 20 DVB-T receivers (ten “can” tuners and ten silicon tuners) under interference from an LTE BS with three traffic loadings (idle, 50%, and 100%). These measurements use the recommended test methodology. This information was used in the generation of Recommendation ITU-R BT.1368-8.



Annex 1F

Annex 1G contains measurements of protection ratios in co- and adjacent channels for five DVB-T receivers (two “can” tuners and three silicon tuners) under interference from an LTE BS with two traffic loadings (idle and 100%). These measurements use the recommended test methodology.



Annex 1G

Annex 1H investigates the adjacent channel interfering effect of an LTE BS and UE in 17 DVB-T receivers. Protection ratios were measured for transmission channels which included Gaussian, static Rayleigh, and time-varying Rayleigh. Results for 90<sup>th</sup> percentiles are tabulated. Note the tests used the ESR<sub>5</sub> methodology which may differ from the recommended methodology used for other measurements in this report, and for future measurements.



Annex 1H

Annex 1I contains measurements of protection ratios and overload thresholds in adjacent channels for 14 DVB-T receivers (seven “can” tuners and seven silicon tuners) under interference from an LTE BS and UE with three traffic loadings (idle, 50%, and 100%). The measurements consider the reduced ACLR performance of the UE, compared to the BS, and the resulting increase in the protection ratios.



Annex 1I

Annex 1J contains measurements of protection ratios and overload thresholds in adjacent channels for five DVB-T receivers with silicon tuners under interference from an LTE UE with traffic loadings of 1 Mbps, 10 Mbps, and 20 Mbps. The measurements use the recommended test methodology and is the same as that used in Annex 1F.



Annex 1J

Annex 1K contains an updated analysis of the protection ratios and overload thresholds in adjacent channels from measurements made in Annexes 1I and 1J for DVB-T receivers under interference from an LTE UE. This analysis provided revisions to Tables 38 and 38A in Recommendation ITU-R BT.1368-9.



Annex 1K

Annex 1L contains measurements of protection ratios and overload thresholds for five DVB-T receivers under interference from an LTE UE in an adjacent channel with a 9 MHz guard band. The measurements were made with an LTE UE ACLR of 60 and 70 dB as well as a continuous and a discontinuous LTE UE transmission. The impact of a bandpass filter centred in the DVB-T channel was also measured.



Annex 1L



## Annex 2

### DVB-T2 receiver performance in the presence of interfering signals from DVB-T2 and LTE

Annex 2 contains measurements of protection ratios and overload thresholds for DVB-T2 receivers in the presence of various interfering signals including DVB-T2, base stations (BS) and user equipment (UE) for LTE, and wireless broadband access systems.

#### A.2.1 Measurements of protection ratios and overload thresholds for DVB-T2 receivers under interference from DVB-T2 in other channels

Annex 2A contains measurements of protection ratios and overload thresholds for 14 DVB-T2 receivers, three with “can” tuners and eleven with silicon tuners, under interference from DVB-T2 in adjacent channels.



Annex 2A

#### A.2.2 Measurements of protection ratios and overload thresholds for DVB-T2 receivers under interference from LTE BS in co- and adjacent channels

Annex 2B contains measurements of protection ratios and overload thresholds for 14 DVB-T2 receivers under interference from an LTE BS in adjacent channels. The measurements were made with the same receivers as those used in Annex 2A. The protection ratios have been corrected for the actual LTE BS ACLR values.



Annex 2B

Annex 2C compares the protection ratios and overload threshold measurements of DVB-T2 receivers at 690 MHz, 754 MHz and 786 MHz with LTE BS interference in adjacent channels. Four DVB-T2 receivers (two with “can” tuners and two with silicon tuners) and the test setup described in Annex 2B were used.



Annex 2C

Annex 2D contains measurements of protection ratios for nine DVB-T2 receivers (four with “can” tuners and five with silicon tuners) under interference in co- and adjacent channels from an LTE BS with traffic loadings at idle and 100 percent. Measurements were made at three DVB-T2 signal levels (-70, -50, and -30 dBm). The test setup and interfering signals are described in Annex 1G.



Annex 2D

s



Annex 2E

Annex 2F contains measurements of protection ratios and overload thresholds for 14 DVB-T2 receivers under interference in adjacent channels from an LTE UE with traffic loadings of 1 Mbps, 10 Mbps, and 20 Mbps. The measurements were made with the same receivers as those used in Annex 2A. The degradation of the protection ratios due to the ACLR (75, 85, and 90 dB) of the LTE UE was evaluated.



Annex 2F

Annex 2G contains measurements of protection ratios for three DVB-T2 receivers with different silicon tuners under interference in adjacent channels from an LTE UE with full traffic loading. The tests evaluated the performance with and without filtering at the DVB-T2 receiver input as well as the LTE UE output. The results include calculation of the receiver adjacent channel selectivity (ACS) as well as calculation of the minimum separation distance for fixed rooftop and portable indoor DTT reception.



Annex 2G

Annex 2H contains measurements of protection ratios and overload thresholds for five DVB-T2 receivers under interference from an LTE UE in an adjacent channel with a 9 MHz guard band. The measurements were made with an LTE UE ACLR of 60 and 70 dB as well as a continuous and a discontinuous LTE UE transmission. The impact of a bandpass filter centred in the DVB-T channel was also measured. The test setup and LTE UE signals are described in Annex 1L.



Annex 2H

#### **A.2.4 Measurements of protection ratios and overload thresholds for DVB-T2 receivers under interference from wireless broadband access systems in co- and adjacent channels**

Annex 2I contains measurements of protection ratios for three DVB-T2 test receivers under interference from a simulated wireless broadband access system operating in co- and adjacent channels. Field measurements were made in addition to the initial laboratory tests.



Annex 2I

## Annex 3

### ATSC receiver performance in the presence of interfering signals from ATSC

Annex 3 contains measurements of protection ratios and receiver desensitisation for ATSC receivers in the presence of various interfering signals including ATSC.

#### A.3.1 Measurements of protection ratios and receiver desensitisation for ATSC receivers under interference from ATSC on adjacent channels

Annex 3A contains measurements of protection ratios and receiver desensitization for 48 ATSC receivers (26 “can”, 18 silicon, and four indeterminate) under interference from ATSC signals on adjacent channel pairs. Protection ratios for specific wanted ATSC signal levels (Weak: -68dBm and Moderate: -53 dBm) are interpolated.



Annex 3A

Annex 3B contains measurements of protection ratios for 12 ATSC receivers under interference from single and multiple ATSC signals on adjacent channels. Protection ratios at wanted ATSC signal levels of -68 dBm (Weak) and -78 dBm (Very Weak) were measured. The results confirm that ATSC receivers, in general, meet the ATSC performance targets. In addition, the selectivity of ATSC receivers can be degraded by the presence of third order intermodulation products.



Annex 3B

Annex 3C contains measurements of protection ratio thresholds and receiver desensitization for 24 ATSC receivers (six “can”, 14 silicon, and four indeterminate tuners) under interference from ATSC signals on adjacent channels. The results demonstrate that the protection ratios in Recommendation ITU-R BT.1368 are adequate except for weak wanted signals. Significantly increases in signal field strength are required in order to maintain reception in the presence of adjacent channel interference.



Annex 3C

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